#### INTRODUCTION

KA8510 is a monolithic circuit which can be used for high performance 46/49MHz MCA type Cordless Phone System. It is a transceiver IC for FM/FSK transmitting/receiving system, and is a complete one chip FM/FSK transceiver IC of 46/ 49MHz which includes transmitting and receiving functions for FM/FSK, a COMPANDOR to remove external noise, and PLL (Phase Locked Loop) of channel selection function which blocks surrounding frequency interference.

It has wide range of applications to FM/FSK transmitting/receiving of VHF bandwidth, including the cordless phone, and narrow band voice and data transmitting/receiving system. To make applications easy and simple, peripheral parts were minimized.



## **ORDERING INFORMATION**

Device	Package	Operating Temperature
KA8510Q	64-QFP-1420C	-25°C ~ + 75°C

## FEATURES

- Operating voltage range : 2 ~ 6V
- Typical supply current : 20mA at 3V
- Built-in dual conversion receiver, transmitter, compandor and PLL
- Compandor part
  - Easy gain control to use external component
  - Included ALC (automatic level control) circuit
- Mute logic, Data amp
- FM Transmitter part
  - Included variable capacitance
  - Adjustable power amp gain
- FM Receiver part
  - Complete dual conversion circuit with RF amplifier
  - Excellent input sensitivity (0.7µVrms at 12dB SINAD)
- PLL part
  - 10/15 channels selectable (both transmitter/receiver)
  - Include oscillation circuit with external crystal (10.24MHz)
  - Unlock detector (Phase difference more than 6.25µs)



### **BLOCK DIAGRAM**





# **1CHIP ANALOG CORDLESS TRANSCEIVER**

#### **PIN CONFIGURATION**



## **PIN DESCRIPTION**

Pin No	Symbol	Description
1	МО	It Generates carrier signal and modulates frequency. The audio signal enters the MI terminal and causes the internal reactance
		to vary, which changes the frequency generated from the coil tank, resul- ting in frequency modulation.
		Input terminal of RF carrier signal.
2	RF OSC	By varying the capacitor value, which is connected between MO terminals,
2	11 000	the amount of frequency modulation (Hz/mV) can be adjusted.
		Larger capacitor value increases the amount of modulation.
3	GND <sub>TX</sub>	Ground.
5	GNDIX	Ground of a transmitter part.
		Variable capacitor included in the chip. Since the RF oscillation frequency
4	VARICAP	changes according as the detected pulse error at PLL influences the variable
		capacitor, it becomes useful when a channel is being changed.
5	MD	AC bypass terminal of FM modulator. A capacitor is connected between
5	UIVI	this terminal and GND.
		Input terminal of RF modulator. It inputs an audio signal through the comp-
6	MI	ressor of compandor to modulate FM.
Ŭ	1011	Variation of RF generated frequency per voltage (mV) at this terminal beco-
		mes the sensitivity of frequency modulation.



Pin No	Symbol	Description
7	со	It is a COMPRESSOR output terminal of COMPANDOR, and is connected to the modulation input terminal of transmitter.
8	PDT	It is a phase detector output terminal of transmitter at PLL. If $f_{TX} > f_{REF}$ or $f_{TX}$ is leading -> output is negative pulse. If $f_{TX} < f_{REF}$ or $f_{TX}$ is lagging -> output is positive pulse. If $f_{TX} = f_{REF}$ and the same phase -> output is high impedance.
9	PDR	It is a phase detector output terminal of transmitter at PLL. If $f_{RX} > f_{REF}$ or $f_{RX}$ is leading -> output is negative pulse. If $f_{RX} < f_{REF}$ or $f_{RX}$ is lagging -> output is positive pulse. If $f_{RX} = f_{REF}$ and the same phase -> output is high impedance.
10	VCO <sub>RX</sub>	This is a variable capacitor terminal, and is included in the chip. It is used as an input terminal where 1'st local oscillation frequency is oscillates. The internal variable capacitor has the value of 10~25pF according to the applied voltage (0.7~V <sub>CC</sub> )
11	LDT	It is an output terminal of lock detector of transmitter stage at PLL. Output is low if PLL is in lock state, and is high if PLL is in unlock state.
12	GND <sub>PLL</sub>	Ground It is used as ground of logic section at PLL.
13	TEST1	This terminal tests operation of PLL. It is operated together with test2. When it is not in use, a resistor is connected between V <sub>CC</sub> and this terminal.
14	TEST2	It is an input terminal for PLL test. When test is set to high, outputs of Rx- programmable divider and Tx-programmable divider can be tested at LDT terminal.
15	MODE	It is PLL's mode selection terminal. This terminal is set to high if it is used for the base set of a cordless phone, and is set to low if it is used for the hand set of a cordless phone.
16 17 18 19	D0 D1 D2 D3	It is a decode input terminal for PLL's channel selection. Four bit binary code of D0(LSB) ~ D3(MSB) are given by MICOM. These inputs have internal pull down devices
20	E MUTE	It is EXPANDER MUTE terminal of COMPANDOR, and is the final MUTE block of an EXPANDER located next to the receiver terminal. It blo- cks the data signal of MICOM being transmitted to an user, and is connec- ted to the Rx MUTE terminal of MICOM. Expanding is executed if it is low.



Pin No	Symbol	Description
		It is a COMPRESSOR MUTE terminal of a COMPANDOR. MUTE block
		is inserted before the data input terminal of COMPRESSOR to avoid dupli-
		cation of data transmission from MICOM (between the base and hand set)
21	C MUTE	with the voice signal.
		It is connected to the TX MUTE terminal of MICOM.
		Compressing is executed if this terminal is high, and COMPRESSOR
		MUTE is executed if it is low.
		It is a data input terminal of COMPRESSOR, and is used to transmit data of
22	DI	handset MICOM to baseset, or data of baseset MICOM to handset.
		This terminal is connected to the data output terminal of MICOM.
		This terminal is used for bypassing an AC element at the feedback loop
23	AGIC	which comes from the SUM AMP block of COMPRESSOR. A capacitor
		should be connected between this terminal and GND.
		This terminal is used for converting waveform from the full wave rectifier
24	CRC	to DC element at the rectifier block of COMPRESSOR.
		(RC = 22msec)
25	CPI -	It is a PRE AMP inverting input terminal of COMPRESSOR, and is used
25	CPI-	for adjusting the negative feedback loop gain. (in application, gain is 5)
26	CPI +	It is a PRE AMP non-iverting input terminal of COMPRESSOR, and is used
20	CPI+	as an input terminal for voice signal.
07		Ground
27		Ground of COMPANDOR
		Supply voltage
28	V <sub>CC(COMP)</sub>	Power supply terminal of COMPANDOR
	10	It is an output terminal of EXPANDER, which a regenerated voice sig-
29	EO	nal comes out.
		This terminal is used for inverting waveform from the full wave rectifier to
30	ERC	DC element at the rectifier block of EXPANDER.
		(RC = 22 msec)
		It is a SUM AMP input terminal of EXPANDER. After the demodulated signal
31	EPI	passed through the 2'nd order low pass filter, the voice signal enters this
		terminal.



Pin No	Symbol	Description
		It is a input reference voltage terminal of ALC (Automatic Level Control).
		ALC circuit may be turned off according to the ALC reference voltage, or
		magnitude of output voltage may be limited if it is higher than a certain
32	ALC	level.
		It is used for adjusting THD of output voltage of COMPRESSOR to less
		than 3% or to limit the frequency deviation of TX in case the input is higher
		than a certain level.
33	V <sub>REF(COMP)</sub>	It is a voltage reference ( $V_{REF} = 1V$ ) used for supplying a constant voltage
	V REF(COMP)	to the COMPRESSOR and EXPANDER of COMPANDOR.
		It is an output terminal of a carrier detector, and outputs the contents of
		a buffer of a Meter Driver which is turned on/off according to a certain sig-
34	CDO	nal level detected by the Meter driver. Since this terminal is an open collec-
		tor output type, it requires a pull-up resistor, and is connected to RSSI ter-
		minal of MICOM.
		It is an output terminal of a Meter Driver.
	MDO	(RSSI : Received Signal Strength Indicator)
35		Amplitude of RF input signal for useful frequency is detected by a Meter
00	MEG	Driver circuit.
		The Meter Driver circuit has perfect linear characteristic of 60dB range for
		input signal level. (0.1μA/dB)
		It is an output terminal of Data Slicing Comparator.
36	DSCO	It separates FSK (Frequency-Shift Keying) Serial data and executes data
		shaping and limiting.
37	DSCI	It is an input terminal of Data Slicing Comparator, and is a non-inverting
		type with negative input terminal biased to $1/2 V_{CC}$ .
38	RAO	It is a recovered audio output terminal. Voice signal which is detected by
		Quadrature Detector is amplified and then is output via this terminal.
		It is a Quadrature Coil input terminal. The 455KHz Oscillator circuit is com-
39	QCI	prised of $L_{P}$ = 680 $\mu H,$ $C_{P}$ = 180 pF, LC Tank circuit. Voice signal is detected
		by mixture of 455KHz (by phase difference) with is converted from se-
		cond Mixer.
40	GND <sub>RX</sub>	Ground
	GIUDRX	Ground for receiver



Pin No	Symbol	Description
41	LD1	It is a Limiter input and Decoupling terminal, and is used for removing am-
		plitude modulation element caused by fading or noise of FM signal.
42	LD2	Limiting IF amplifier amplifies and limits second intermediate frequency
		455KHz signal.
43	LI	The input impedance of Limiting IF amplifier is designed to 1.5Kohm.
		While FM wave is transmitted with constant magnitude, its magnitude is
		slightly modulated due to the reflection from obstacles, fading phenome-
		non, noise wave, etc., and mixed with AM wave element before it enters
		the receiver's antenna.
		The Limiter makes uniform amplitude by removing these AM wave ele-
		ment.
44		Supply Voltage
44	V <sub>CC(RX)</sub>	This terminal is used for supplying power to a receiver.
		It is an output terminal of second mixer. Second intermediate frequency
45	2 MO	455KHz, generated by mixing first intermediate frequency (10.7MHz) and
		second local oscillator, is output.
40		It is an input terminal of second local oscillator. It generates second local
46	2 LOI	oscillator frequency to convert output from first Mixer (10.7MHz) to second
47	2 LOI	intermediate frequency. It is an oscillator with crystal of 10.24MHz,10.245
		MHz.
40	0.14	It is an input terminal of second Mixer. Output from first Mixer is entered
48	2 MI	to second Mixer input terminal via 10.7MHz ceramic filter. Second Mixer
49	2 MI	converts frequency to second intermediate frequency (455KHz : AM IF).
		It is an output terminal of first Mixer. Signal from the first Mixer and freque-
		ncy of first local oscillator are mixed and become first intermediate freque-
50	1 MO	ncy, which is then output through this terminal.
		The output terminal is made of emitter follower with an output impedance
		of 330ohm to match 330ohm input/output impedance of 10.7MHz ceramic
		filter.
F.4	410	It is an input terminal of first local oscillator.
51	1 LOI	The local oscillator is a voltage controlled oscillator. Local oscillation fre-
52	1 LOI	quency and received frequency are mixed at the first Mixer and then are
		converted to first intermediate frequency of 10.7MHz.



Pin No	Symbol	Description
53	1 MI	It is an input terminal of first Mixer. This Mixer is made of double balanced
54	1 MI	multiplier. Received signal (46/49 MHz) amplified at RF AMP is input to
		this terminal.
		It is an collector terminal of RF amplifier transistor.
55	RFC	Amount of amplification can be selected using the resistor load at RF amp-
		lifier collector.
		It is an emitter terminal of RF amplifier transistor.
		The gain of RF AMP can be adjusted using the resistance of emitter dege-
56	RFE	neration.
		$I_E$ = 1.5mA when emitter of RF AMP is connected to the ground, and vol-
		tage gain on 1 Kohm of collector load is Av = 20dB
		It is a base terminal of RF amplifier transistor.
		Base of TR is biased internally to about 0.8V from the common-emitter am-
57	RFB	plifier design.
		This RF AMP amplifies subtle signal received by the antenna to proper
		size for conversion into first intermediate frequency (10.7MHz).
50	GND	Ground
58	GND	Ground for analog at PLL.
50		Supply Voltage
59	V <sub>CC(PLL RX)</sub>	This terminal supplies power from PLL to Rx ECL area.
<u></u>		Supply Voltage
60	V <sub>CC(PLL TX)</sub>	This terminal supplies power from PLL to Tx ECL area.
		This terminal outputs transmission frequency (46/49MHz) from RF oscilla-
61	RF OUT	tor. Tx spurious characterestics can be adjusted by connection a capacitor
		between pin 63 and this terminal.
		This is an output terminal of a POWER AMP, and is open collector type.
62	AMPO TX	It provides 2 point oscillation and amplification to narrow the bandwidth
		of transmitter output.
		This terminal controls the gain of a POWER AMP, and is open base type.
63	AMPI TX	The AMP gain can be selected by connecting a variable resistor between
		V <sub>cc</sub> and this terminal.
		Supply Voltage
64	V <sub>cc</sub> (TX)	It is a $V_{CC}$ of transmitter. It permits voltage supply to use seperately if the tran-
	,	
	V <sub>cc</sub> (TX)	V <sub>cc</sub> and this terminal. Supply Voltage



## ABSOLUTE MAXIMUM RATINGS (Ta = 25 C)

Characteristics	Symbol	Value	Unit
Maximum Supply Voltage	V <sub>cc</sub>	7	V
Power Dissipation	PD	600	mW
Junction Temperature	TJ	+ 125	°C
Operation Temperature	T <sub>OPR</sub>	-25 ~ + 75	°C
Storage Temperature	T <sub>STG</sub>	-65 ~ + 150	°C

## ELECTRICAL CHARACTERISTICS

Characteristic	Symbol	Test Conditions	Min	Тур	Мах	Unit
Operating Voltage	V <sub>cc</sub>	-	2.0	-	6.0	V
Total Operating Current	Icc	-	-	20	-	mA
RECEIVER (V <sub>CC</sub> = 5V, f	<sub>c</sub> = 49.7MHz	z, f <sub>DEV</sub> = ± 3KHz, f <sub>MOD</sub> = 1KHz, Ta =	25°C, Unle	ess otherwi	se specifie	(b
Operating Current 1	I <sub>CC</sub> 1	RFin = 1mVrms V <sub>CC</sub> = 3V	-	3.5	6.0	mA
Operating Current 2	I <sub>CC</sub> 2	$V_{CC} = 5V$ , RFin = 1mVrms	-	4.5	7.0	mA
Input for -3dB Sensitivity	V <sub>LIM</sub>	$\label{eq:RFin} \begin{array}{l} RFin = 1mVrms \to 0dB \\ \\ RFin = 1mVrms \to -3dB \end{array}$	-	0.7	2.0	μVrms
Input for 20dB Sensitivity	V <sub>I(SEN)</sub>	Modulation Input None Modulation Input	-	0.7	2.0	μVrms
S/N Ratio	S/N	Modulation Input None Modulation Input	48	55	-	dB
Recoverd Audio Output	V <sub>O(RA)</sub>	RFin = 1mVrms f <sub>mod</sub> = 1KHz f <sub>dev</sub> = ± 3KHz	210	350	490	mVrms
Noise Output Level	V <sub>NO</sub>	RFin = None Input	-	250	400	mVrms
Recoverd Audio Output Voltage Drop	V <sub>O(RAD)</sub>	$V_{CC} = 5V \rightarrow 2V$ RFin = 1mVrms	-8	-1.5	-	dB
Direct Output Voltage	V <sub>O(DET)</sub>	RFin = 1mVrms	1.6	2.3	3.0	V <sub>dc</sub>
Carrier Detector Throshold	V <sub>TH(DET)</sub>	RFin = None Input $R_{L(34)} = 10K\Omega$	0.49	0.60	0.73	V <sub>dc</sub>



## ELECTRICAL CHARACTERISTICS (continued)

Characteristics	Symbol	Test Conditions	Min	Тур	Max	Unit	
RECEIVER ( $T_a = 25^{\circ}C$ , Unless otherwise specified)							
Comparator Threshold Voltage Difference	$\Delta \; V_{\text{TH (COMP)}}$	$V_{1 (COMP)} = 1 KHz, 150 mV_{P-P}$ $R_{L (37)} = 180 K\Omega$	70	110	150	mW	
Comparator Output Voltage 1	V <sub>OH (COMP)</sub>	"	3.8	4.25	-	V <sub>dc</sub>	
Comparator Output Voltage 2	V <sub>OL (COMP)</sub>	"	-	0.25	0.5	V <sub>dc</sub>	
Second Mixer Coversion Voltage Gain	$\Delta G_{V(2M)}$	$V_{1 (54)} = 1mVrms$ $R_{L (50)} = 330\Omega$	14	18	22	dB	
First Mixer Coversion Voltage Gain	$\Delta  G_{V(1M)}$	$V_{1 (48)} = 1mVrms$ $R_{L (45)} = 330 \text{ K}\Omega$	17	21	26	dB	
AM Rejection Ratio	AMR	RFin = 1mVrms AM MOD = 30%	25	35	-	dB	
Detector Output Distortion	THD DET	RFin = 1mVrms	-	2.0	3.0	%	
Detector Output Resistance	R <sub>O (DET)</sub>	RFin = 1mVrms	-	1.4	2.0	KΩ	
Detector Output DC Voltage Change Ratio	$\Delta V_{O(\text{DET})}$	RFin = 1mVrms	-	0.12	0.2	V/KHz	
Meter Drive Slope	MDS	-	70	100	135	nA/dB	
First Mixer Input Resistance	R <sub>I (1M)</sub>	f <sub>C</sub> = 50MHz	500	690	-	Ω	
First Mixer Input Capacitance	C <sub>I (1M)</sub>	f <sub>C</sub> = 50MHz	-	7.2	10	pF	
Limitter Input Sensitivity	V <sub>I (LIM)</sub>	f <sub>C</sub> = 455MHz, 20dB S/N	-	100	250	μVrms	
Second Mixer Input Sensitivity	S <sub>V (2M)</sub>	f <sub>C</sub> = 10.7MHz, 20dB S/N	-	10	25	μVrms	
First Mixer 3rd Order Intercept	3RD	-	-	-2.2	-	μVrms	
RF Amp DC Current Gain	G <sub>I (AMP)</sub>	V <sub>CC</sub> = 5Vdc	1.0	1.5	2.5	μVrms	



Unit

mΑ

KHz

V

pF

mΑ

 $\mathsf{V}_{\mathsf{dc}}$ 

dB

dB

%

dB

 $V_{P-P}$ 

dB

dB

dB

%

dB

#### Characteristics Symbol **Test Conditions** Min Тур Max TRANSMITTER ( $V_{CC}$ = 3V, $f_C$ = 49.86MHz, $T_a$ = 25°C, Unless otherwise specified) **Operating Current** Icc $V_{CC} = 3V$ 1.0 1.5 2.0 Modulation Senstitivity $S_{\text{MOD}}$ $V_{in}=0.9\sim1.1V$ 100 150 200 Hz/mV Maximum Deviation $V_{in} = 0 \sim 2.0 V$ ± 40 ± 70 ± 120 $\Delta~{\rm f}_{\rm (MAX)}$ 0.5 0.7 First Amp bias Voltage VBIAS V<sub>in</sub> = Open 0.9 $f_{O} = 49.860 MHz$ First Amp RF Out Voltage $\mathsf{V}_{\mathsf{RF}}$ 100 300 mVrms -Variable Capacitance Zero Bias 30 VCAP --COMPRESSOR ( $V_{CC}$ =3V, f=1MHz, Ta = 25°C, Unless otherwise specified) **Operating Current** Icc No signal 3.6 6.0 Reference Voltage $\mathsf{V}_{\mathsf{REF}}$ No signal 0.9 1.0 1.1 Standard Output Voltage $V_{inc} = 13mVrms \rightarrow 0dB$ 240 300 340 mVrms V<sub>O(COMP)</sub> $V_{inc} = -20 dB$ -0.5 0 +0.5 $\Delta \; G_{\text{V1(COMP)}}$ Compressor Gain Difference $V_{inc} = -40 dB$ -1.0 0 +1.0 $\Delta G_{V2(COMP)}$ Compressor Output Distortion THD<sub>COMP</sub> $V_{inc} = 0 dB$ 0.5 1.0 Mute Attenuation Ratio $\mathsf{ATT}_{\mathsf{MUTE}}$ $V_{inc} = 0dB, CMUTE = GND$ 60 80 -Compressor Limitting Voltage V<sub>inc</sub> = Variable 1.15 V<sub>LIM(COMP)</sub> 1.35 1.50 ALC $V_{\text{ALC}}$ 280 325 360 mVrms EXPANDER (V<sub>CC</sub> = 3V, f = 1MHz, Ta = 25°C, Unless otherwise specified) Standard Output Voltage V<sub>O(EXP)</sub> $V_{ine} = 180 mVrms \rightarrow 0 dB$ 110 130 160 mVrms $V_{in} = -10 dB$ -0.5 0 +0.5 $\Delta \; G_{\text{V1(EXP)}}$ Expander Gain Difference $V_{in} = -20 dB$ -1.0 0 +0.1 $\Delta \; \textbf{G}_{\text{V2(EXP)}}$ $V_{in} = -30 dB$ -1.5 +1.5 0 $\Delta G_{V3(EXP)}$ Expander Output Distortion THD<sub>EXP</sub> $V_{in} = 0 dB$ -0.5 1.5 Mute Attenuation Ratio $\mathsf{ATT}_{\mathsf{MUTE}}$ $V_{in} = 0 dB$ 60 85 -V<sub>ine</sub> = Variable Expander Maximum -500 600 mVrms VOEXP(MAX)

THD = 10%

#### **ELECTRICAL CHARACTERISTICS (continued)**



Output Voltage

Characteristic	Symbol	Test Conditions	Min	Тур	Max	Unit		
	PLL (V <sub>CC</sub> = 3V, Ta = 25°C, Unless otherwise specified)							
Input Current	Iн	$V_{in} = V_{CC} - 0.5V$	30	50	80	μA		
input Gunent	Ι <sub>ΙL</sub>	$V_{in} = 0V$	-	0	1	μΑ		
Input Voltage	V <sub>IH</sub>	I <sub>in</sub> = 20μA	1.2	1.5	1.7			
Input Voltage	VIL	I <sub>in</sub> = 3μA	0	0.3	0.5	V		
	V <sub>OH 1</sub>	PDT, PDR : I <sub>0</sub> = 1mA(sourcing)	2.3	2.7	3.0	V		
Output Voltage	V <sub>OL 1</sub>	PDT, PDR : I <sub>0</sub> = 1mA(Sinking)	0	0.3	0.5	v		
oupur volago	V <sub>OH 2</sub>	LDT : I <sub>O</sub> = 1mA(sourcing)	2.3	2.7	3.0	V		
	V <sub>OL 2</sub>	LDT : I <sub>0</sub> = 1mA(Sinking)	0	0.3	0.5	v		
Tx Counter	f <sub>TX</sub>	$f_{in(TX)} = 42MHz$ $300mV_{p-p}$	-1	Fout	+1	Hz		
Rx Counter	f <sub>RX</sub>	$f_{in(RX)} = 42MHz$ 300mV <sub>p-p</sub>	-1	Fout	+1	Hz		

## ELECTRICAL CHARACTERISTICS (continued)



## APPLICATION CIRCUIT



