

LM1866 Low Voltage AM/FM Receiver

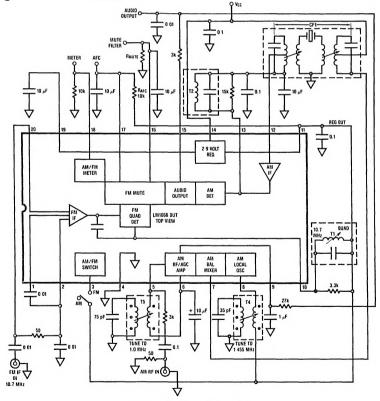
General Description

The LM1866 has been designed for high quality battery powered medium wave AM and FM receiver applications requiring operation down to 3V. The AM section contains a fully balanced, wide dynamic range, gain controlled mixer stage buffered from a single pin local oscillator. A two pin compound IF amplifier and internal detector provide a low distortion high level audio output. An AM/FM signal strength meter voltage is provided to a single output pin. The FM section contains a six stage limiting IF amplifier, quadrature detector, AFC output, deviation audio muting and noise operated audio muting. While designed for the high ripple, high battery impedance conditions found at the end of life for four "C" or "D" cells, the LM1866 will operate equally well at supply voltages up to 15V.

Features

- Operation from 3V to 5V
- □ Excellent power supply ripple rejection
- □ Fully balanced, wide dynamic range, AM mixer stage
- □ Internal AM detector for minimum tweet interference
- □ Single pole DC AM/FM mode switching
- Six stage FM IF limiting amplifier for excellent AM rejection
- "Soft" FM deviation and noise operated audio muting
- □ FM quadrature detector
- □ Single pin AM/FM meter output
- □ Single pin matched level AM/FM audio output

Block Diagram and Test Circuit



TL/H/7908-1

Coil Data: T2, Toko 159GC-A3785 CF1, Toko CFU-90D Order Number LM1866N See NS Package Number N20A

T1, Toko KAC K2318HM T4 = T5, Toko RBO6A5105 Toko America 1250 Feehanville Drive Mount Prospect, IL 60056 (312) 297-0070

Absolute Maximum Ratings

If Military/Aerospace specified devices are required, contact the National Semiconductor Sales Office/ Distributors for availability and specifications.

Supply Voltage (Pin 14) 15V Package Dissipation (Note 1) 1900 mW Storage Temperature Range -55°C to +150°C

Operating Temperature Range 0°C to +70°C

Lead Temperature (Soldering, 10 sec.) 260°C

Electrical Characteristics (Test Circuit, TA = 25°C)

Parameter	Conditions	Min	Тур	Max	Units
STATIC DC CHARACTERISTICS: e _{IN} = 0	, R _{MUTE} = 0Ω, V _{CC} = 6V				
Operating Supply Range, V14		3	6	15	V
Supply Current, I ₁₄	AM Mode		15	27	mA
Supply Current, I ₁₄	FM Mode		16	24	mA
Regulator Output Voltage, V11			2.9		V
Meter Output Voltage, V18	AM Mode		0	0.2	V
Meter Output Voltage, V18	FM Mode		0	0.2	V
AFC Output Voltage, V17	FM Mode	l	2.9	Į.	V
AM/FM Audio Output Resistance, R _o 15			3		kΩ
AM DYNAMIC CHARACTERISTICS: fam	= 1 MHz, f _{MOD} = 1 kHz, m = 0.3, V _{CC} = 6V				
Maximum Sensitivity	e _{AM} for e _o = 6 mV		9		μ٧
20 dB Quieting Sensitivity	e _{AM} for e _o = 20 dB S/N		25		μV
Signal to Noise Ratio	e _{AM} = 10 mV	40	50	ł	dB
Total Harmonic Distortion	e _{AM} = 10 mV		0.3		%
Total Harmonic Distortion	$e_{AM} = 10 \text{ mV}, m = 0.8$		1	2	%
Audio Output Level	e _{AM} = 10 mV	70	120		mV
Overload Distortion	$e_{AM} = 50 \text{ mV, m} = 0.8$		2	12	%
Meter Output Voltage	e _{AM} = 1 mV		2.0	3.0	v
Meter Output Voltage	e _{AM} = 50 mV		3.0	3.5	V
M DYNAMIC CHARACTERISTICS: f _{FM} =	= 10.7 MHz, $f_{MOD} = 400$ Hz, $\Delta f = \pm 75$ kHz,	V _{CC} = 6V			
-3 dB Limiting Sensitivity	e _{FM} for -3 dB Limiting Sensitivity		20	35	μ٧
Signal to Noise Ratio	e _{FM} = 10 mV	60	76		dB
AM Rejection	e _{FM} = 10 mV, 30% AM Mod	40	55		dB
Total Harmonic Distortion	e _{FM} = 10 mV		0.5	1	%
Audio Output Level	e _{FM} = 10 mV, 30% FM Mod	60	120		m∨
Meter Output Level	e _{FM} = 1 mV		1.3	2.3	V
Meter Output Level	e _{FM} = 50 mV		2.0	2.8	l v
± Deviation Mute (Notes 2, 4)	e _{FM} = 10 mV, R _{AFC} = 10k		40		kHz
R _{MUTE} for Noise Mute (Notes 3, 4)	Set e _{FM} for −3 dB Limiting Sensitivity	2	5	10	kΩ
Max Audio Mute Attenuation		60	75		dB

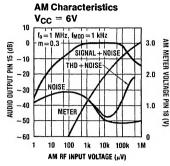
Note 1: Above $T_A = 25^{\circ}C$, derate based on $T_{J(max)} = 150^{\circ}C$ and $\theta_{JA} = 65^{\circ}C/W$.

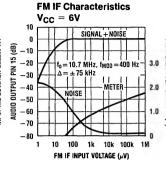
Note 2: $R_{MUTE} = 2 k\Omega$, $e_{FM} = 10$ mV, adjust center frequency for $V_{AFC} = V_{REG}$, record f_{FM} , adjust $\pm f_{FM}$ for >50 dB audio mute attenuation.

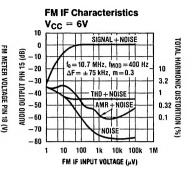
Note 3: Adjust R_{MUTE} from 2k to 10k for >50 dB audio mute attenuation. Set e_{FM} = 10 mV and check for mute off.

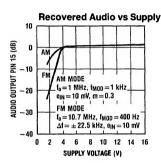
Note 4: When $R_{MUTE} = 0\Omega$, the deviation and noise operated mute functions are disabled. When $R_{MUTE} = 2 \, k\Omega$, only the noise mute function is disabled. The deviation mute bandwidth is set by the R_{AFC} resistor. The noise mute threshold is set by the R_{AUTE} resistor. Test circuit noise bandwidth characteristics prevent noise mute operation for IF input levels below the -3 dB limiting threshold. When the FM IF is used with a tuner, full noise mute capability is accessible (See Applications Information).

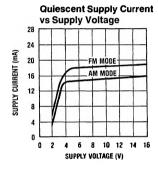
Typical Performance Characteristics (Test Circuit)











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Applications Information

(See Typical Applications and LM1866 Schematic Diagram)

VOLTAGE REGULATOR SECTION

Because of the wide supply voltage range and high ripple conditions expected in battery or low cost transformer supplies, the LM1866 uses a band gap referenced active voltage regulator which is externally compensated at pin 19. This capacitor, when made large enough, improves the supply rejection and decreases the noise bandwidth to a level well below the AM reception frequencies. A 0.1 μF capacitor will compensate the regulator for low noise operation while 50 μF (max) will improve supply rejection and the maximum FM audio mute attenuation characteristics. During power turn on, the pin 19 capacitor is quick-charged to its normal operating voltage so that the AM or FM sections are in operation before the audio amplifier turn on delay has timed out. See LM1895/LM2895 and LM1896/LM2896 data sheets for additional audio amplifier information.

AM SECTION

The AM section contains a fully balanced mixer stage with the RF input applied to a differential, diode degenerated, transistor pair at pins 5 and 6. DC feedback is provided by the loopstick secondary winding. The mixer output 1st IF transformer at pin 7 should be returned to $V_{\rm CC}$ at pin 14 to allow maximum undistorted output swing when tuning between stations. RF and AGC decoupling at pin 6 removes noise and lowers audio distortion.

The mixer upper pairs are switched differentially by a buffer amplifier from the pin 8 local oscillator. DC feedback is provided by the oscillator coil secondary winding to the pin 11 regulator voltage.

The oscillator frequency is given by:

$$f_O = \frac{0.159}{\sqrt{LC}},$$

and the peak swing is given by: $V_P = IZ$ ($I = 700 \mu A$, Z = tank impedance at resonance). V_P should be between 0.3V and 0.5V to maintain an undistorted output at low supplies.

The two stage AM IF amplifier at pins 12 and 13 requires output to input DC feedback and external decoupling. The IF gain is given by:

$$A_V = \frac{Z_L}{12}$$

Applications Information (Continued)

where Z_L equals resonant unloaded tank impedance in parallel with R_{EXT} . In most applications $Z_L = 10k$ and

$$Q_L = \frac{Z_L}{X_C} = 5$$

where R_{EXT} = an external IF gain setting resistor and X_{C} = impedance of tank tuning capacitor. A rule of thumb for setting the IF gain would be to adjust REXT for 20 dB audio S/N when the audio has dropped 10 dB below the level found at the AGC threshold. (Because of the low QL, a non-tuned coil is acceptable.)

The output of the IF amplifier drives an internal detector which is operating at low currents. This results in very low 2nd and 3rd IF harmonic radiation for minimal tweet interference

FM SECTION

The FM section contains a six stage limiting amplifier, quadrature detector. AFC output, deviation mute detector and a high frequency noise mute detector. (See Figure 1 for the Simplified Mute Circuit Schematic.) The output of the quadrature detector is split into three current source pairs. The ± audio current and internal load resistor R84 provide the audio output voltage via Q56 to pin 15. The ± AFC current, external load resistor (RAFC) and the 10 µF capacitor provide an audio decoupled AFC voltage to pin 17. The ± noise current and internal load resistor R114 provide a wideband detector output that is limited in frequency by CSTRAY. With the addition of internal C4 and R120 a band pass filter (fo = 1 MHz) is realized at the input of the peak to peak detector. The output current, flowing in resistor RMUTE and filtered by a capacitor, provides a mute voltage at pin 16. When the mute voltage rises to approximately one VBF, transistor Q139 will start to shunt the ± audio current away from R84, muting the audio output. The value of the RMITE resistor will determine the minimum audio signal to noise ratio at which one wishes to mute. The deviation mute detector will output a current only when the AFC voltage is offset above or below the VREG voltage. Load resistor R121 and transistor Q154 will convert this current to a mute voltage at pin 16. This is done to prevent interaction between the two detector output currents. The external RAFC resistor is used to set the deviation mute bandwidth so that the pin 16 mute voltage is one VBE at the desired frequency band edge. When disabling the mute functions, pin 16 is shorted to ground, preventing Q139 from becoming active.

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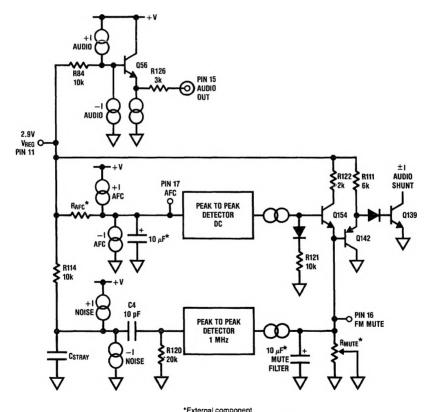


FIGURE 1. Simplified Mute Circuit Schematic

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Variable Tuning Capacitor

Capacitance: AM C1A 4 pF – 142 pF, C1B 4 – 60 pF FM 2.5 pF – 20 pF C1C, C1D

Type: QT-22124 Toko

Applications Information (Continued)

Applications	TABLE L Typical Applic	nuea) ation External Coil and Componen	t Salaction Guida		
Component	Typical Value		urpose		
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C1A, B, C, D	— 00000 0.04 F	AM/FM tuning capacitor			
R1, C2, C3	330Ω, 0.01 μF	FM IF decoupling, filter match and DC feedback			
C4	1 μF-10 μF	AM/RF/AGC decoupling			
R5, C5	27k, 1 μF	Sets AM AGC time constant			
R6	120k-150k	Optional: decreases AM audio output but improves AM meter threshold			
C6	0.1 μF	Regulator output decoupling			
C7, C8	0.1 μF, 10 μF	AM IF/audio decoupling			
R4 (R _{EXT})	15k	Sets AM IF gain			
R7, C15, C14	10Ω , 0.1 μF, 100 μF	Supply decoupling			
R3, C10	3k, 0.01 μF	Sets FM de-emphasis/AM smoothing			
		Audio post filter pole is given by:	$f = \frac{0.159}{1.000}$		
			H _T C10'		
		when $R_T = R3 + R_015 = R3 +$			
R _{MUTE} , C11	0 to 10k, 10 μF	Sets noise mute threshold, filter. 0Ω will turn off mute function.			
R _{AFC} , C12	10k, 10 μF	Sets deviation mute bandwidth, a	The state of the s		
C13	10 μF	Regulator decoupling and supply	rejection filter		
T1		T2	Т3		
r	7	F======	F		
	7 !				
i	147 2!!	1 5";	141 3 !! \$ 21		
! C _T	> 3 i	c_ o 3 i	! 		
	3ii!	2427	! T 3;;E !		
i L					
L		L			
	TL/H/7908-4	TL/H/7908-5	TL/H/7908-6		
$C_T = 82 pF$		$C_T = 180 pF$	C _T ≈ 50 pF		
Qu ≥ 70 f = 10.7 MHz		Qu = 14	Qu = 80		
	t = 10.7 MHz				
Apollo Electronics Corp.					
T4 and T5 MW Oscillator Coil CF1					
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	TL/H/7908-7	∇			
L = 360 µH		<u> </u>			
f = 796 kHz			TL/H/7908-8		
Qu = 160	= 985 kHz – 2105 kHz	Toko CFU-090D or equivalent f = 455 kHz, BW > 4.8 kHz			
Part no. RBO		1 - 455 KHZ, BW > 4.6 KHZ			
MW Antenna Coil		Dummy Antenna for FM			
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S II ← ANTENNA INPUT					
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	104T \\ \\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\\		TI		
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	0—1 	8-9	TI /H/7908-10		

L7 SWG #20, N = 3 1/2T, ID = 5 mm L5 SWG #20, N = 3 1/2T, ID = 5 mm L6 L = 0.44 μH, N = 4 1/2T, Qu = 70

Tuning freq. = 530 kHz-1650 kHz

 $L = 650 \, \mu H$

f = 796 kHz

Qu = 200

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