

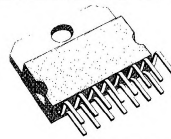
## DUAL SWITCH-MODE SOLENOID DRIVER

### PRELIMINARY DATA

- HIGH CURRENT CAPABILITY (up to 2.5A per channel)
- HIGH VOLTAGE OPERATION (up to 46V for power stage)
- HIGH EFFICIENCY SWITCHMODE OPERATION
- REGULATED OUTPUT CURRENT (adjustable)
- FEW EXTERNAL COMPONENTS
- SEPARATE LOGIC SUPPLY
- THERMAL PROTECTION

is completely controlled by means of a switching technique allowing very efficient operation. Furthermore, it includes an enable input and dual supplies (for interfacing with peripherals running at a higher voltage than the logic).

The L295 is particularly suitable for applications such as hammer driving in matrix printers, step motor driving and electromagnet controllers.



Multiwatt 15

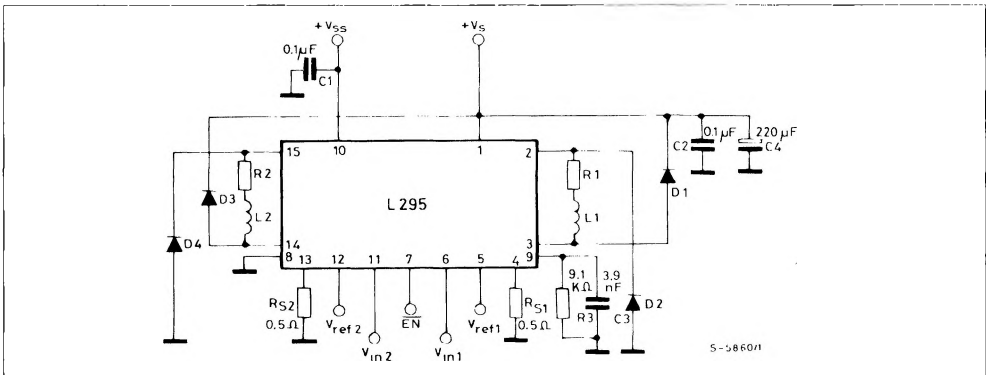
ORDER CODE : L295

The L295 is a monolithic integrated circuit in a 15-lead Multiwatt<sup>®</sup> package ; it incorporates all the functions for direct interfacing between digital circuitry and inductive loads. The L295 is designed to accept standard microprocessor logic levels at the inputs and can drive 2 solenoids. The output current

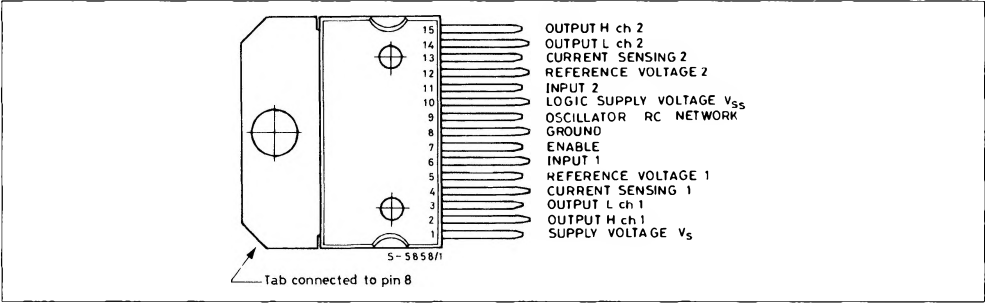
### ABSOLUTE MAXIMUM RATINGS

$V_s$	Supply voltage	50	V
$V_{ss}$	Logic supply voltage	12	V
$V_{EN}, V_i$	Enable and input voltage	7	V
$V_{ref}$	Reference voltage	7	V
$I_o$	Peak output current (each channel)		
	– non repetitive ( $t = 100 \mu\text{sec}$ )	3	A
	– repetitive (80 % on – 20 % off ; $T_{on} = 10 \text{ ms}$ )	2.5	A
	– DC operation	2	A
$P_{tot}$	Total power dissipation (at $T_{case} = 75^\circ\text{C}$ )	25	W
$T_{stg}, T_j$	Storage and junction temperature	– 40 to 150	$^\circ\text{C}$

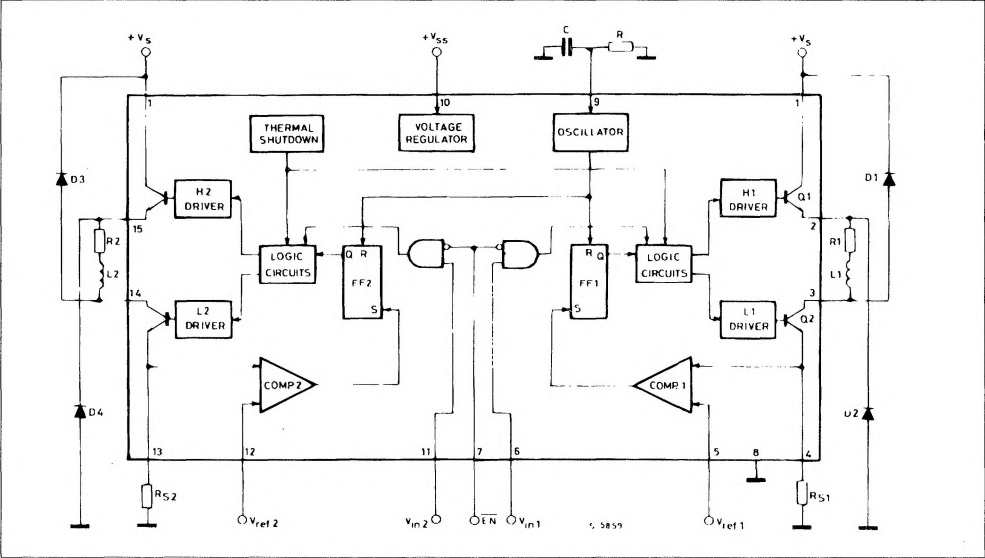
### APPLICATION CIRCUIT



CONNECTION DIAGRAM (top view)



BLOCK DIAGRAM



THERMAL DATA

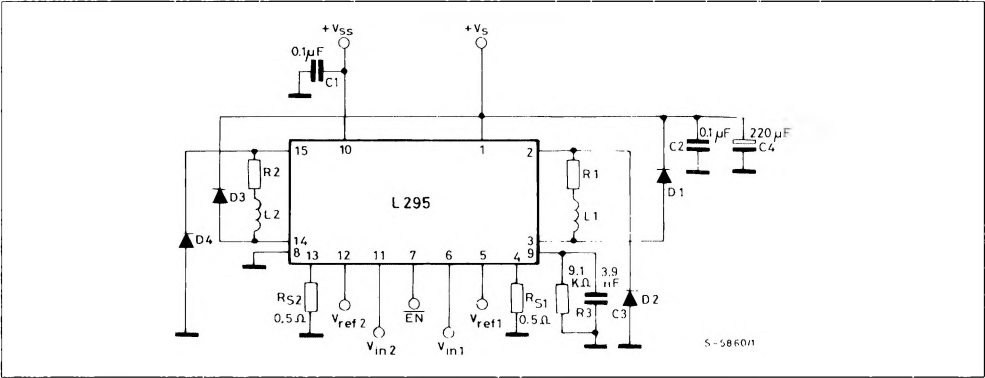
$R_{th\ j-case}$	Thermal resistance junction-case	max	3	°C/W
$R_{th\ j-amb}$	Thermal resistance junction-ambient	max	35	°C/W

**ELECTRICAL CHARACTERISTICS** (Refer to the application circuit,  $V_{SS} = 5V$ ,  $V_S = 36V$ ;  $T_j = 25^\circ C$ ; L = low; H = high; unless otherwise specified)

Parameter		Test conditions	Min.	Typ.	Max.	Unit
$V_S$	Supply Voltage		12		46	V
$V_{SS}$	Logic Supply Voltage		4.75		10	V
$I_D$	Quiescent drain current (from VSS)	$V_S = 46V$ ; $V_{I1} = V_{I2} = V_{EN} = L$			4	mA
$I_{SS}$	Quiescent drain current (from VS)	$V_{SS} = 10V$			46	mA
$V_{I1}, V_{I2}$	Input Voltage	Low	-0.3		0.8	V
		High	2.2		7	
$V_{EN}$	Enable Input Voltage	Low	-0.3		0.8	V
		High	2.2		7	
$I_{I1}, I_{I2}$	Input Current	$V_{I1} = V_{I2} = L$			-100	$\mu A$
		$V_{I1} = V_{I2} = H$			10	
$I_{EN}$	Enable Input Current	$V_{EN} = L$			-100	$\mu A$
		$V_{EN} = H$			10	
$V_{ref1}, V_{ref2}$	Input Reference Voltage		0.2		2	V
$I_{ref1}, I_{ref2m}$	Input Reference Voltage				-5	$\mu A$
$F_{osc}$	Oscillation Frequency	$C = 3.9 nF$ ; $R = 9.1 K\Omega$		25		KHz
$I_p, V_{ref}$	Transconductance (each ch.)	$V_{ref} = 1V$ $R_S = 0.5\Omega$	1.9	2	2.1	A/V
$V_{drop}$	Total output voltage drop (each channel) (*)	$I_o = 2A$		2.8	3.6	V
$V_{sens1}, V_{sens2}$	External sensing resistors voltage drop				2	V

(\*)  $V_{drop} = V_{CEsat Q1} + V_{CEsat Q2}$ .

APPLICATION CIRCUIT



D2, D4 = 2A High speed diodes  
D1, D3 = 1A High speed diodes  
R1 = R2 = 2Ω  
L1 = L2 = 5 mH  
t<sub>rr</sub> ≤ 200 ns

FUNCTIONAL DESCRIPTION

The L295 incorporates two independent driver channels with separate inputs and outputs, each capable of driving an inductive load (see block diagram).

The device is controlled by three microprocessor compatible digital inputs and two analog inputs. These inputs are :

- EN chip enable (digital input, active low) , enables both channels when in the low state.
- V<sub>in1</sub>, V<sub>in2</sub> channel inputs (digital inputs, active high), enable each channel independently. A channel is activated when both  $\overline{\text{EN}}$  and the appropriate channel input are active.
- V<sub>ref1</sub>, V<sub>ref2</sub> reference voltages (analog inputs), used to program the peak load currents. Peak load current is proportional to V<sub>ref</sub>.

Since the two channels are identical, only channel one will be described.

The following description applies also the channel two, replacing FF2 for FF1, V<sub>ref2</sub> for V<sub>ref1</sub> etc.

When the channel is activated by a low level on the EN input and a high level on the channel input, V<sub>in2</sub>,

the output transistors Q1 and Q2 switch on and current flows in the load according to the exponential law :

$$I = \frac{V}{R1} \left( 1 - e^{-\frac{R1 t}{L1}} \right)$$

where : R1 and L1 are the resistance and inductance of the load and V is the voltage available on the load (V<sub>s</sub> - V<sub>drop</sub> - V<sub>sense</sub>).

The current increases until the voltage on the external sensing resistor, R<sub>S1</sub>, reaches the reference voltage, V<sub>ref1</sub>. This peak current, I<sub>p1</sub>, is given by :

$$I_{p1} = \frac{V_{ref1}}{R_{S1}}$$

At this point the comparator output, Comp1, sets the RS flip-flop, FF1, that turns off the output transistor, Q1. The load current flowing through D2, Q2, R<sub>S1</sub>, decreases according to the law :

$$I = \left( \frac{V_A}{R1} + I_{p1} \right) e^{-\frac{R1 t}{L1}} - \frac{V_A}{R1}$$

where V<sub>A</sub> = V<sub>CEsat</sub> Q2 + V<sub>sense</sub> 1 + V<sub>D2</sub>

If the oscillator pin (9) is connected to ground the load current falls to zero as shown in fig. 1.

At the time  $t_2$  the channel 1 is disabled, by taking the inputs  $V_{in1}$  low and/or  $\overline{EN}$  high, and the output transistor Q2 is turned off. The load current flows through D2 and D1 according to the law :

$$I = \left( \frac{V_B}{R_1} + I_{T2} \right) e^{-\frac{R_1 t}{L_1}} - \frac{V_B}{R_1}$$

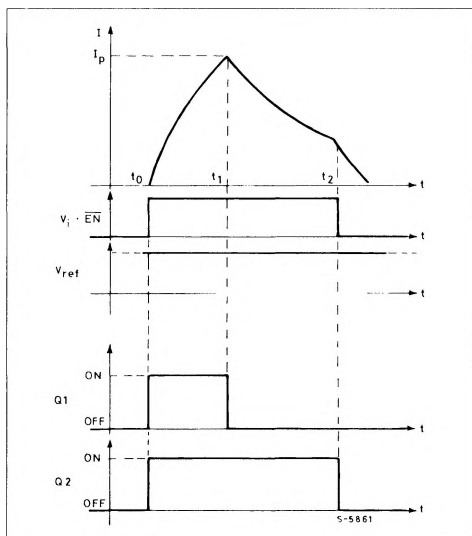
where  $V_B = V_S + V_{D1} + V_{D2}$

$I_{T2}$  = current value at the time  $t_2$ .

Fig. 2 in shows the current waveform obtained with an RC network connected between pin 9 and ground. From to  $t_1$  the current increases as in fig. 1. A difference exists at the time  $t_2$  because the current starts to increase again. At this time a pulse is produced by the oscillator circuit that resets the flip flop, FF1, and switches on the output transistor, Q1. The current increases until the drop on the sensing resistor  $R_{S1}$  is equal to  $V_{ref1}$  ( $t_3$ ) and the cycle repeats.

## SIGNAL WAVEFORMS

**Figure 1 :** Load current waveform with pin 9 connected to GND.



The switching frequency depends on the values of R and C, as shown in fig. 4 and must be chosen in the range 10 to 30 KHz.

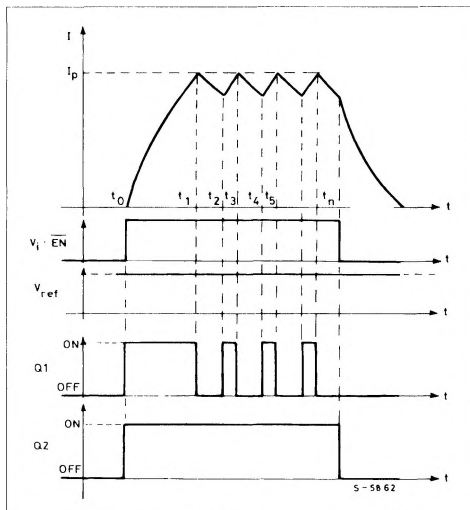
It is possible with external hardware to change the reference voltage  $V_{ref}$  in order to obtain a high peak current  $I_p$  and a lower holding current  $I_h$  (see fig.3).

The L295 is provided with a thermal protection that switches off all the output transistors when the junction temperature exceeds 150 °C. The presence of a hysteresis circuit makes the IC work again after a fall of the junction temperature of about 20 °C.

The analog input pins ( $V_{ref1}$ ,  $V_{ref2}$ ) can be left open or connected to  $V_{SS}$ ; in this case the circuit works with an internal reference voltage of about 2.5V and the peak current in the load is fixed only by the value of  $R_S$  :

$$I_p = \frac{2.5}{R_S}$$

**Figure 2 :** Load current waveform with external R-C network connected between pin 9 and ground.



SIGNAL WAVEFORMS (continued)

Figure 3 : With  $V_{ref}$  changed by hardware.

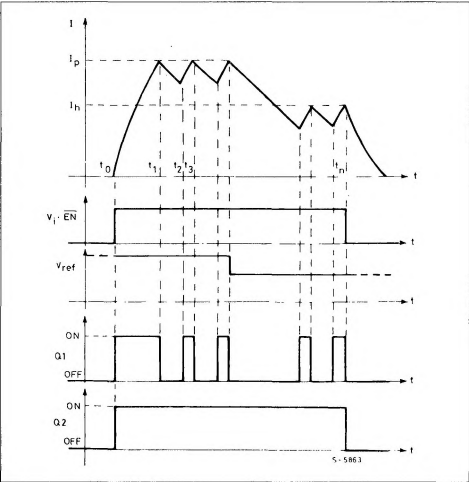


Figure 4 : Switching frequency vs. values of  $R$  and  $C$ .

