CMOS LSI

No. 4799A

SANYO

LC7538JM

Electronic Volume Control System for Car Audio

Overview

The LC7538JM is a fully equipped electronic volume IC which permits significant reductions in externally connected components while providing ample volume, balance, loudness, fader, bass and treble control functions.

Features

- Volume : 81 positions ranging from 0 dB to -79 dB (in 1 dB increments) plus -∞. Separate left and right control provides excellent balance function.

- Loudness : Loudness operation provided by externally attached CR to activate tap at the -20 dB position of the volume ladder resistor.

- Fader : Fader function traversing 16 positions with rear or front attenuated output only (these 16 positions consist of 2 dB step intervals ranging from 0 dB to -20 dB, 5 dB step intervals ranging from -20 dB to -45 dB, plus the end settings of -60 dB and -∞).

- Bass and Treble: Using externally attached C (capacitor), the LC7538JM provides bass-treble mutual 15-position control and formats a NF-form tone control circuit (LUX form).

- On-chip op amplifier for caching applications reduces external components.

- Reduced switching noise with silicon gate CMOS processor.

- All controls performed using serial data input (C^2B).

Package Dimensions

unit : mm

3204

SANYO : MFP36S
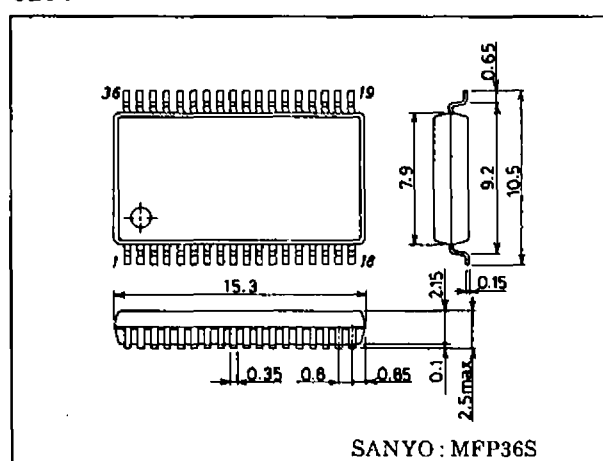
Specifications

Absolute Maximum Ratings at Ta = 25°C, V_{SS} = 0 V

Parameter	Symbol	Conditions	Ratings	Unit
Maximum supply voltage	V_{DD} max	V_{DD}	11	V
Maximum input voltage	V_{IN} max1	CL, DI, CE	$V_{SS} - 0.3$ to $V_{DD} + 0.3$	V
	V_{IN} max2	LTIN, RTIN, L5dBIN, R5dBIN, L1dBIN, R1dBIN, LFIN, RFIN	$V_{SS} - 0.3$ to $V_{DD} + 0.3$	V
Allowable power dissipation	Pd max	Ta ≤ 85°C	210	mW
Operating temperature	Topr		−40 to +85	°C
Storage temperature	Tstg		−50 to +125	°C

Allowable Operation Conditions at Ta = 25°C, V_{SS} = 0 V

| Parameter	Symbol	Conditions	Ratings			Unit
			min	typ	max	
Supply voltage	VDD	*1	7.0		10.0	V
Input high level voltage	V_{IH}	CL, DI, CE	4.0		V_{DD}	V
Input low level voltage	VIL	CL, DI, CE	V_{SS}		1.0	V
Input amplitude voltage	V_{IN}	LTIN, RTIN, L5dBIN, R5dBIN, L1dBIN, R1dBIN, LFIN, RFIN	V_{SS}		V_{DD}	Vp-p
Input pulse width	$t_{\varphi W}$	CL	1			µs
Setup time	t_{SETUP}	CL, DI, CE	1			µs
Hold time	t_{HOLD}	CL, DI, CE	1			µs
Operating Frequency	fopg	CL			500	kHz

Note: 1. A capacitor rated at 2000 pF or less should be installed between all power supply pins and V_{SS}.

Electrical Characteristics at Ta = 25°C, V_{DD} = 9 V, V_{SS} = 0 V

Parameter	Symbol	Conditions	Ratings			Unit
			min	typ	max	
Total harmonic distortion	THD (1)	V_{IN} = 1 Vrms, f = 1 kHz, total overall flat		0.04		%
	THD (2)	V_{IN} = 1 Vrms, f = 20 kHz, total overall flat		0.06		%
Crosstalk	CT	V_{IN} = 1 Vrms, f = 1 kHz, total overall flat, Rg = 1 kΩ	60	87		dB
Maximum Output Reduction	Vo min	V_{IN} = 1 Vrms, f = 1 kHz, main volume −∞, fader volume −∞, C = 1000 µF between Vref and V_{SS} for L/R		82		dB
All Resistance Value	R_{VOL} (1)	5 dB step	15	25	35	kΩ
	R_{VOL} (2)	1 dB step	12	20	28	kΩ
	R_{FADER}		12	20	28	kΩ
	R_{BASS}		48	80	112	kΩ
	R_{TREBLE}		30	50	70	kΩ
Input high level current	I_{IH}	VI = 8 V (CL, CE, DI pins)		·	10	µA
Input low level current	I_{IL}	VI = 0 V (CL, CE, DI pins)	−10			µA
Output noise voltage	V_N	All overall flat (IHF-A), Rg = 1 kΩ		7.5	15	µV
Current dissipation	I_{DD}	V_{DD}-V_{SS} = 10 V		15	21	mA
Analog switch on resistance	R_{ON}	CT1	1.8	3.0	4.2	kΩ
		Between CT2 and Vref	0.6	1.0	1.4	kΩ
		Fader S1 to S4	1.8	3.0	4.2	kΩ
		−∞	0.6	1.0	1.4	kΩ
		All other cases	6.0	10.0	14.0	kΩ

Equivalent Circuit Block Diagram

A01020
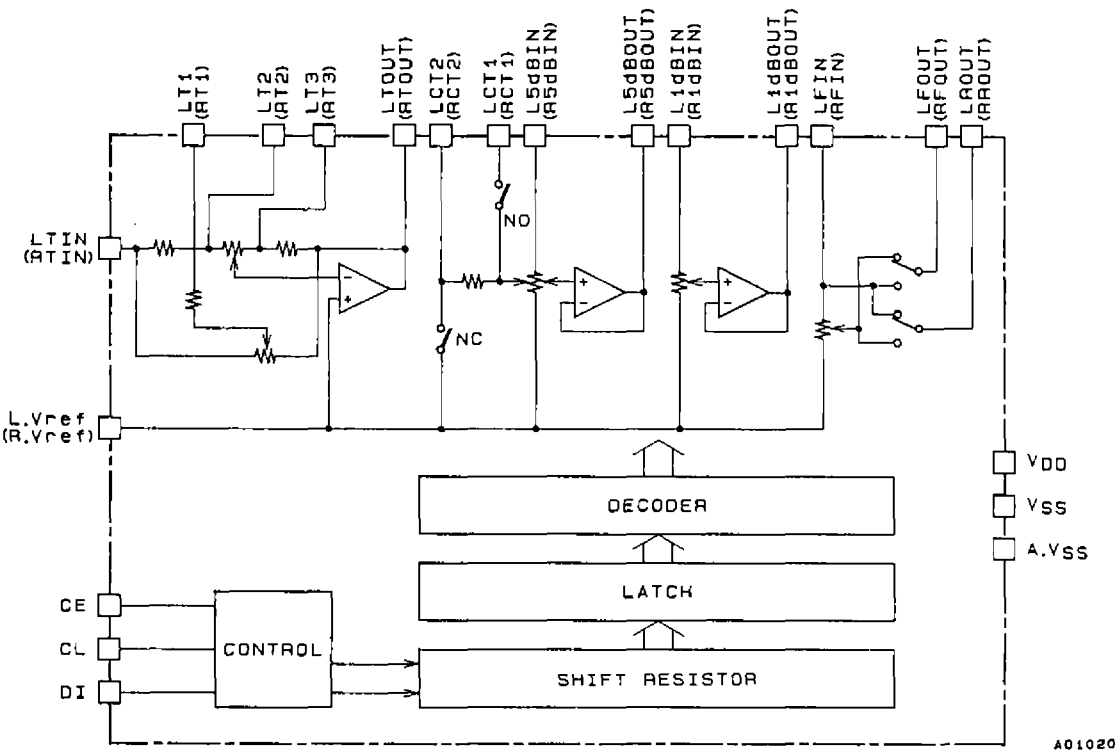
Test Circuit

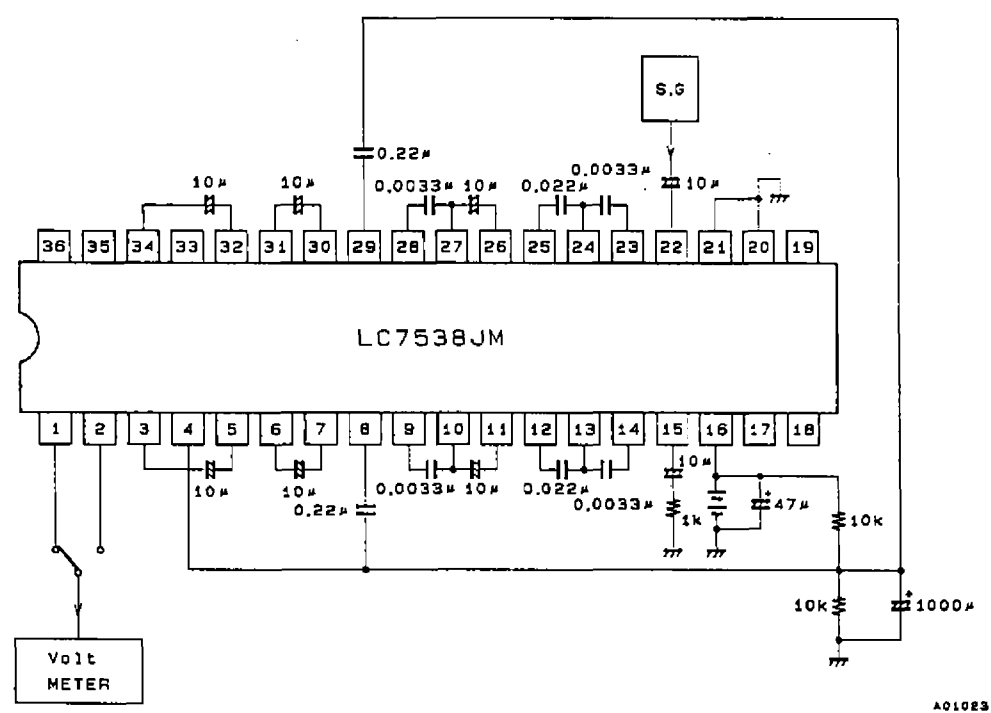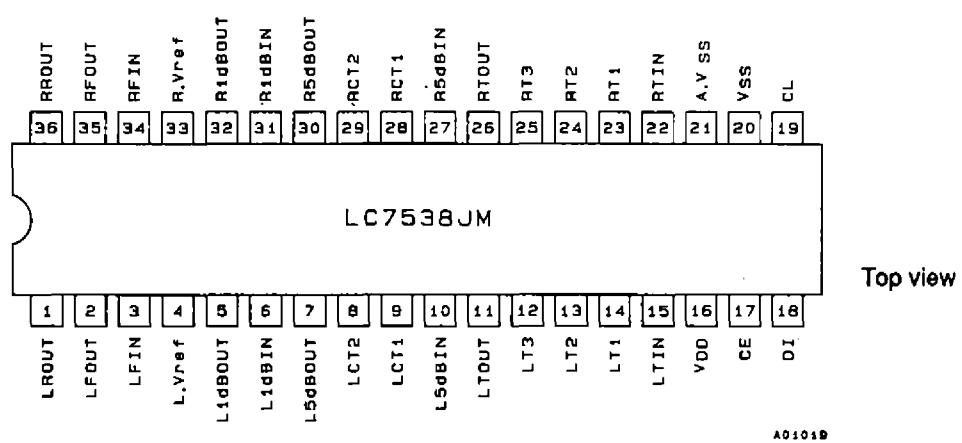### a) Total Harmonic Distortion

A01021

Unit (resistance: Ω, capacitance: F)
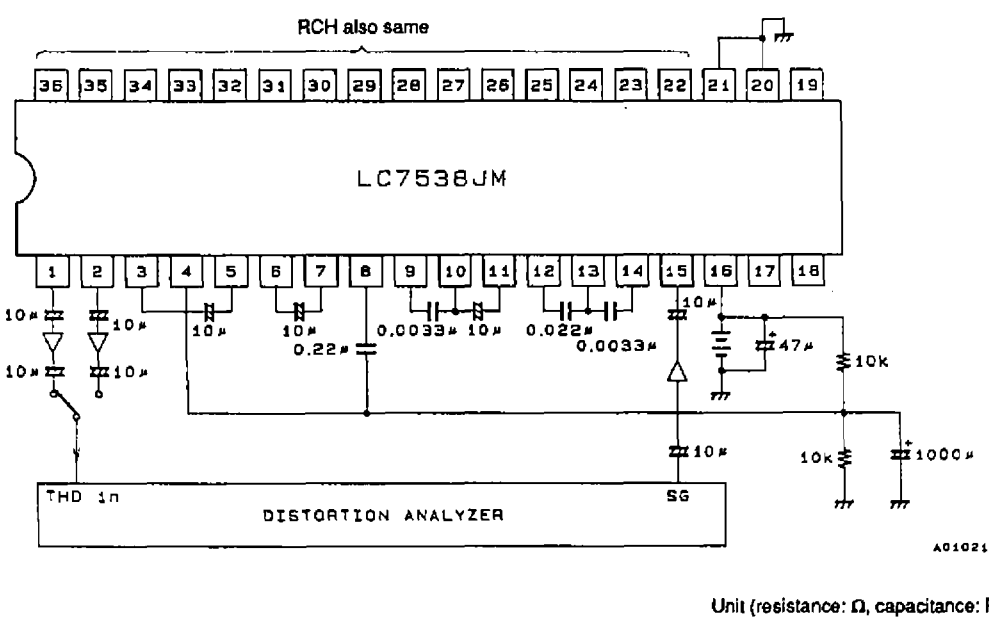
b) Output Noise Voltage

Unit (resistance: Ω, capacitance: F)

RCH also same

A01022

c) Crosstalk

A01023

Pin Assignment

Top view

A01019
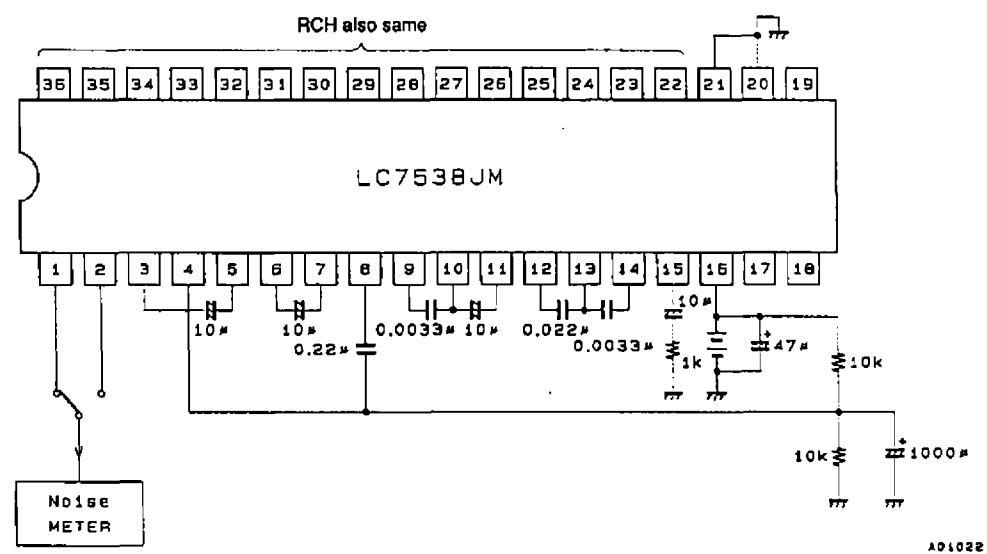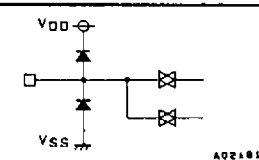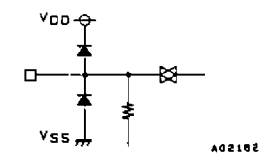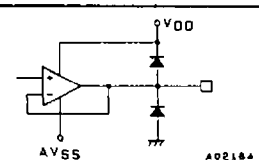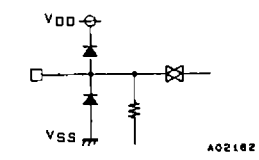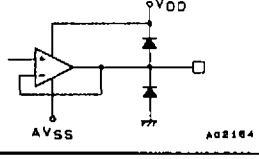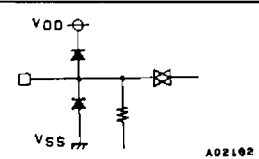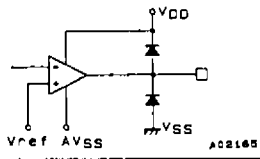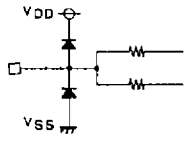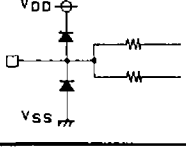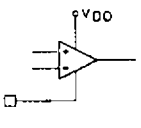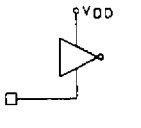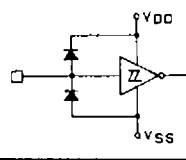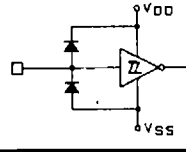
Pin Descriptions

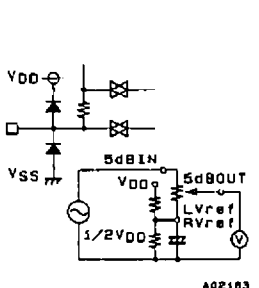| Pin name	Pin No.	Description	Remarks
LROUT	1	• These pins function as output pins for the fader. Output reduction for rear and front is performed separately for each. Attenuation capacity is unified for both left and right. Step positioning is designed using an open circuit so that reception is performed using high impedance.	
LFOUT	2		
RROUT	36		
RFOUT	35		
LFIN	3	• When utilizing the fader function, these pins function as input pins. • Low impedance driven.	
RFIN	34		
LVREF	4	• These pins are common pins for fader volume, tone and main volume. The pattern impedance connected here should be lowered as much as possible. • LVref and RVref are not connected to V_{SS}. • Connections for LVref and RVref to V_{SS} should be established externally to match all specifications. Notably, attention should be paid to capacity since capacitors are subject to residual resistance during volume output reduction when installed between LVref (RVref) and V_{SS} as is the case with single power sources. • Normally, high voltage applied from V_{DD}.	
RVREF	33		
L1dBOUT	5	• These pins are output pins for the 1 dB step attenuator located in the section main volume.	
R1dBOUT	32		
L1dBIN	6	• These pins are input pins for the 1 dB step attenuator located in the section main volume. • Low impedance driven.	
R1dBIN	31		
L5dBOUT	7	• These pins are output pins for the 5 dB step attenuator located in the section main volume.	
R5dBOUT	30		
LCT1	9	• These pins are for loudness control. Connect a high-band compensation capacitor between CT1 to 5dB IN and a low-band compensation capacitor between CT2 to Vref.	
LCT2	8		
RCT1	28		
RCT2	29		
L5dBIN	10	• These pins are input pins for the 5 dB step attenuator located in the section main volume. • Low impedance driven.	
R5dBIN	27		
LTOUT	11	• These pins are output pins for tone control.	
RTOUT	26		

Continued from preceding page.

Pin name	Pin No.	Description	Remarks
LT3	12	• These pins are for connecting bass and treble compensation for the tone circuit. Connect a high-band compensation capacitor between T1 and T2. Connect a low-band compensation capacitor between T2 and T3.	
LT2	13		
LT1	14		
RT3	25		
RT2	24		
RT1	23		
LTIN	15	• These pins are tone control input pins. • Low impedance driven.	
RTIN	22		
V$_{DD}$	16	• Supply voltage pin.	
A. V$_{SS}$	21	• Ground pin for on-chip op amp.	
V$_{SS}$	20	• Ground pin for internal logic.	
CE	17	• This is the chip enable pin. According to the timing of the switch from high to low, data is written to an internal latch and all analog switches operate. Data transfer with high-level switches to enable.	
DI	18	• These are input pins for the clock and serial data for control.	
CL	19		

Equivalent Circuit for Main Volume Section

Unit (resistance: Ω)

L5dBIN 0
L5dBOUT
L1dBIN 0
L1dBOUT

7.147 4.019 2.260 1.271 | -5 -10 -15 -20

2.175 1.936 1.728 1.540 | -1 -2 -3 -4

12.619 | -∞

LCT1

1.952

4.376 2.542 1.383 0.779 0.438 0.246 0.138 0.078 0.044 0.024 0.014 0.018

LCT2

-25 -30 -35 -40 -45 -50 -55 -60 -65 -70 -75 -∞

Same as right channel

LVref

unit:kΩ

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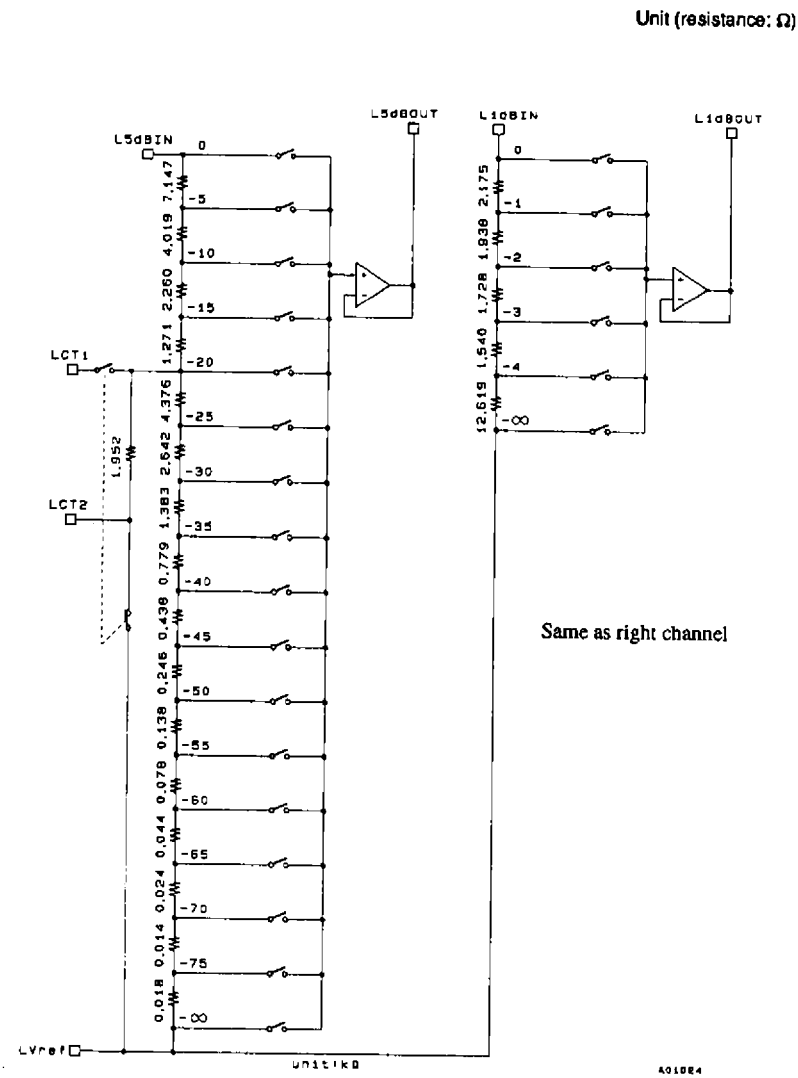
Equivalent Circuit for Fader Volume Section

Unit (resistance: Ω)

FADER = "1" turns S2, S3 on
FADER = "0" turns S1, S4 on

Same as right channel
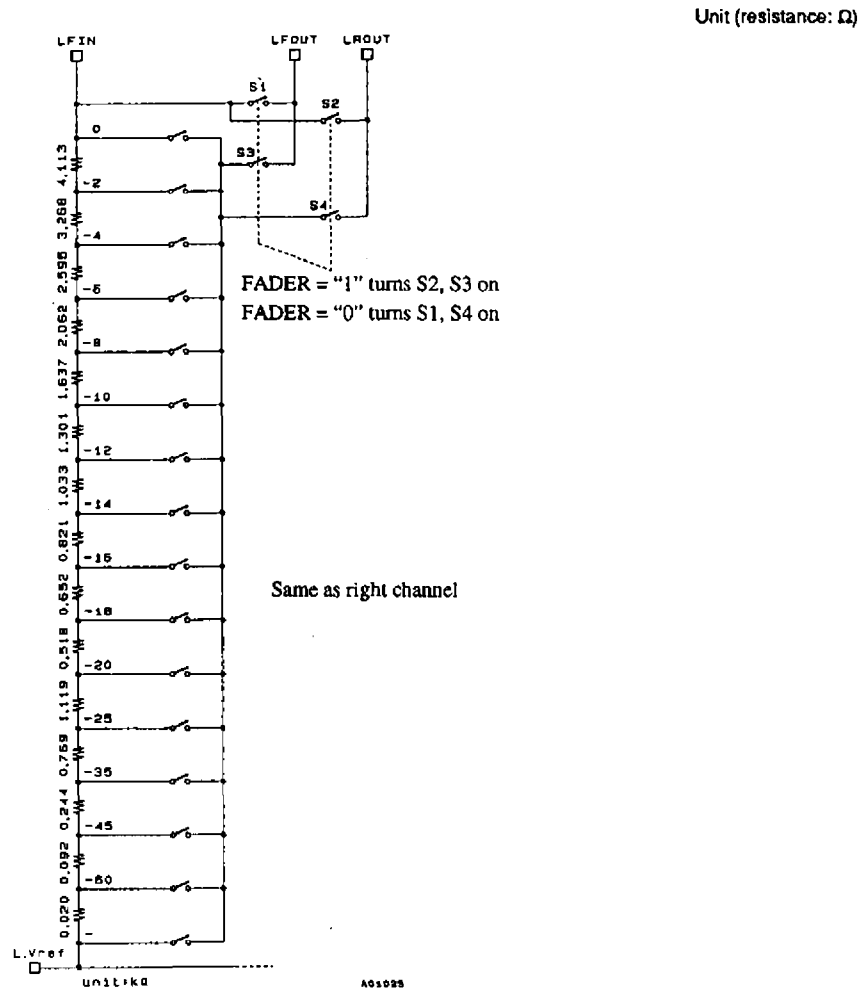
When data of -∞ is transferred to main volume
control 1 dB step, S1 and S2 open and S3, S4 are
turned on simultaneously.

Equivalent Circuit for Tone Section
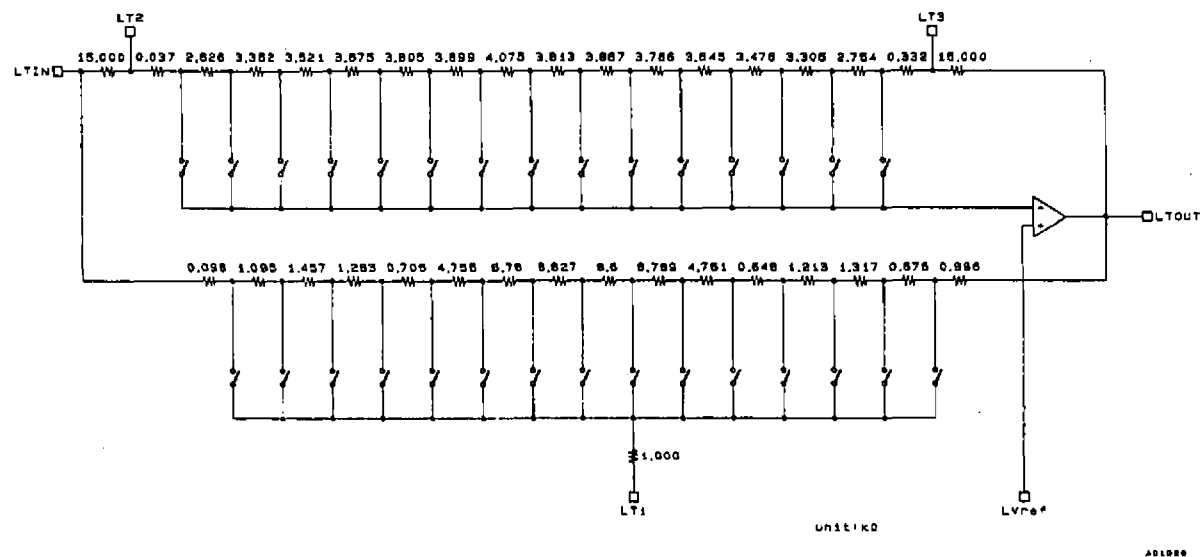
Control System Timing and Data Format

Controlling of LC7538JM involves the input of regulating serial data to CE, CL and DI pins. Data format consists of 36 bits composed of an 8-bit address and 28-bit data.

CE

DI | B0 | B1 | B2 | B3 | A0 | A1 | A2 | A3 | D0 | D1 | D2 | D3 | D4 | D5 | D6 | D7 | D23 | D24 | D25 | 10 | 11 |

CL

CE

CL

DI

1μs min
1μs min
1μs min
1μs min

1μs min

1μs≤T DEST

B	B	B	B	A	A	A	A	D	D	D	D	D	D	D	D	D	D	D	D	D	D	D	D	D	D	D	D	D	D	D	D	D	D	1	1
0	1	2	3	0	1	2	3	0	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15	16	17	18	19	20	21	22	23	24	25	0	1

Don't care

Address Code

1	0	0	0	0	0	0	1

Channel Selection

0	0	0	0	
1	0	0	0	: Rch
0	1	0	0	: Lch
1	1	0	0	: L/R simultaneously

1 dB Step Control

0	0	0	0	: −∞
1	0	0	0	: −4dB
0	1	0	0	: −3dB
1	1	0	0	: −2dB
0	0	1	0	: −1dB
1	0	1	0	: 0dB

Rear/Front Control

0	Rear
1	Front

Loudness Control

0	OFF
1	ON

BASS Control

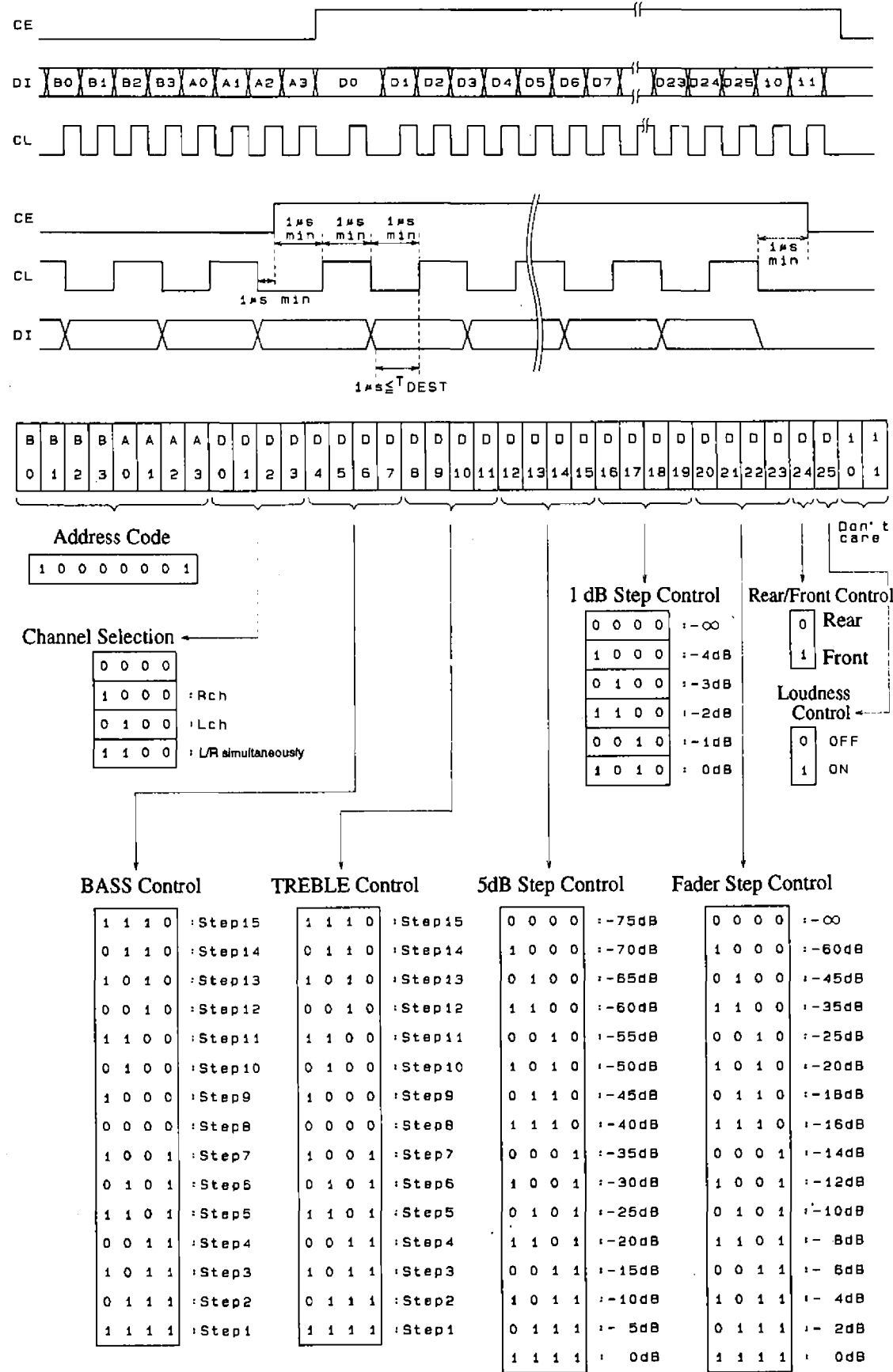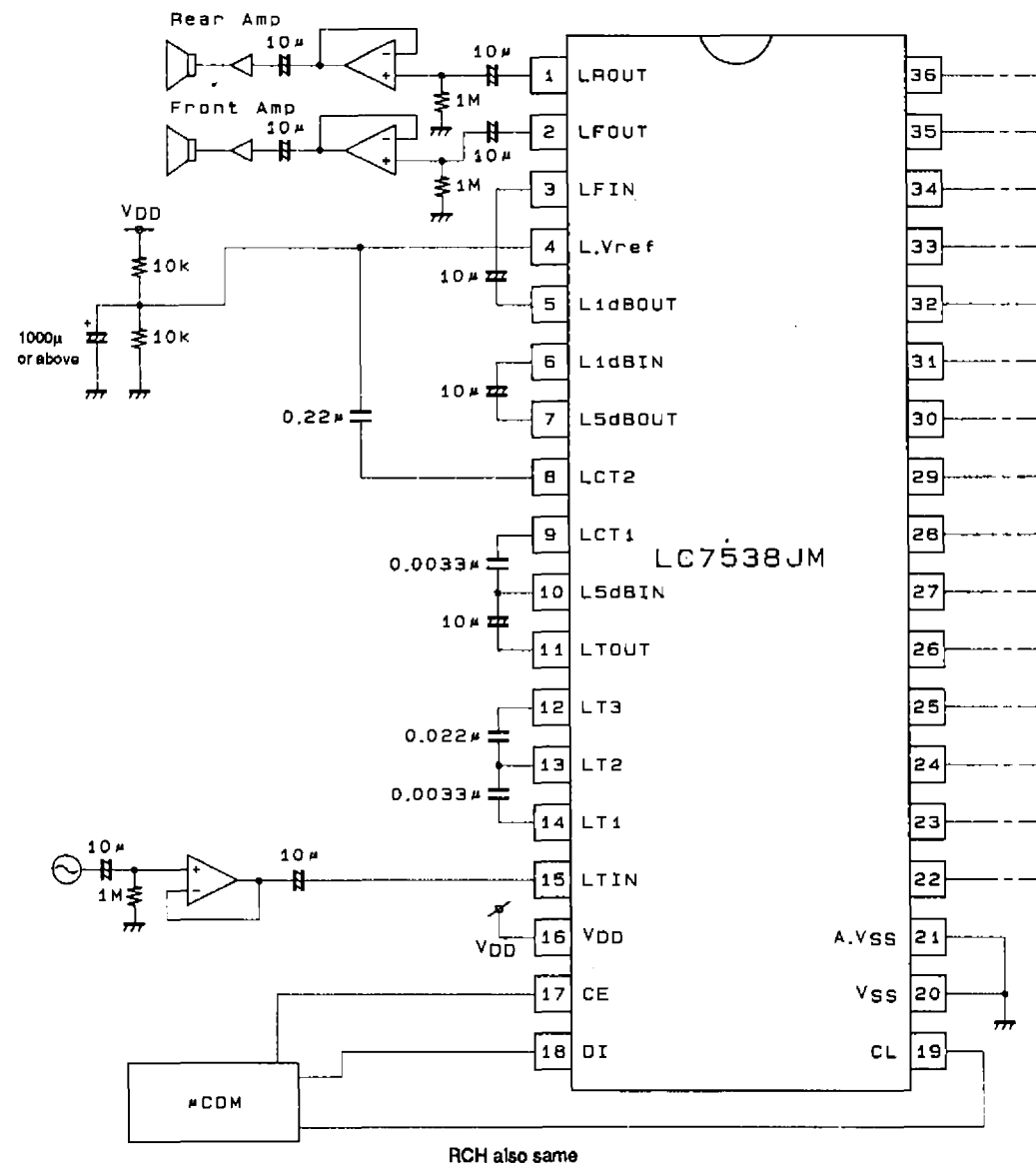| 1	1	1	0	: Step15
0	1	1	0	: Step14
1	0	1	0	: Step13
0	0	1	0	: Step12
1	1	0	0	: Step11
0	1	0	0	: Step10
1	0	0	0	: Step9
0	0	0	0	: Step8
1	0	0	1	: Step7
0	1	0	1	: Step6
1	1	0	1	: Step5
0	0	1	1	: Step4
1	0	1	1	: Step3
0	1	1	1	: Step2
1	1	1	1	: Step1

TREBLE Control

1	1	1	0	: Step15
0	1	1	0	: Step14
1	0	1	0	: Step13
0	0	1	0	: Step12
1	1	0	0	: Step11
0	1	0	0	: Step10
1	0	0	0	: Step9
0	0	0	0	: Step8
1	0	0	1	: Step7
0	1	0	1	: Step6
1	1	0	1	: Step5
0	0	1	1	: Step4
1	0	1	1	: Step3
0	1	1	1	: Step2
1	1	1	1	: Step1

5dB Step Control

0	0	0	0	: −75dB
1	0	0	0	: −70dB
0	1	0	0	: −65dB
1	1	0	0	: −60dB
0	0	1	0	: −55dB
1	0	1	0	: −50dB
0	1	1	0	: −45dB
1	1	1	0	: −40dB
0	0	0	1	: −35dB
1	0	0	1	: −30dB
0	1	0	1	: −25dB
1	1	0	1	: −20dB
0	0	1	1	: −15dB
1	0	1	1	: −10dB
0	1	1	1	: − 5dB
1	1	1	1	: 0dB

Fader Step Control

0	0	0	0	: −∞
1	0	0	0	: −60dB
0	1	0	0	: −45dB
1	1	0	0	: −35dB
0	0	1	0	: −25dB
1	0	1	0	: −20dB
0	1	1	0	: −18dB
1	1	1	0	: −16dB
0	0	0	1	: −14dB
1	0	0	1	: −12dB
0	1	0	1	: −10dB
1	1	0	1	: − 8dB
0	0	1	1	: − 6dB
1	0	1	1	: − 4dB
0	1	1	1	: − 2dB
1	1	1	1	: 0dB

A01027

Sample Application Circuit

Unit (resistance: Ω, capacitance: F)

A01028

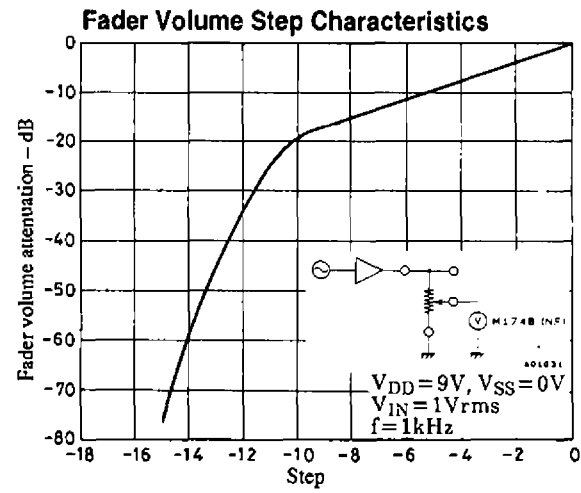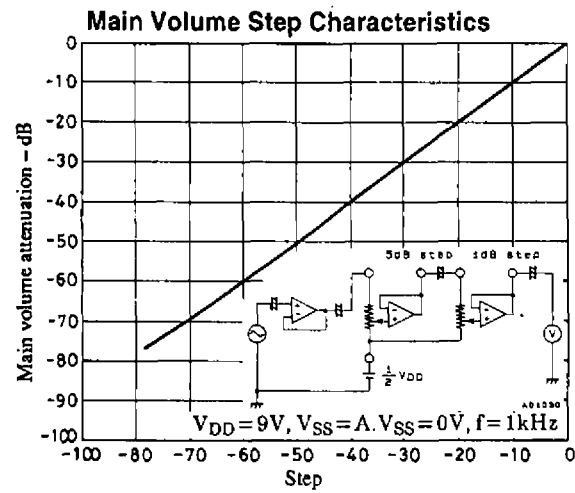RCH also same

Note: Bipolar electrolytic capacitors should be used as widely as possible where others are not recommended directly.

Tone Characteristics

Output level – dB

step15
step14
step13
step12
step11
step10
step9
step8
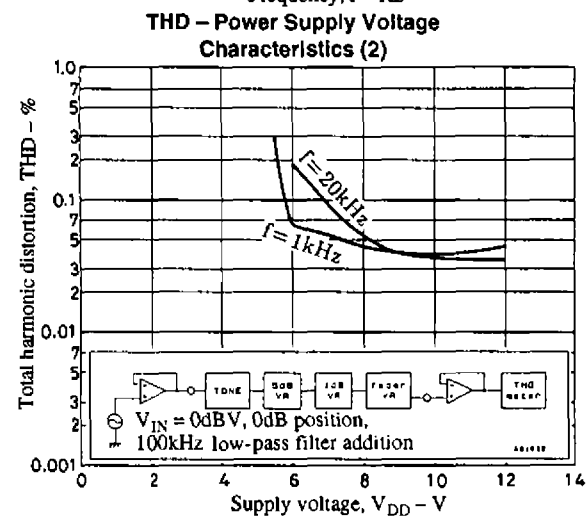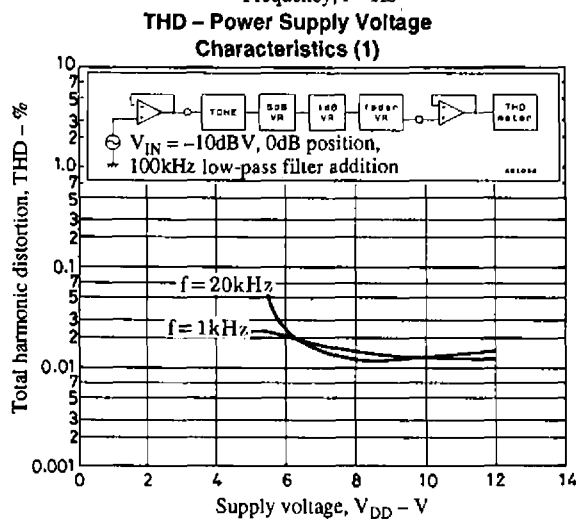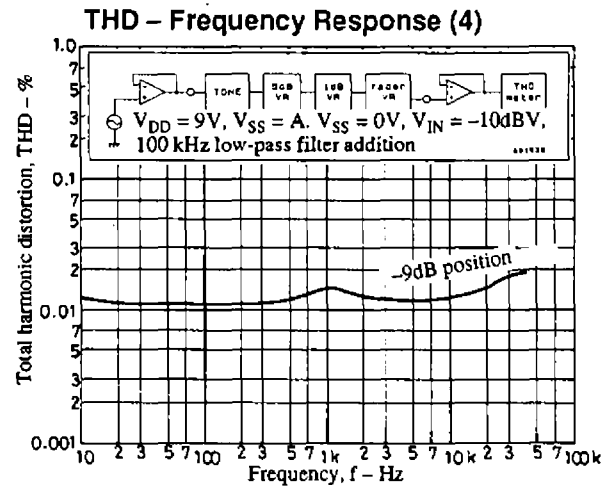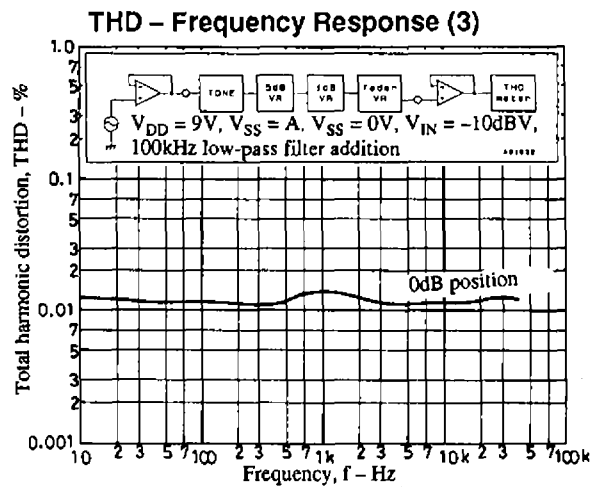step7
step6
step5
step4
step3
step2
step1

step16
step14
step13
step12
step11
step10
step9
step8
step7
step6
step5
step4
step3
step2
step1

0.0033μ   0.022μ
T2   T3

TIN   TOUT

½ VDD

T1

$V_{DD} = 9V$, $V_{SS} = A$. $V_{SS} = 0V$, –20dBV input

Unit (capacitance: F)

Frequency, f – Hz

Main Volume Step Characteristics

Main volume attenuation – dB

5dB step   1dB step

½ VDD

$V_{DD} = 9V$, $V_{SS} = A$. $V_{SS} = 0V$, $f = 1kHz$

Step

Fader Volume Step Characteristics

Fader volume attenuation – dB

M1748 (NS)

$V_{DD} = 9V$, $V_{SS} = 0V$
$V_{IN} = 1Vrms$
$f = 1kHz$

Step

THD – Frequency Response (1)

Total harmonic distortion, THD – %

0dB position

TONE   5dB VR   1dB VR   Fader VR   THD meter

$V_{DD} = 9V$, $V_{SS} = A$, $V_{SS} = 0V$, $V_{IN} = 0dBV$,
100 kHz low pulse addition

Frequency, f – Hz

THD – Frequency Response (2)

Total harmonic distortion, THD – %

–19dB position

TONE   5dB VR   1dB VR   Fader VR   THD meter

$V_{DD} = 9V$, $V_{SS} = A$, $V_{SS} = 0V$, $V_{IN} = 0dBV$,
100 kHz low pulse addition

Frequency, f – Hz

THD – Frequency Response (3)

$V_{DD} = 9V$, $V_{SS} = A$. $V_{SS} = 0V$, $V_{IN} = -10dBV$, 100kHz low-pass filter addition

Total harmonic distortion, THD – %

0dB position

Frequency, f – Hz

THD – Frequency Response (4)

$V_{DD} = 9V$, $V_{SS} = A$. $V_{SS} = 0V$, $V_{IN} = -10dBV$, 100 kHz low-pass filter addition

Total harmonic distortion, THD – %

–9dB position

Frequency, f – Hz

THD – Power Supply Voltage Characteristics (1)

$V_{IN} = -10dBV$, 0dB position, 100kHz low-pass filter addition

Total harmonic distortion, THD – %

f = 20kHz

f = 1kHz

Supply voltage, V_{DD} – V

THD – Power Supply Voltage Characteristics (2)

f = 20kHz

f = 1kHz

$V_{IN} = 0dBV$, 0dB position, 100kHz low-pass filter addition

Total harmonic distortion, THD – %

Supply voltage, V_{DD} – V

THD – Input Level Characteristics

Total harmonic distortion, THD – %

f = 1kHz

f = 20kHz

$V_{DD} = 9V$, $V_{SS} = A$. $V_{SS} = 0V$, 0dB position, 100 kHz low-pass filter addition

Input level, V_{IN} – dBV

Loudness Characteristics

Output level – dB

0dB
–5dB
–10dB
–15dB
–20dB
–25dB

$V_{DD} = 9V$, $V_{SS} = A$. $V_{SS} = 0V$

Frequency, f – Hz

Loudness External Constant Calculation Sample

First, refer to page 7 where the 5 dB step internal equivalent circuit for the LC7538JM is shown. Using this information, an external constant for loudness can be added to establish a simplified circuit for computation as shown in Figure 1. Computations gaining a 5 dB boost with f = 100 Hz using this configuration are shown in the following.
(f = 100 Hz and 5 dB boost)
Within figure 1, when R and C are defined as:
R1 = R2 = 10 kΩ
R3 = 1 kΩ
C1 =Z1, C2 = Z2, then the following equation can be established:

$$\underset{(\text{at} = 1\text{kHz})}{\text{VOUT}} = \frac{\dfrac{R2\,(R3 + Z2)}{R2 + R3 + Z2}}{\dfrac{R1 \cdot Z1}{R1 + Z1} + \dfrac{R2\,(R3 + Z2)}{R2 + R3 + Z2}} = -20\text{dB}$$
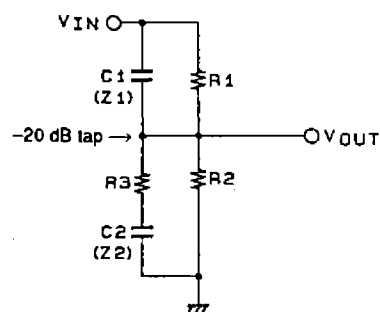
$$\underset{(\text{at} = 100\text{Hz})}{\text{VOUT}} = \frac{\dfrac{R2\,(R3 + 10 \cdot Z2)}{R2 + R3 + 10 \cdot Z2}}{\dfrac{R1 \cdot 10 \cdot Z1}{R1 + 10 \cdot Z1} + \dfrac{R2(R3 + 10 \cdot Z2)}{R2 + R3 + 10 \cdot Z2}} = -15\text{dB}$$

thereby resulting in,

$Z1 \neq 178.3$ kΩ and $Z2 = 176$ Ω.

Under such conditions where f = 1 kHz, specifications may be satisfied if C (capacitor) having these impedances is supplied externally. The end result is that C1 = 893 pF and C2 = 0.9 μF.

R1, R2 and R3 : LC7538JM on-chip resistors
C1 : External high-band compensation capacitor
C2 : External low-band compensation capacitor

Figure 1

Notes for Above Applications

- When the power supply is turned on, the internal analog switch becomes inexact. Until data is set, counter measures such as those required for muting are performed externally.

- In order to prevent crossover into the analog system of high-frequency digital signals transferred to the CL, DI and CE pins, transfer along these signal lines should occur along shielded lines, or the signal lines should be protected by using the grounding pattern or the circuit.

- For volume steps with large attenuation levels (over –20 dB), when the loudness circuit is off the high frequency region (above about 4 kHz) will be attenuated by about 3 dB relative to the low frequency region (about 400 Hz) due to the influence of the resistance of the loudness circuit analog switch. Therefore we recommend using tone control compensation together with the volume step described above.

- When sending data immediately after power on, send data as follows:

— When sending independent left and right data, send data at least four times.

— Alternatively, when sending data that drives the left and right channels at the same time, send the data at least twice.