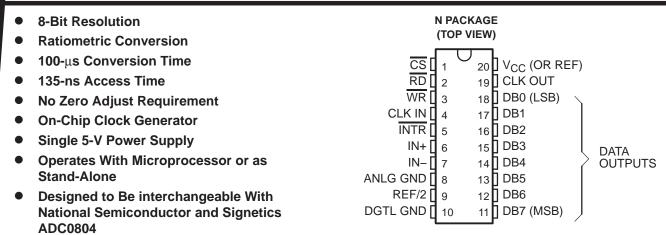
SLAS035 - OCTOBER 1983 - REVISED OCTOBER 1988



### description

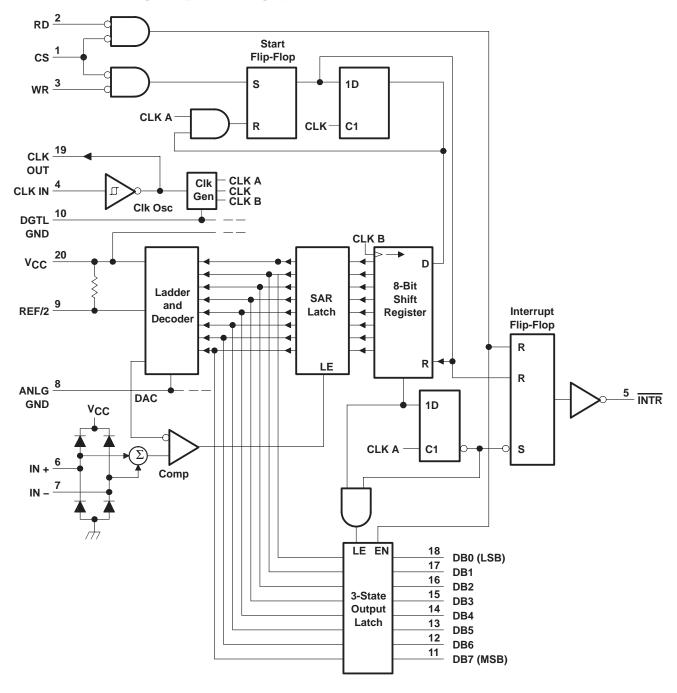
The ADC0804 is a CMOS 8-bit successive-approximation analog-to-digital converter that uses a modified potentiometric (256R) ladder. The ADC0804 is designed to operate from common microprocessor control buses, with the 3-state output latches driving the data bus. The ADC0804 can be made to appear to the microprocessor as a memory location or an I/O port. Detailed information on interfacing to most popular microprocessors is readily available from the factory.

A differential analog voltage input allows increased common-mode rejection and offset of the zero-input analog voltage value. Although REF/2 is available to allow 8-bit conversion over smaller analog voltage spans or to make use of an external reference, ratiometric conversion is possible with REF/2 open. Without an external reference, the conversion takes place over a span from  $V_{CC}$  to ANLG GND. The ADC0804 can operate with an external clock signal or, with an additional resistor and capacitor, can operate using an on-chip clock generator.

The ADC0804C is characterized for operation from 0°C to 70°C. The ADC0804I is characterized for operation from –40°C to 85°C.

SLAS035 - OCTOBER 1983 - REVISED OCTOBER 1988

## functional block diagram (positive logic)



SLAS035 - OCTOBER 1983 - REVISED OCTOBER 1988

## absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Supply voltage, V <sub>CC</sub> (see Note 1)	6.5 V
Input voltage range: CS, RD, WR	
Other inputs	. $-0.3 \text{ V}$ to $V_{CC}$ + $0.3 \text{ V}$
Output voltage range	$-0.3 \text{ V to V}_{CC} + 0.3 \text{ V}$
Operating free-air temperature range: ADC0804C	0°C to 70°C
ADC0804I	40°C to 85°C
Storage temperature range	65°C to 150°C
Lead temperature 1,6 mm (1/16 inch) from case for 10 seconds	260°C

NOTE 1: All voltage values are with respect DGTL GND with DGTL GND and ANLG GND connected together (unless otherwise noted.)

### recommended operating conditions

		MIN	NOM	MAX	UNIT
Supply voltage, V <sub>CC</sub>		4.5	5	6.3	V
Voltage at REF/2, V <sub>REF/2</sub> (see Note 2),		0.25	2.5		V
High-level input voltage at $\overline{\text{CS}}$ , $\overline{\text{RD}}$ , or $\overline{\text{WR}}$ , $V_{\text{IH}}$		2		15	V
Low-level input voltage at CS, RD, or WR, V <sub>IL</sub>				0.8	V
Analog ground voltage (see Note 3)			0	1	V
Analog input voltage (see Note 4)		-0.05		V <sub>CC</sub> + 0.05	V
Clock iput frequency, f <sub>clock</sub> (see Note 5)		100	640	1460	kHz
outy cycle for f <sub>clock ≥</sub> 640 kHz (see Note 5)				60	%
Pulse durartion, clock input (high or low) for f <sub>clock</sub> < 640 kHz, t <sub>W</sub> (CLK) (see Note 5)		275	781		ns
Pulse durartion, WR input low, (start conversion), t <sub>W</sub> (WR)					ns
Operating free air temperature T.	ADC0804C	0		70	°C
Operating free-air temperature, T <sub>A</sub>	ADC0804I	-40		85	-0

- NOTES: 2. The internal reference voltage is equal to the voltage applied to REF/2 or approximately equal to one-half of the V<sub>CC</sub> when REF/2 is left open. The voltage at REF/2 should be one-half the full-scale differential input voltage between the analog inputs. Thus, the differential input voltage range when REF/2 is open and V<sub>CC</sub> = 5 V is 0 V to 5 V. V<sub>REF/2</sub> for an input voltage range from 0.5 V to 3.5 V (full-scale differential voltage of 3 V) is 1.5 V.
  - 3. These values are with respect to DGTL GND.
  - 4. When the differential input voltage  $(V_{||+} V_{||-})$  is less than or equal to 0 V, the output code is 0000 0000.
  - 5. Total unadjusted error is specified only at an f<sub>Clock</sub> of 640 kHz with a duty cycle of 40% to 60% (pulse duration 625 ns to 937 ns). For frequencies above this limit or pulse duration below 625 ns, error may increase. The duty cycle limits should be observed for an f<sub>Clock</sub> greater than 640 kHz. Below 640 kHz, this duty cycle limit can be exceeded provided t<sub>W(CLK)</sub> remains within limits.



SLAS035 - OCTOBER 1983 - REVISED OCTOBER 1988

## electrical characteristics over recommended range of operating free-air temperature, $V_{CC} = 5 \text{ V}$ , $f_{clock} = 640 \text{ kHz}$ , $V_{REF/2} = 2.5 \text{ V}$ (unless otherwise noted)

PARAMETER			TEST CONDITIONS			TYP <sup>†</sup>	MAX	UNIT
	High-level output voltage	All outputs	$V_{CC} = 4.75 \text{ V},$	I <sub>OH</sub> = -360 μA	2.4			V
VOH		DB and INTR	$V_{CC} = 4.75 \text{ V},$	$I_{OH} = -10  \mu A$	4.5			
VOL	Low-level output voltage	Data outputs	$V_{CC} = 4.75 \text{ V},$	I <sub>OL</sub> = 1.6 mA			0.4	V
		INTR output	$V_{CC} = 4.75 \text{ V},$	I <sub>OL</sub> = 1 mA			0.4	
		CLK OUT	$V_{CC} = 4.75 \text{ V},$	I <sub>OL</sub> = 360 μA			0.4	
V <sub>T+</sub>	Clock positive-going threshold voltage				2.7	3.1	3.5	V
V <sub>T</sub> –	Clock negative-going thresho	ld voltage			1.5	1.8	2.1	V
$V_{T+}-V_{T-}$	Clock input hysteresis				0.6	1.3	2	V
lΗ	High-level input current					0.005	1	μΑ
I <sub>IL</sub>	Low-level input current					-0.005	-1	μΑ
			V <sub>O</sub> = 0				-3	
IOZ	IOZ Off-state output current		V <sub>O</sub> = 5 V				3	μΑ
I <sub>OHS</sub>	Short-current output current	Output high	$V_{O} = 0$ ,	T <sub>A</sub> = 25°C	-4.5	-6		mA
lols	Short-circuit output current	Output low	$V_0 = 5 V$ ,	T <sub>A</sub> = 25°C	9	16		mA
ICC	Supply current plus reference current		V <sub>REF/2</sub> = open,	$T_A = 25^{\circ}C$ , $\overline{CS} = 5 \text{ V}$		1.9	2.5	mA
R <sub>REF/2</sub>	Input resistance to reference ladder		See Note 6		1	1.3		kΩ
Ci	Input capacitance (control)					5	7.5	pF
Co	Output capacitance (DB)					5	7.5	pF

## operating characteristics over recommended operating free-air temperature, $V_{CC}$ = 5 V, $V_{REF/2}$ = 2.5 V, $f_{clock}$ = 640 kHz (unless otherwise noted)

PARAMETER		TEST CONDITI	MIN	TYP†	MAX	UNIT	
	Supply-voltage-variation error (see Notes 2 and 7)	$V_{CC} = 4.5 \text{ V to } 5.5 \text{ V}$			±1/16	±1/8	LSB
	Total unadjusted error (see Notes 7 and 8)	V <sub>REF/2</sub> = 2.5 V				±1	LSB
	DC common-mode error (see Note 8)				±1/16	±1/8	LSB
t <sub>en</sub>	Output enable time	C <sub>L</sub> = 100 pF			135	200	ns
tdis	Output disable time	$C_L = 10 pF$ ,	$R_L = 10 \text{ k}\Omega$		125	200	ns
td(INTR)	Delay time to reset INTR				300	450	nx
tconv	Conversion cycle time (see Note 9)	f <sub>clock</sub> = 100 kHz to 1.46 MHz		65 1/2		72 1/2	clock cycles
	Conversion time			103		114	μs
CR	Free-running conversion rate	INTR connected to WR,	CS at 0 V			8827	conv/s

<sup>†</sup> All typical values are at  $T_A = 25$ °C.

NOTES: 2. The internal reference voltage is equal to the voltage applied to REF/2 or approximately equal to one-half of the V<sub>CC</sub> when REF/2 is left open. The voltage at REF/2 should be one-half the full-scale differential input voltage between the analog inputs. Thus, the differential input voltage when REF/2 is open and V<sub>CC</sub> = 5 V is 0 to 5 V. V<sub>REF</sub>/2 for an input voltage range from 0.5 V to 3.5 V (full-scale differential voltage of 3 V) is 1.5 V.

- 6. The resistance is calculated from the current drawn from a 5-V supply applied to ANLG GND and REF/2.
- 7. These parameters are specified for the recommended analog input voltage range.
- 8. All errors are measured with reference to an ideal straight line through the end points of the analog-to-digital transfer characteristic
- Although internal conversion is completed in 64 clock periods, a CS or WR low-to-high transition is followed by 1 to 8 clock periods
  <u>before</u> conversion starts. After conversion is completed, part of another clock period is required before a high-to-low transition of
  <u>INTR</u> completes the cycle.



SLAS035 - OCTOBER 1983 - REVISED OCTOBER 1988

### PARAMETER MEASUREMENT INFORMATION

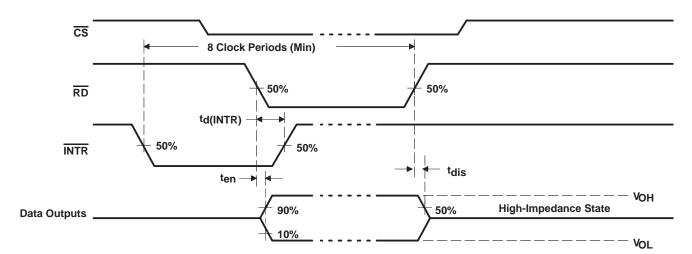
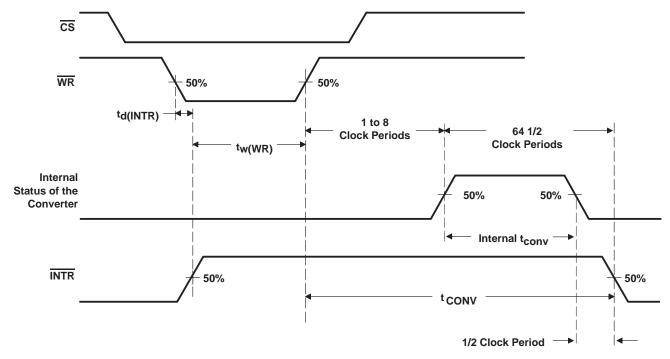


Figure 1. Read Operation Timing Diagram



**Figure 2. Write Operation Timing Diagram** 

SLAS035 - OCTOBER 1983 - REVISED OCTOBER 1988

#### PRINCIPLES OF OPERATION

The ADC0804 contains a circuit equivalent to a 256-resistor network. Analog switches are sequenced by successive-approximation logic to match an analog differential input voltage  $(V_{I+} - V_{I-})$  to a corresponding tap on the 256-resistor network. The most significant bit (MSB) is tested first. After eight comparisons (64 clock periods), an 8-bit binary code (1111 1111 = full scale) is transferred to an output latch and the interrupt ( $\overline{\text{INTR}}$ ) output goes low. The device can be operated in a free-running mode by connecting the  $\overline{\text{INTR}}$  output to the write ( $\overline{\text{WR}}$ ) input and holding the conversion start ( $\overline{\text{CS}}$ ) input at a low level. To ensure startup under all conditions, a low-level  $\overline{\text{WR}}$  input is required during the power-up cycle. Taking  $\overline{\text{CS}}$  low anytime after that will interrupt a conversion in process.

When WR goes low, the ADC0804 successive-approximation register (SAR) and 8-bit shift register are reset. As long as both  $\overline{CS}$  and  $\overline{WR}$  remain low, the ADC0804 remains in a reset state. One to eight clock periods after  $\overline{CS}$  or  $\overline{WR}$  makes a low-to-high transition, conversion starts.

When  $\overline{\text{CS}}$  and  $\overline{\text{WR}}$  are low, the start flip-flop is set and the interrupt flip-flop and 8-bit register are reset. The next clock pulse transfers a logic high to the output of the start flip-flop. The logic high is ANDed with the next clock pulse, placing a logic high on the reset input of the start flip-flop. If either  $\overline{\text{CS}}$  or  $\overline{\text{WR}}$  have gone high, the set signal to the start flip-flop is removed, causing it to be reset. A logic high is placed on the D input of the 8-bit shift register and the conversion process is started. If  $\overline{\text{CS}}$  and  $\overline{\text{WR}}$  are still low, the start flip-flop, the 8-bit shift register, and the SAR remain reset. This action allows for wide  $\overline{\text{CS}}$  and  $\overline{\text{WR}}$  inputs with conversion starting from one to eight clock periods after one of the inputs goes high.

When the logic high input has been clocked through the 8-bit shift register, completing the SAR search, it is applied to an AND gate controlling the output latches and to the D input of a flip-flop. On the next clock pulse, the digital word is transferred to the 3-state output latches and the interrupt flip-flop is set. The output of the interrupt flip-flop is inverted to provide an INTR output that is high during conversion and low when the conversion is completed.

When a low is at both  $\overline{CS}$  and  $\overline{RD}$ , an output is applied to the DB0 through DB7 outputs and the interrupt flip-flop is reset. When either the  $\overline{CS}$  or  $\overline{RD}$  inputs return to a high state, the DB0 through DB7 outputs are disabled (returned to the high-impedance state). The interrupt flip-flop remains reset.



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