

# L4952

## VOLTAGE REGULATOR FOR CAR RADIO APPLICATIONS

#### PRELIMINARY DATA

- OPERATION SUPPLY VOLTAGE 9.5 TO 28V
- PEAK SUPPLY OVERVOLTAGE PULSE UP TO 52V (VS. GND)
- FIXED OUTPUT VOLTAGE OF 8.6 ± 0.2V AND OUTPUT CURRENT UP TO 150mA
- VERY LOW DROP OUTPUT STAGE WITH LOW OUTPUT VOLTAGE DIAGNOSTIC
- OUTPUT SHORT CIRCUIT AND THERMAL OVERLOAD PROTECTION
- VERY LOW STANDBY CURRENT (DEVICE DISABLED)
- TWO INTERNAL OUTPUT VOLTAGE SWITCHES WITH LOW DROPOUT FOR AM/FM SWITCHING
- DRIVER CIRCUIT FOR EXTERNAL HIGH SIDE SWITCH WITH DROP MONITORING
- ENABLE INPUT TO SWITCH ON THE DE-VICE
- ADDITIONAL 10V VERY LOW DROP REGU-LATOR TO SUPPLY THE VARICAP



#### DESCRIPTION

The L4952 is a monolithic voltage regulator in an advanced BCD60II process with high efficient pchannel transistor. This device is optimized for Car Radio applications to obtain optimum performance and supply system integration with high flexibility and minimum peripheral components.

## **BLOCK DIAGRAM**



This is preliminary information on a new product now in development or undergoing evaluation. Details are subject to change without notice.

#### **ABSOLUTE MAXIMUM RATINGS**

Symbol	Parameter	Value	Unit
Vs	DC supply voltage	-2.2 +28	V
Vst	Transient operating supply voltage for t = 400ms	-2.2 +52	V
V <sub>01,2,3,4</sub>	Output voltages	-0.3 +12	V
V <sub>EN</sub>	Input voltage of EN	-0.3 +18	V
V <sub>IND</sub>	Input voltage for EN2, EN3, HSON	-18 +18	V
V <sub>s</sub> -V <sub>IND</sub>	Max Voltage Difference between $V_{\text{s}}$ and logic inputs EN2, EN3, HSON	52	V
V <sub>DIAG</sub>	Diagnostic output	-0.3 +20	V
I <sub>o1</sub>	Output current out1, out4	short circuit protected	mA
I <sub>02,3</sub>	Output current out2,3 (Note 1)	internally limited	mA
I <sub>oDIAG</sub>	Output current Diagnostic	10	mA
I <sub>HSSD</sub>	Driver current for external pnp High side switch internally limited		mA
TJ	Operating junction temperature	-40 +150	°C
Po	Power dissipation with on board heat sink 2cm <sup>2</sup>	1	W
V <sub>ESD</sub>	ESD voltage capability (MIL 883 C)	±2000	V

Note: Current limiter of OUT1 will also protect OUT2 and OUT3 as long as OUT1 is not reverse supplied. Output capacitors up to 100µF between OUT and GND will not harm this protection.

#### **PIN CONNECTION**



## THERMAL DATA

Symbol	Parameter	Value	Unit
R <sub>th j-amb</sub>	Thermal Resistance Junction to Ambient with on board heatsink $2\text{cm}^2$	60	°C/W
T <sub>JSD</sub>	Thermal shutdown junction temperature	>150	°C



Symbol	bol Parameter Test Condition		Min.	Тур.	Max.	Unit
I <sub>s</sub>	V <sub>s</sub> Quiescent current	no Load		1.6	5	mA
I <sub>ssb</sub>	V <sub>s</sub> standby current	EN = 0V		0	20	μA
VINLEN	Low Voltage EN		-0.3		0.5	V
VINHEN	High Voltage EN	V <sub>S</sub> = 18V	3.5		18	V
V <sub>IN L EN2,3</sub>	EN2, EN3, Low Voltage		-18		1.05	V
VIN H EN2,3	EN2, EN3, High Voltage	V <sub>S</sub> = 18V	1.45		18	V
VIN L HSON	HSON Low Voltage		-18		1.0	V
V <sub>IN H HSON</sub>	HSON High Voltage	V <sub>S</sub> = 18V	2.0		18	V
I <sub>INH</sub>	Current EN2, EN3, HSON	$V_{In} = 5V$		30	70	μA
I <sub>INL</sub>	Current EN2, EN3, HSON	$V_{In} = 0V$	-10		10	μA
IIN ON/OFF	Current High EN	V <sub>In</sub> = 5V	-10	100	200	μA
	gulator (OUT 1)	•				
V <sub>o1</sub>	Output voltage	no load	8.4	8.6	8.8	V
V <sub>DP1</sub>	Dropout voltage	$V_s = 8.4V, V_{DP1} = V_s - V_{out1}$				
		I <sub>01</sub> = 0.15A, I <sub>02,3</sub> = 0mA		0.16	0.6	V
		$I_{01} = 0.1A, I_{02,3} = 0mA$		0.11	0.4	V
		I <sub>01</sub> = 0.05A, I <sub>02,3</sub> = 0mA		0.06	0.2	V
SVRR	Supply Voltage ripple rejection	$f = 100Hz, C = 10\mu F ESR = 5\Omega$	60	70		dB
V <sub>oLo</sub>	Load regulation	10mA < I <sub>o</sub> < 150 mA		30	60	mV
I <sub>oLim1</sub>	Current limits	V <sub>OUT</sub> = 8V	150	300	800	mA
Voltage Re	gulator (OUT 4)					
V <sub>o4</sub>	Output voltage	no load	9.5	10.0	10.5	V
V <sub>DP4</sub>	Dropout voltage	$I = 3mA V_s = 8.6V$		0.14	0.3	V
5		I = 1.5mA V <sub>s</sub> = 8.6V		0.075	0.15	V
SVRR	Supply Voltage ripple rejection	f = 100Hz C = 10μF ESR = 5Ω	30	60		dB
V <sub>o4Lo</sub>	Load regulation	0.5mA < I < 3 mA		100		mV
I <sub>o4Lim</sub>	Current limits	V <sub>OUT</sub> = 8V	8	60		mA
Diagnostic	Output					
V <sub>LDIAG</sub>	Output Diagnostic Low voltage	$R_{DIAG}$ to 5 V = 10K $\Omega$			0.4	V
$\Delta V_{out}$	Output voltage drop before diagnostic activation	$V_{S} > 6V I_{O1} = 100mA$ $I_{O2} = I_{O3} = 0mA$		30	200	mV
Voltage Sw	itches Vout2/3	• • • •		•		
V <sub>DP2</sub>	Dropout V <sub>out2</sub>	I <sub>o2</sub> = 50mA, En <sub>2</sub> = H		0.25	0.5	V
0,2		$I_{02} = 25 \text{mA}, \text{ En}_2 = \text{H}$		0.125	0.25	V
V <sub>DP3</sub>	Dropout V <sub>out3</sub>	$I_{03} = 75 \text{mA}, \text{ En}_3 = L$		0.25	0.5	V
		$I_{03} = 40 \text{mA}, \text{En}_3 = L$		0.14	0.28	V
High Side S	Switch Driver					L
V <sub>DPHS</sub>	Low Drop Voltage	I <sub>out</sub> = 50mA, HSON = H		0.26	0.8	V
V <sub>DPHS</sub>	Drop detection threshold	$V_{\text{Dth}} = V_{\text{S}} - V_{14}$ , HSON = H	0.7	1	1.3	V

## **ELECTRICAL CHARACTERISTICS** ( $V_s = 14V$ ; $T_J = 25^{\circ}C$ unless otherwise specified)

#### FUNCTIONAL DESCRIPTION

The L4952 is a monolithic voltage regulator with an output voltage of typically 8.6V and a maximum output current of 150mA. It 's a device for audio applications in carradios.

The device contain a precision Bandgap reference, a output amplifier with overcurrent protection, two voltage switches, a driver for an external

pnp switch, a discharge circuitry for theft detection and a overtemperature protection.

For automatic PCB assembly the package is SO12+4+4. To use the maximal current of 150mA a small copper area of 2cm<sup>2</sup> as heat sink on board (Rth=60K/W) is necessary.

The device has a very low quiescent current in standby mode. If the digital Input EN is Low the device is switched off (stand by mode) and if EN



## Figure 1: Application Circuit Diagram



## Figure 2.



## Figure 3.



57

is High the device is switched on. The diagnostic circuitry detect the low drop voltage.

In this case the DIAG output is going low and can mute the power output stage to avoid noise on the loudspeaker. The two internal switches can switch the stabilized output voltage with P-MOS Transistors to one of the outputs 2 and 3 with low drop. This is useful to switch the AM and FM circuitry on or off.

To control it there are two digital inputs  $En_2$  and  $En_3$ one for each switch.  $EN_2$  is High active and  $EN_3$  is Low active. It's possible to drive the AM/FM switch with one digital line ( $EN_2$  and  $EN_3$  together).

The driver for the external High side switch can switch on and off the external pnp transistor. The drop detection circuitry avoid the damage of the external power pnp transistor.

To supply the varicaps and the PLL-opamp of the car radio a second very low drop 10 V regulator is available. This regulator in dropout has a typical resistance of  $50\Omega$ .

#### **Typical Characteristics (Note 4)**

## Figure 3: Stand by consumption versus temperature



Supply voltage rejection (C =  $10\mu$ F, ESR =  $4.7\Omega$ , Load at OUT4 =  $10k\Omega$ )

Figure 5: Supply rejection versus Frequency.



## Function of the external High side switch driver

Fig 2 shows the principle circuitry of the external High side switch. Fig. 3 shows the switch on/off phase of the external High side switch. At the time t0 the microcontroller switches on (curve 1 =output signal of the microcontroller).

The signal on the HSON pin of the L4952 is shown on curve 2. At t1 the external Power pnp is switched on. At t2 ( $V_{HSON} = V_{SON2}$ ) the internal comparator compares the drop of the external pnp. In case of normal operation the drop is smaller than  $V_{Dth}$  and no failure will be detected (curve 3).

In case of failure (that mean a higher drop than  $V_{Dth}$ ) the external power pnp will be switched off (curve 4). If an error is detected it will be stored in the internal error flip-flop. The external pnp can only be switched on again after having turned HSON off (V<sub>HSON</sub> < V<sub>SON1</sub>) again.





OUT2 propagation delay (Load =  $100\Omega$ )

**Figure 6:** OUT2 propagation delay (Load =  $100\Omega$ )







**Figure 7:** OUT3 propagation delay (Load =  $100\Omega$ )









#### Note 4

Typical charcteristics shown by the Figures 3 to 12 are typical parameters. Depending on produc-





Figure 10: OUT1 voltage versus temperature



Figure 12: OUT4 voltage versus temperature



tion spread certain deviations may occure. For limits see pages 2 to 4.

SGS-THOMSON MICROELECTROMICS

57



MAX. 0.104 0.012 0.096 0.019 0.013

0.512

0.419

0.299

0.050

0.030

DIM.		mm			inch
	MIN.	TYP.	MAX.	MIN.	TYP.
А			2.65		
a1	0.1		0.3	0.004	
a2			2.45		
b	0.35		0.49	0.014	
b1	0.23		0.32	0.009	
С		0.5			0.020

1.27

11.43

### SO20 PACKAGE MECHANICAL DATA

12.6

10

7.4

0.5

c1

D

Е

e e3

F

L

Μ



45 (typ.)

0.496

0.394

0.291

0.020

0.050

0.450

13.0

10.65

7.6

1.27

0.75

SGS-THOMSON MICROELECTRONICS

Information furnished is believed to be accurate and reliable. However, SGS-THOMSON Microelectronics assumes no responsibility for the consequences of use of such information nor for any infringement of patents or other rights of third parties which may result from its use. No license is granted by implication or otherwise under any patent or patent rights of SGS-THOMSON Microelectronics. Specification mentioned in this publication are subject to change without notice. This publication supersedes and replaces all information previously supplied. SGS-THOMSON Microelectronics or systems without express written approval of SGS-THOMSON Microelectronics.

© 1996 SGS-THOMSON Microelectronics - Printed in Italy - All Rights Reserved

SGS-THOMSON Microelectronics GROUP OF COMPANIES

Australia - Brazil - Canada - China - France - Germany - Hong Kong - Italy - Japan - Korea - Malaysia - Malta - Morocco - The Netherlands -Singapore - Spain - Sweden - Switzerland - Taiwan - Thailand - United Kingdom - U.S.A.

