

Burr-Brown Products from Texas Instruments



DAC7642 DAC7643

SBAS233 - DECEMBER 2001

16-Bit, Dual Voltage Output DIGITAL-TO-ANALOG CONVERTER

FEATURES

- LOW POWER: 4mW
- UNIPOLAR OR BIPOLAR OPERATION
- SETTLING TIME: 10µs to 0.003% FSR
- 15-BIT LINEARITY AND MONOTONICITY: -40°C to +85°C
- RESET TO MID-SCALE (DAC7642) OR ZERO-SCALE (DAC7643)
- DATA READBACK
- DOUBLE-BUFFERED DATA INPUTS

APPLICATIONS

- PROCESS CONTROL
- CLOSED-LOOP SERVO-CONTROL
- MOTOR CONTROL
- DATA ACQUISITION SYSTEMS
- DAC-PER-PIN PROGRAMMERS

DESCRIPTION

The DAC7642 and DAC7643 are dual channel, 16-bit, voltage output Digital-to-Analog Converters (DACs) which provide 15-bit monotonic performance over the specified temperature range. They accept 16-bit parallel input data, have double-buffered DAC input logic (allowing simultaneous update of all DACs), and provide a readback mode of the internal input registers. Programmable asynchronous reset clears all registers to a mid-scale code of $8000_{\rm H}$ (DAC7642) or to a zero-scale code of $0000_{\rm H}$ (DAC7643). These DACs can operate from a single +5V supply or from +5V and -5V supplies, providing an output range of 0 to +2.5V or -2.5V to +2.5V, respectively.

Low power and small size per DAC make the DAC7642 and DAC7643 ideal for automatic test equipment, DAC-per-pin programmers, data acquisition systems, and closed-loop servo-control. The DAC7642 and DAC7643 are available in a LQFP-32 package and specified over a -40° C to $+85^{\circ}$ C temperature range.





Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.



ABSOLUTE MAXIMUM RATINGS⁽¹⁾

| | 0.01/ += 441/ |
|--|---------------------------------|
| V _{CC} to V _{SS} | |
| V _{CC} to GND | –0.3V to 5.5V |
| V _{REF} L to V _{SS} | $-0.3V$ to $(V_{CC} - V_{SS})$ |
| V _{CC} to V _{REF} H | $-0.3V$ to $(V_{CC} - V_{SS})$ |
| V _{REF} H to V _{REF} L | $-0.3V$ to $(V_{CC} - V_{SS})$ |
| Digital Input Voltage to GND | –0.3V to V _{CC} + 0.3V |
| Digital Output Voltage to GND | –0.3V to V _{CC} + 0.3V |
| Maximum Junction Temperature | +150°C |
| Operating Temperature Range | 40°C to +85°C |
| Storage Temperature Range | –65°C to +125°C |
| Lead Temperature (soldering, 10s) | +300°C |

NOTE: (1) Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. Exposure to absolute maximum conditions for extended periods may affect device reliability.

PACKAGE/ORDERING INFORMATION



This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

| PRODUCT | MONOTONICITY | PACKAGE-LEAD | PACKAGE DESIGNATOR ⁽¹⁾ | SPECIFIED TEMPERATURE RANGE | PACKAGE MARKING | ORDERING NUMBER | TRANSPORT MEDIA, QUANTITY |
|------------|--------------|--------------|--------------------------------------|-----------------------------------|--------------------|--------------------|------------------------------|
| DAC7642VF | 14 Bits | LQFP-32 | VF | –40°C to +85°C | DAC7642 | DAC7642VFT | Tape and Reel, 250 |
| " | " | " | " | " | " | DAC7642VFR | Tape and Reel, 1000 |
| DAC7642VFB | 15 Bits | LQFP-32 | VF | –40°C to +85°C | DAC7642B | DAC7642VFB T | Tape and Reel, 250 |
| " | " | " | " | " | " | DAC7642VFB R | Tape and Reel, 1000 |
| DAC7643VF | 14 Bits | LQFP-32 | VF | –40°C to +85°C | DAC7643 | DAC7643VFT | Tape and Reel, 250 |
| " | " | " | " | " | " | DAC7643VFR | Tape and Reel, 1000 |
| DAC7643VFB | 15 Bits | LQFP-32 | VF | –40°C to +85°C | DAC7643B | DAC7643VFBT | Tape and Reel, 250 |
| " | " | " | " | " | " | DAC7643VFBR | Tape and Reel, 1000 |

NOTE: (1) For the most current specifications and package information, refer to our web site at www.ti.com.



ELECTRICAL CHARACTERISTICS (Dual Supply)

At $T_A = T_{MIN}$ to T_{MAX} , $V_{CC} = +5V$, $V_{SS} = -5V$, $V_{REF}H = +2.5V$, and $V_{REF}L = -2.5V$, unless otherwise noted.

| | | | DAC7642V | | | AC7642VF AC7643VF | | |
|---|---|-----------------------------------|--|--|--------|--|--|--|
| PARAMETER | CONDITIONS | MIN | ТҮР | MAX | MIN | ТҮР | МАХ | UNITS |
| ACCURACY Linearity Error Linearity Match Differential Linearity Error Monotonicity, T _{MIN} to T _{MAX} Bipolar Zero Error Bipolar Zero Error Drift Full-Scale Error Full-Scale Error Drift Bipolar Zero Matching | Channel-to-Channel Matching | 14 | ±3 ±4 ±2 ±1 5 ±1 5 ±1 | ±4 ±3 ±3 10 ±3 10 ±3 10 ±3 | 15 | +2 +2 +1 * * * * + 1 | ±3 ±2 * * * * * * | LSB LSB Bits mV ppm/°C mV ppm/°C mV |
| Full-Scale Matching Power-Supply Rejection Ratio (PSRR) | Channel-to-Channel Matching At Full-Scale | | ±1 10 | ±3 100 | | ±1 * | ±3 * | mV ppm/V |
| ANALOG OUTPUT Voltage Output Output Current Maximum Load Capacitance Short-Circuit Current Short-Circuit Duration | $R_L = 10k\Omega$ No Oscillation GND, V _{CC} or V _{SS} | V _{REF} L -1.25 | 500 -10, +30 Indefinite | V _{REF} H +1.25 | * | * * | * * | V mA pF mA |
| REFERENCE INPUT Ref High Input Voltage Range Ref Low Input Voltage Range Ref High Input Current Ref Low Input Current | | V _{REF} L + 1.25 -2.5 | 500 500 | +2.5 V _{REF} H – 1.25 | * * | * * | * * | ν ν μΑ μΑ |
| DYNAMIC PERFORMANCE Settling Time Channel-to-Channel Crosstalk Digital Feedthrough Output Noise Voltage DAC Glitch | To ±0.003%, 5V Output Step See Figure 5 f = 10kHz 7FFF _H to 8000 _H or 8000 _H to 7FFF _H | | 8 0.5 2 60 40 | 10 | | * * * * | * | µs LSB nV-s nV/√Hz nV-s |
| DIGITAL INPUT V _{IH} V _{IL} I _{IH} I _{IL} | | 0.7 • V _{CC} | | 0.3 • V _{CC} ±10 ±10 | * | | * * * | V V μΑ μΑ |
| DIGITAL OUTPUT V _{OH} V _{OL} | I _{OH} = -0.8mA I _{OL} = 1.2mA | 3.6 | 4.5 0.3 | 0.4 | * | * * | * | V V |
| POWER SUPPLY V _{CC} V _{SS} I _{CC} I _{SS} Power | | +4.75 -5.25 -1.2 | +5.0 -5.0 0.7 -0.8 7.5 | +5.25 -4.75 1.1 11.5 | * * | * * * * | * * * | V V mA mA mW |
| TEMPERATURE RANGE Specified Performance | | -40 | | +85 | * | | * | °C |

* Specifications same as DAC7642VF and DAC7643VF.



ELECTRICAL CHARACTERISTICS (Single Supply)

At $T_A = T_{MIN}$ to T_{MAX} , $V_{CC} = +5V$, $V_{SS} = 0V$, $V_{REF}H = +2.5V$, and $V_{REF}L = 0V$, unless otherwise noted.

| | | | DAC7642V | | | AC7642VF AC7643VF | | |
|---|--|--------------------------------|---|--|--------|---|--|--|
| PARAMETER | CONDITIONS | MIN | TYP | МАХ | MIN | TYP | MAX | UNITS |
| ACCURACY Linearity Error ⁽¹⁾ Linearity Match Differential Linearity Error Monotonicity, T _{MIN} to T _{MAX} Zero-Scale Error Zero-Scale Error Drift Full-Scale Error Drift Zero-Scale Error Drift Zero-Scale Matching Full-Scale Matching | Channel-to-Channel Matching Channel-to-Channel Matching | 14 | ±3 ±4 ±2 ±1 5 ±1 5 ±1 5 ±1 | ±4 ±3 ±3 10 ±3 10 ±3 ±3 | 15 | +2 +2 +1 * * * +1 +1 | ±3 ±2 * * * * ±3 ±3 | LSB LSB Bits mV ppm/°C mV ppm/°C mV mV |
| Power-Supply Rejection Ratio (PSRR) | At Full-Scale | | 10 | ±3 100 | | * | ±3 * | ppm/V |
| ANALOG OUTPUT Voltage Output Output Current Maximum Load Capacitance Short-Circuit Current Short-Circuit Duration | $R_L = 10k\Omega$ No Oscillation GND or V _{CC} | 0 -1.25 | 500 -10, +30 Indefinite | V _{REF} H +1.25 | * | * * * | * * | V mA pF mA |
| REFERENCE INPUT Ref High Input Voltage Range Ref Low Input Voltage Range Ref High Input Current Ref Low Input Current | | V _{REF} L + 1.25 0 | 250 -250 | +2.5 V _{REF} H – 1.25 | * * | * * | * * | V V μΑ μΑ |
| DYNAMIC PERFORMANCE Settling Time Channel-to-Channel Crosstalk Digital Feedthrough Output Noise Voltage, f = 10kHz DAC Glitch | To ±0.003%, 2.5V Output Step See Figure 6 7FFF _H to 8000 _H or 8000 _H to 7FFF _H | | 8 0.5 2 60 40 | 10 | | * * * * | * | µs LSB nV-s nV/√Hz nV-s |
| DIGITAL INPUT V _{IH} V _{IL} I _{IH} I _{IL} | | 0.7 • V _{CC} | | 0.3 • V _{CC} ±10 ±10 | * | | * * * | V V μΑ μΑ |
| DIGITAL OUTPUT V _{OH} V _{OL} | I _{OH} = -0.8mA I _{OL} = 1.2mA | 3.6 | 4.5 0.3 | 0.4 | * | * * | * | V V |
| POWER SUPPLY V _{CC} V _{SS} I _{CC} Power | | +4.75 0 | +5.0 0 0.5 2.5 | +5.25 0 0.9 4.5 | * * | * * * * | * * * * | V V mA mW |
| TEMPERATURE RANGE Specified Performance | | -40 | | +85 | * | | * | °C |

* Specifications same as DAC7642VF and DAC7643VF.

NOTE: (1) If $V_{SS} = 0V$, specification applies at Code 0040_H and above due to possible negative zero-scale error.

PIN CONFIGURATION



PIN DESCRIPTIONS

| PIN | NAME | DESCRIPTION | PIN | NAME | DESCRIPTION |
|-----|-----------------|-------------------------|-----|--------------------------|--|
| 1 | V _{cc} | Positive Power Supply | 20 | R/W | Enabled by $\overline{\text{CS}}$, Controls Data Read from and Write |
| 2 | GND | Ground | | | to the Input Registers. |
| 3 | DB15 | Data Bit 15, MSB | 21 | LOADDACS | DAC Output Registers Load Control. Rising edge |
| 4 | DB14 | Data Bit 14 | | | triggered. Transfers Data from the Input Registers to the DAC Registers, Updating the DAC Output. |
| 5 | DB13 | Data Bit 13 | 22 | RST | Reset, Rising Edge Triggered. DAC7642 resets to |
| 6 | DB12 | Data Bit 12 | | | mid-scale, DAC7643 resets to zero. (Resets Both |
| 7 | DB11 | Data Bit 11 | | | Input Registers and DAC Registers) |
| 8 | DB10 | Data Bit 10 | 23 | DACSEL | Enabled by \overline{CS} . Selects the individual DAC Input Registers. (LOW Selects Register A, HIGH Selects |
| 9 | DB9 | Data Bit 9 | | | Register B) |
| 10 | DB8 | Data Bit 8 | 24 | V _{SS} | Negative Power Supply |
| 11 | DB7 | Data Bit 7 | 25 | V _{OUT} B | DAC B Voltage Output |
| 12 | DB6 | Data Bit 6 | 26 | V _{OUT} B Sense | DAC B Output Amplifier Inverting Input. Used to |
| 13 | DB5 | Data Bit 5 | | | close the feedback loop at the load. |
| 14 | DB4 | Data Bit 4 | 27 | V _{REF} H Sense | DAC A and B Reference High Sense Input |
| 15 | DB3 | Data Bit 3 | 28 | V _{REF} H | DAC A and B Reference High Input |
| 16 | DB2 | Data Bit 2 | 29 | V _{OUT} L | DAC A and B Reference Low Input |
| 17 | DB1 | Data Bit 1 | 30 | V _{REF} L Sense | DAC A and B Reference Low Sense Input |
| 18 | DB0 | Data Bit 0, LSB | 31 | V _{OUT} A Sense | DAC A Output Amplifier Inverting Input. Used to close the feedback loop at the load. |
| 19 | CS | Chip Select, Active LOW | 32 | V _{OUT} A | DAC A Output Voltage |



TYPICAL CHARACTERISTICS: $V_{SS} = 0V$

At $T_A = +25^{\circ}C$, $V_{CC} = +5V$, $V_{SS} = 0V$, $V_{REF}H = +2.5V$, $V_{REF}L = 0V$, representative unit, unless otherwise specified.



TYPICAL CHARACTERISTICS: V_{SS} = 0V (Cont.)

At $T_A = +25^{\circ}C$, $V_{CC} = +5V$, $V_{SS} = 0V$, $V_{REF}H = +2.5V$, $V_{REF}L = 0V$, representative unit, unless otherwise specified.





TYPICAL CHARACTERISTICS: $V_{SS} = 0V$ (Cont.)

At $T_A = +25^{\circ}C$, $V_{CC} = +5V$, $V_{SS} = 0V$, $V_{REF}H = +2.5V$, $V_{REF}L = 0V$, representative unit, unless otherwise specified.



















TYPICAL CHARACTERISTICS: V_{SS} = 0V (Cont.)

At $T_A = +25^{\circ}C$, $V_{CC} = +5V$, $V_{SS} = 0V$, $V_{REF}H = +2.5V$, $V_{REF}L = 0V$, representative unit, unless otherwise specified.



$V_{SS} = -5V$

At $T_A = +25^{\circ}C$, $V_{CC} = +5V$, $V_{SS} = -5V$, $V_{REF}H = +2.5V$, $V_{REF}L = -2.5V$, representative unit, unless otherwise specified.







 LINEARITY ERROR AND DIFFERENTIAL LINEARITY ERROR vs CODE (DAC B, +25°C)

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Digital Input Code





TYPICAL CHARACTERISTICS: $V_{SS} = -5V$ (Cont.)

At $T_A = +25^{\circ}C$, $V_{CC} = +5V$, $V_{SS} = -5V$, $V_{REF}H = +2.5V$, $V_{REF}L = -2.5V$, representative unit, unless otherwise specified.





TYPICAL CHARACTERISTICS: $V_{SS} = -5V$ (Cont.)

At $T_A = +25^{\circ}C$, $V_{CC} = +5V$, $V_{SS} = -5V$, $V_{REF}H = +2.5V$, $V_{REF}L = -2.5V$, representative unit, unless otherwise specified.





TYPICAL CHARACTERISTICS: $V_{SS} = -5V$ (Cont.)

At $T_A = +25^{\circ}C$, $V_{CC} = +5V$, $V_{SS} = -5V$, $V_{REF}H = +2.5V$, $V_{REF}L = -2.5V$, representative unit, unless otherwise specified.





THEORY OF OPERATION

The DAC7642 and DAC7643 are dual channel, voltage output, 16-bit DACs. The architecture is an R-2R ladder configuration with the three MSB's segmented followed by an operational amplifier that serves as a buffer. Each DAC has its own R-2R ladder network, segmented MSBs, and output op amp, as shown in Figure 1. The minimum voltage output (zero-scale) and maximum voltage output (full-scale) are set

by the external voltage references $V_{REF}L$ and $V_{REF}H$, respectively. The digital input is a 16-bit parallel word and the DAC input registers offer a readback capability. The converters can be powered from either a single +5V supply or a dual ±5V supply. Each device offers a reset function which immediately sets all DAC output voltages, DAC registers and Input registers to mid-scale, code $8000_{\rm H}$ (DAC7642), or to zero-scale, code $0000_{\rm H}$ (DAC7643). See Figures 2 and 3 for the basic configurations of the DAC7642 and DAC7643.



FIGURE 1. DAC7642 and DAC7643 Architecture.





FIGURE 2. Basic Single-Supply Operation of the DAC7642 and DAC7643.



FIGURE 3. Basic Dual-Supply Operation of the DAC7642 and DAC7643.

ANALOG OUTPUTS

When $V_{SS} = -5V$ (dual-supply operation), the output amplifier can swing to within 2.25V of the supply rails over the -40° C to $+85^{\circ}$ C temperature range. When $V_{SS} = 0V$ (single-supply operation), and with R_{LOAD} also connected to ground, the output can swing to ground. Care must also be taken when measuring the zero-scale error when $V_{SS} = 0V$. Since the DAC output cannot swing below ground, the output voltage may not change for the first few digital input codes (0000_{H} , 0001_{H} , 0002_{H} , etc.) if the output amplifier has a negative offset. At the negative limit of -2mV, the first specified output starts at code 0040_{H} .

Due to the high accuracy of these DACs, system design problems such as grounding and contact resistance become very important. A 16-bit converter with a 2.5V full-scale range has a 1LSB value of 38μ V. With a load current of 1mA, a series wiring and connector resistance of only $40m\Omega$ (R_{W2}) will cause a voltage drop of 40μ V, as shown in Figure 4. To understand what this means in terms of a system layout, the resistivity of a typical 1 ounce copper-clad printed circuit board is 1/2 m Ω per square. For a 1mA load, a 10 milli-inch wide printed circuit conductor 600 milli-inches long will result in a voltage drop of 30μ V.

The DAC7642 and DAC7643 offer a force and sense output configuration for the high open-loop gain output amplifiers. This feature allows the loop around the output amplifier to be closed at the load (shown in Figure 4), thus ensuring an accurate output voltage.

REFERENCE INPUTS

The reference inputs, $V_{REF}L$ and $V_{REF}H$, can be any voltage between V_{SS} + 2.5V and V_{CC} – 2.5V provided that $V_{REF}H$ is at least 1.25V greater than $V_{REF}L$. The minimum output of each DAC is equal to $V_{REF}L$ plus a small offset voltage (essentially, the offset of the output op amp). The maximum output is equal to $V_{REF}H$ plus a similar offset voltage. Note



FIGURE 4. Analog Output Closed-Loop Configuration. R_W represents wiring resistances.

that V_{SS} (the negative power supply) must either be connected to ground or must be in the range of -4.75V to -5.25V. The voltage on V_{SS} sets several bias points within the converter. If V_{SS} is not in one of these two configurations, the bias values may be in error and proper operation of the device may be affected.

The current into the $V_{REF}H$ input and out of $V_{REF}L$ depends on the DAC output voltages and can vary from a few microamps to approximately 0.5mA. The reference input appears as a varying load to the reference. If the references applied can sink or source the required current, a reference buffer is not required. The DAC7642 and DAC7643 feature reference drive and sense connections such that the internal errors caused by the changing reference current and the circuit impedances can be minimized. Figures 5 through 13 show different reference configurations and the effect on the linearity and differential linearity.



FIGURE 5. Dual Supply Configuration-Buffered References, Used for Dual-Supply Characteristic Curves.









FIGURE 7. Integral Linearity and Differential Linearity Error Curves for Figure 6.



FIGURE 8. Integral Linearity and Differential Linearity Error Curves for Figure 9.



FIGURE 9. Single-Supply Buffered Reference with V_{REF}L = +1.25V and V_{REF}H = +2.5V.





FIGURE 10. Single-Supply Buffered V_{REF}H.



FIGURE 11. Linearity and Differential Linearity Error Curves for Figure 10.



FIGURE 12. Low-Cost Single-Supply Configuration.



FIGURE 13. Linearity and Differential Linearity Error Curves for Figure 12.

DIGITAL INTERFACE

See Table I for the basic control logic of the DAC7642 and DAC7643. Note that each internal register is edge triggered and not level triggered. When the LOADDACS signal is transitioned from LOW to HIGH, the digital word existing in the input register is latched into the DAC register. The first set of registers (the input registers) are triggered via the DACSEL, \overline{R}/W , and CS inputs. Only one of these registers can be transparent at any given time.

The double-buffered architecture is designed mainly so each DAC input register can be written to at any time without affecting the DAC outputs. All DAC voltages are updated simultaneously by the rising edge of LOADDACS. It also allows multiple devices to be updated simultaneously by sharing the LOADDACS control from the host with each device.





| DACSEL | R/W | CS | RST | LOADDACS | INPUT REGISTER | DAC REGISTER | MODE | DAC |
|--------|-----|----|------|----------|-------------------|-----------------|-------------|-----|
| L | L | L | L, H | Х | Write | Hold | Write Input | Α |
| н | L | L | L, H | Х | Write | Hold | Write Input | В |
| L | н | L | L, H | Х | Read | Hold | Read Input | A |
| н | н | L | L, H | Х | Read | Hold | Read Input | В |
| X | Х | н | L, H | 1 | Hold | Write | Update | All |
| X | Х | н | L, H | L, H | Hold | Hold | Hold | All |
| х | Х | х | ↑ | L, H | Reset | Reset | Reset | All |

TABLE I. DAC7642 and DAC7643 Logic Truth Table.

DIGITAL TIMING

Figure 14 and Table II provide detailed timing for the digital interface of the DAC7642 and DAC7643.

DIGITAL INPUT CODING

The DAC7642 and DAC7643 input data is in Straight Binary format. The output voltage is given by Equation 1:

$$V_{OUT} = V_{REF}L + \frac{(V_{REF}H - V_{REF}L) \cdot N}{65,536}$$
(1)

where N is the digital input code. This equation does not include the effects of offset (zero-scale) or gain (full-scale) errors.



FIGURE 14. Digital Input and Output Timing.

| SYMBOL | DESCRIPTION | MIN | ТҮР | МАХ | UNITS |
|------------------|---------------------------------------|-----|-----|-----|-------|
| t _{RCS} | CS LOW for Read | 150 | | | ns |
| t _{RDS} | R/W HIGH to CS LOW | 10 | | | ns |
| t _{RDH} | R/W HIGH after CS HIGH | 10 | | | ns |
| t _{DZ} | CS HIGH to Data Bus in High Impedance | 10 | | 100 | ns |
| t _{CSD} | CS LOW to Data Bus Valid | | 100 | 150 | ns |
| t _{wcs} | CS LOW for Write | 40 | | | ns |
| t _{WS} | R/W LOW to CS LOW | 0 | | | ns |
| t _{WH} | R/W LOW after CS HIGH | 10 | | | ns |
| t _{AS} | DACSEL Valid to CS LOW | 0 | | | ns |
| t _{AH} | DACSEL Valid after CS HIGH | 10 | | | ns |
| t _{LS} | CS LOW to LOADDACS HIGH | 30 | | | ns |
| t _{LH} | CS LOW after LOADDACS HIGH | 100 | | | ns |
| t _{LX} | LOADDACS HIGH | 100 | | | ns |
| t _{DS} | Data Valid to CS LOW | 0 | | | ns |
| t _{DH} | Data Valid after CS HIGH | 10 | | | ns |
| t _{LWD} | LOADDACS LOW | 100 | | | ns |
| t _{RSS} | RESET LOW | 10 | | | ns |
| t _{RSH} | RESET HIGH | 10 | | | ns |
| t _S | Settling Time | | | 10 | μs |

TABLE II. Timing Specifications ($T_A = -40^{\circ}C$ to +85°C).

DIGITALLY-PROGRAMMABLE CURRENT SOURCE

The DAC7642 and DAC7643 offer a unique set of features that allows a wide range of flexibility in designing applications circuits, such as programmable current sources. The DAC7642 and DAC7643 offer both a differential reference input, as well as an open-loop configuration around the output amplifier. The open-loop configuration around the output amplifier allows a transistor to be placed within the loop to implement a digitally-programmable, unidirectional current source. The availability of a differential reference also allows programmability for both the full-scale and zero-scale currents. The output current is calculated as:

$$I_{OUT} = \left(\left(\frac{V_{REF}H - V_{REF}L}{R_{SENSE}} \right) \bullet \left(\frac{N \text{ Value}}{65,536} \right) \right) + \left(V_{REF}L / R_{SENSE} \right)$$
(2)

Figure 15 shows a DAC7642 and DAC7643 in a 4-20mA current output configuration. The output current can be determined by Equation 3:

$$I_{OUT} = \left(\left(\frac{2.5V - 0.5V}{125\Omega} \right) \bullet \left(\frac{N \text{ Value}}{65,536} \right) \right) + \left(\frac{0.5V}{125\Omega} \right)$$
(3)

At full-scale, the output current is 16mA plus the 4mA for the zero current. At zero scale the output current is the offset current of 4mA ($0.5V/125\Omega$).



FIGURE 15. 4-20mA Digitally Controlled Current Source.

VF (S-PQFP-G32)

PLASTIC QUAD FLATPACK



NOTES: A. All linear dimensions are in millimeters.

B. This drawing is subject to change without notice.







PACKAGING INFORMATION

| Orderable Device | Status | Package Type | • | | Package Qty | Eco Plan | Lead/Ball Finish | MSL Peak Temp | Op Temp (°C) | Top-Side Markings | Samples |
|------------------|----------|--------------|---------|----|-------------|----------------------------|------------------|--------------------|--------------|-------------------|---------|
| | (1) | | Drawing | | | (2) | | (3) | | (4) | |
| DAC7642VFBT | ACTIVE | LQFP | VF | 32 | 250 | Green (RoHS & no Sb/Br) | CU NIPDAU | Level-1-260C-UNLIM | -40 to 85 | DAC7642 B | Samples |
| DAC7642VFBTG4 | ACTIVE | LQFP | VF | 32 | 250 | Green (RoHS & no Sb/Br) | CU NIPDAU | Level-1-260C-UNLIM | -40 to 85 | DAC7642 B | Samples |
| DAC7642VFT | ACTIVE | LQFP | VF | 32 | 250 | Green (RoHS & no Sb/Br) | CU NIPDAU | Level-1-260C-UNLIM | -40 to 85 | DAC7642 | Samples |
| DAC7642VFTG4 | ACTIVE | LQFP | VF | 32 | 250 | Green (RoHS & no Sb/Br) | CU NIPDAU | Level-1-260C-UNLIM | -40 to 85 | DAC7642 | Samples |
| DAC7643VFBR | OBSOLETE | LQFP | VF | 32 | | TBD | Call TI | Call TI | -40 to 85 | DAC7643 B | |
| DAC7643VFBRG4 | OBSOLETE | LQFP | VF | 32 | | TBD | Call TI | Call TI | -40 to 85 | DAC7643 B | |

⁽¹⁾ The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.

TBD: The Pb-Free/Green conversion plan has not been defined.

Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

Pb-Free (RoHS Exempt): This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

⁽³⁾ MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

⁽⁴⁾ Only one of markings shown within the brackets will appear on the physical device.

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PACKAGE MATERIALS INFORMATION

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TAPE AND REEL INFORMATION





QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



| *All dimensions are nominal | | | | | | | | | | | | |
|-----------------------------|-----------------|--------------------|----|-----|--------------------------|--------------------------|------------|------------|------------|------------|-----------|------------------|
| Device | Package Type | Package Drawing | | SPQ | Reel Diameter (mm) | Reel Width W1 (mm) | A0 (mm) | B0 (mm) | K0 (mm) | P1 (mm) | W (mm) | Pin1 Quadrant |
| DAC7642VFBT | LQFP | VF | 32 | 250 | 330.0 | 16.4 | 9.6 | 9.6 | 1.9 | 12.0 | 16.0 | Q2 |
| DAC7642VFT | LQFP | VF | 32 | 250 | 330.0 | 16.4 | 9.6 | 9.6 | 1.9 | 12.0 | 16.0 | Q2 |

TEXAS INSTRUMENTS

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PACKAGE MATERIALS INFORMATION

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*All dimensions are nominal

| Device | Package Type | Package Drawing | Pins | SPQ | Length (mm) | Width (mm) | Height (mm) |
|-------------|--------------|-----------------|------|-----|-------------|------------|-------------|
| DAC7642VFBT | LQFP | VF | 32 | 250 | 367.0 | 367.0 | 38.0 |
| DAC7642VFT | LQFP | VF | 32 | 250 | 367.0 | 367.0 | 38.0 |

MECHANICAL DATA

MTQF002B - JANUARY 1995 - REVISED MAY 2000



NOTES: A. All linear dimensions are in millimeters. B. This drawing is subject to change without notice.



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