

LM124/LM224/LM324, LM124A/LM224A/LM324A, LM2902 Low Power Quad Operational Amplifiers

General Description

The LM124 series consists of four independent, high gain, internally frequency compensated operational amplifiers which were designed specifically to operate from a single power supply over a wide range of voltages. Operation from split power supplies is also possible and the low power supply current drain is independent of the magnitude of the power supply voltage.

Application areas include transducer amplifiers, DC gain blocks and all the conventional op amp circuits which now can be more easily implemented in single power supply systems. For example, the LM124 series can be directly operated off of the standard $\pm 5~V_{DC}$ power supply voltage which is used in digital systems and will easily provide the required interface electronics without requiring the additional $\pm 15~V_{DC}$ power supplies.

Unique Characteristics

- In the linear mode the input common-mode voltage range includes ground and the output voltage can also swing to ground, even though operated from only a single power supply voltage.
- The unity gain cross frequency is temperature compensated.
- The input bias current is also temperature compensated.

Advantages

- Eliminates need for dual supplies
- Four internally compensated op amps in a single package
- Allows directly sensing near GND and V_{OUT} also goes to GND
- Compatible with all forms of logic
- Power drain suitable for battery operation

Features

- Internally frequency compensated for unity gain
- Large DC voltage gain

100 dE

Wide bandwidth (unity gain)

1 MHz

(temperature compensated)

■ Wide power supply range:

3 V_{DC} to 32 V_{DC}

Single supply or dual supplies

 \pm 1.5 V_{DC} to \pm 16 V_{DC}

- $\hfill \square$ Very low supply current drain (800 $\mu A)$ —essentially independent of supply voltage (1 mW/op amp at $+~5~V_{DC})$
- Low input biasing current (temperature compensated)

45 nA_{DC}

■ Low input offset voltage and offset current

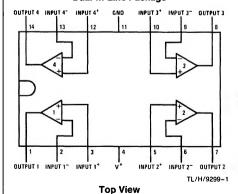
2 mV_{DC} 5 nA_{DC}

- ☐ Input common-mode voltage range includes ground
- Differential input voltage range equal to the power supply voltage
- Large output voltage swing 0 V_{DC}

0 V_{DC} to V^+ - 1.5 V_{DC}

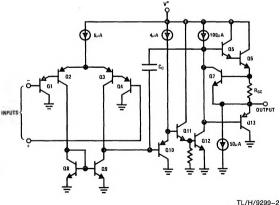
Connection Diagram

Dual-In-Line Package



Order Number LM124J, LM124AJ, LM224J, LM224AJ, LM324J, LM324AJ, LM324AM, LM324AM, LM2902M, LM324N, LM324AN or LM2902N See NS Package Number J14A. M14A or N14A

Schematic Diagram (Each Amplifier)



TL/H/9299

Absolute Maximum Ratings

If Military/Aerospace specified devices are required, contact the National Semiconductor Sales Office/Distributors for availability and specifications. (Note 9)

_			_			_			-						_		
	LM2902	-65°C to +150°C	260°C			260°C		215°C	220°C	duct Reliability" for							
	LM124/LM224/LM324 LM124A/LM224A/LM324A	-65°C to +150°C	conds) 260°C			260°C		215°C	220°C	nods and Their Effect on Proc	nount devices.						
	3	Storage Temperature Range	Lead Temperature (Soldering, 10 seconds) 260°C	Soldering Information	Dual-In-Line Package	Soldering (10 seconds)	Small Outline Package	Vapor Phase (60 seconds)	Infrared (15 seconds)	See AN-450 "Surface Mounting Methods and Their Effect on Product Reliability" for	other methods of soldering surface mount devices.	ESD rating to be determined.					
	LM2902	26 V _{DC} or ±13 V _{DC}	26 V _{DC}	-0.3 V _{DC} to +26 V _{DC}		1130 mW	1260 mW	800 mW			Continuous		50 mA	-40°C to +85°C			
	LM124/LM224/LM324 LM124A/LM224A/LM324A	32 V_{DC} or \pm 16 V_{DC}	32 V _{DC}	-0.3 V _{DC} to +32 V _{DC}		1130 mW	1260 mW	800 mW			Continuous		50 mA		0°C to +70°C	-25°C to +85°C	-55°C to +125°C
(Note 9)		Supply Voltage, V ⁺	Differential Input Voltage	Input Voltage	Power Dissipation (Note 1)	Molded DIP	Cavity DIP	Small Outline Package	Output Short Circuit to GND	Oniput Gilor Folicuit to Give	$V^+ \le 15 \text{ Vnc and } T_A = 25^{\circ}\text{C}$	Input Current	(V _{IN} < -0.3 V _{DC}) (Note 3)	Operating Temperature Range	LM324/LM324A	LM224/LM224A	LM124/LM124A

Electrical Characteristics $V^+ = +5.0 \text{ V}_{DC}$, (Note 4), unless otherwise stated

o de care de c	onditions.	LM124A LM224A	24A	LM2	LM224A	LM324A	24A	LM124/LM224	LM224	LM	LM324	LM	LM2902	l diei
raiailielei	Silonipio	Min Typ	Max	Min Typ	Мах	Min Typ	Max	Min Typ	Мах	Min Typ	Мах	Min Typ	Max	2
Input Offset Voltage	(Note 5)	±1	±2	±1	∓3	+2	∓3	75	±5	±2	17	±2	1	mV _{DC}
Input Bias Current (Note 6)	$l_{IN(+)}$ or $l_{IN(-)}$, $V_{CM} = 0V$	20	20	40	80	45	100	45	150	45	250	45	250	nApc
Input Offset Current	$l_{IN(+)} - l_{IN(-)}, V_{CM} = 0V$	±2	±10	7	±15	± 2	∓30	€∓	∓30	45	± 20	72	7 ∓	nApc
Input Common-Mode Voltage Range (Note 7)	$V^{+} = 30 V_{DC},$ (LM2902, $V^{+} = 26 V_{DC})$	0	V ⁺ -1.5	0	V ⁺ -1.5 0		V ⁺ -1.5	0	V ⁺ -1.5	0	V ⁺ -1.5	0	V ⁺ -1.5 V _{DC}	V _{DC}
Supply Current	$R_L = \infty, V^+ = 30V,$ (LM2902 $V^+ = 26V$)	1.5	က	1.5	ဗ	1.5	ဇ	1.5	3	1.5	3	1.5	ဇ	<
	$R_L = \infty$ On All Op Amps Over Full Temperature Range	0.7	1.2	0.7	1.2	0.7	1.2	0.7	1.2	0.7	1.2	0.7	1.2	DO
Large Signal Voltage Gain	$V^+ = 15 V_{DC}$, $R_L \ge 2 k\Omega$, $(V_O = 1 V_{DC} \text{ to } 11 V_{DC})$	50 100		50 100		25 100		50 100		25 100		25 100		V/m/V
Common-Mode Rejection Ratio	DC, $V_{CM} = 0V$ to $V^+ - 1.5 V_{DC}$	70 85		70 85		65 85		70 85		65 85		50 70		쁑
Power Supply Rejection Ratio	DC, V ⁺ = 5 V _{DC} to 30 V _{DC} (LM2902, V ⁺ = 5 V _{DC} to 26 V _{DC})	65 100		65 100		65 100		65 100		65 100		50 100		용

Electrical Ch	Electrical Characteristics $V^+ = +5.0 V_{DC}$ (Note 4) unless otherwise stated (Continued)	e 4) unless other	rwise s	tated (Conti	(penu									
Parameter	Conditions	LM124A		LM224A		LM324A	L	LM124/LM224		LM324		LM2	LM2902	Inite
		Min Typ Max		Min Typ Max		Min Typ Max		Min Typ Max	Min	Min Typ Max	×	Min Typ	Мах	
Amplifier-to-Amplifier Coupling (Note 8)	f = 1 kHz to 20 kHz, (Input Referred)	-120		-120		-120		-120		-120		-120	0	dВ
Output Current Source	Output Current Source $V_{IN}^+ = 1 V_{DC}$, $V_{IN}^+ = 0 V_{DC}$, $V_{T}^+ = 15 V_{DC}$, $V_{DC}^- = 2 V_{DC}$	20 40	20	40	20	40	8	40	50	40	2	20 40		0.0
Sink	$V_{IN}^{-} = 1 V_{DC}, V_{IN}^{+} = 0 V_{DC}, V_{T}^{+} = 15 V_{DC}, V_{O} = 2 V_{DC}$	10 20	10	20	9	20	5	20	9	20	-	10 20		20,
	$V_{IN}^{-} = 1 V_{DC}, V_{IN}^{+} = 0 V_{DC}, V_{V}^{+} = 15 V_{DC}, V_{O} = 200 \text{ m/DC}$	12 50	12	90	12	50	12	50	12	50	-	12 50		мApc
Short Circuit to Ground	(Note 2) $V^+ = 15 V_{DC}$, $T_A = 25^{\circ}C$	40 60		40 60		40 60		40 60		40 60		40	09	mApc
Input Offset Voltage	(Note 5)	±4	_	+4		± 2		1		6∓	6		±10	mVpc
Input Offset Voltage Drift	$R_{S} = 0\Omega$	±7 ±20	0	±7 ±20		±7 ±30		±7		7±		4 7		μV/°C
Input Offset Current	$l_{IN(+)} - l_{IN(-)}, V_{CM} = 0V$	∓30	0	∓30		±75		±100		±150	20	±45	±200	nApc
Input Offset Current Drift	$R_{S} = 0.0$	±10 ±200	00	±10 ±200	0	±10 ±200		±10		±10		±10		pA _{DC} /°C
Input Bias Current	lin(+) or lin(-)	40 100		40 100		40 200		40 300		40 500	0	40	200	nApc
Input Common-Mode Voltage Range (Note 7)	$V^{+} = +30 V_{DC}$ (LM2902, $V^{+} = 26 V_{DC}$)	0 V ⁺ -2	-2 0	V ⁺ -2	2 0	V ⁺ -2	0	V ⁺ -2	0	V ⁺ -2	-2 0		V ⁺ -2	V _{DC}
Large Signal Voltage Gain	$\begin{array}{l} V^+ = +15\text{Vpc} \\ (\text{Vo Swing} = 1\text{Vpc} \text{to}11\text{Vpc}) \\ \text{R}_L \geq 2\text{k}\Omega \end{array}$	25	25		15		25		15		-	15		V/mV
Output Voltage VoH	$V^{+} = +30 V_{DC}, R_{L} = 2 k\Omega$	26	26		26		56		26		22	CJ.		
Swing	$R_{L} \ge 10 k\Omega$ (LM2902, V $^{+} = 26 V_{DC}$)	27 28	27	28	27	28	27	28	27	28	- 53	3 24		VDC
VOL	$V^+ = 5 V_{DC}$, $R_L \ge 10 k\Omega$	5 20		5 20	\Box	5 20		5 20		5 20		5	100	mV _{DC}

Electrical Characteristics $V^+ = +5.0 V_{DC}$ (Note 4) unless otherwise stated (Continued)

Linite		Δ.Δ.	30	
LM2902	Min Typ Max	20	8	
5	Min	10 20	5	
	Мах			
LM324	Тур	10 20	8	
	Min	10	2	
M224	Max			
24/LF	Тур	20	80	
LM	Min	10 20	2	
LM324A LM124/LM224	Мах			
.M324	Тур	10 20	8	
_	Min	10	2	
4	Мах			
LM224A	Тур	10 20	8	
_	Min	10	ro	
¥.	Max			
LM124A	Тур	20	10 15	
	Min	10	10	
Conditions		$V_{IN}^{+} = +1 V_{DC},$ $V_{IN}^{-} = 0 V_{DC}, V^{+} = 15 V_{DC}$	$V_{IN}^{-} = +1 V_{DC},$ $V_{IN}^{+} = 0 V_{DC}, V^{+} = 15 V_{DC}$	
		ant Source V _O = 2 V _{DC}		
į	5	Sour	Sink	
Parame		Output Current		

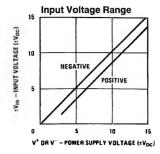
Note 1: For operating at high temperatures, the LM324/LM324A, LM2902 must be derrated based on a + 125°C maximum junction temperature and a thermal resistance of 86°C/W which applies for the device soldered in a printed circuit board, operating in a still air ambient. The LM224/LM224A and LM124/LM124A can be derated based on a +150°C maximum junction temperature. The dissipation is the total of all four amplifiers—use external resistors, where possible, to allow the amplifier to saturate of to reduce the power which is dissipated in the integrated circuit Note 2: Short circuits from the output to V + can cause excessive heating and eventual destruction. When considering short circuits to ground, the maximum output current is approximately 40 mA independent of the magnitude of Note 3: This input current will only exist when the voltage at any of the input leads is driven negative. It is due to the collector-base junction of the input PNP transistors becoming forward biased and thereby acting as input diode clamps. In addition to this diode action, there is also lateral NPN parasitic transistor action on the IC chip. This transistor action can cause the output voltages of the op amps to go to the V ⁺ voltage level (or to ground for a large overdrive) for the time duration that an input is driven negative. This is not destructive and normal output states will re-establish when the input voltage, which was negative, again returns to a value greater than -0.3 V_{DC} (at 25°C). **Note 4:** These specifications are limited to -55° C $\leq T_{A} \leq +125^{\circ}$ C for the LM124/LM124A. With the LM224/LM224A, all temperature specifications are limited to -25° C $\leq T_{A} \leq +85^{\circ}$ C, the LM324/LM324A temperature V $^+$. At values of supply voltage in excess of +15 V_{DC} , continuous short-circuits can exceed the power dissipation ratings and cause eventual destruction. Destructive dissipation can result from simultaneous shorts on all amplifiers. specifications are limited to 0° C $\leq T_{A} \leq +70^{\circ}$ C, and the LM2902 specifications are limited to -40° C $\leq T_{A} \leq +85^{\circ}$ C.

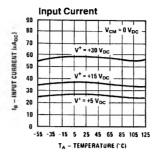
Note 5: Vo = 1.4 Vpc, R_S = 0.0 with V⁺ from 5 Vpc, to 30 Vpc; and over the full input common-mode range (0 Vpc to V⁺ - 1.5 Vpc) at 25°C; for LM2902, V⁺ from 5 Vpc to 26 Vpc.

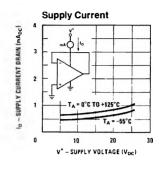
Note 7: The input common-mode voltage of either input signal voltage should not be allowed to go negative by more than 0.3V (at 25°C). The upper end of the common-mode voltage range is V⁺ – 1.5V (at 25°C), but either or both Note 6: The direction of the input current is out of the IC due to the PNP input stage. This current is essentially constant, independent of the state of the output so no loading change exists on the input lines.

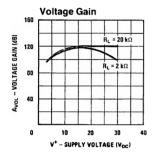
Note 8: Due to proximity of external components, insure that coupling is not originating via stray capacitance between these external parts. This typically can be detected as this type of capacitance increases at higher frequencies. Note 9: Refer to RETS124AX for LM124A military specifications and refer to RETS124X for LM124 military specifications. inputs can go to $+32~{
m V}_{
m DC}$ without damage ($+26~{
m V}_{
m DC}$ for LM2902), independent of the magnitude of V $^+$

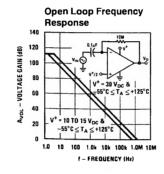
Typical Performance Characteristics

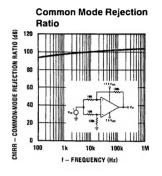


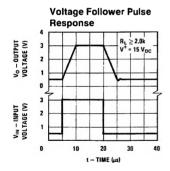


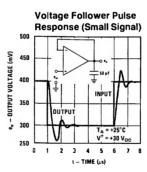


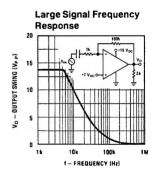


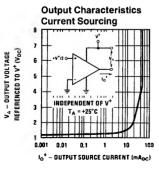


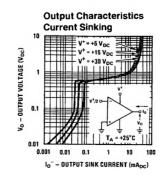


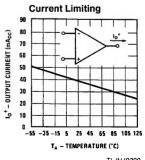




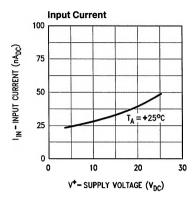


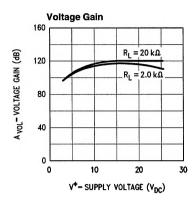






Typical Performance Characteristics (LM2902 only)





TL/H/9299-4

Application Hints

The LM124 series are op amps which operate with only a single power supply voltage, have true-differential inputs, and remain in the linear mode with an input common-mode voltage of 0 V_{DC}. These amplifiers operate over a wide range of power supply voltage with little change in performance characteristics. At 25°C amplifier operation is possible down to a minimum supply voltage of 2.3 V_{DC}.

The pinouts of the package have been designed to simplify PC board layouts. Inverting inputs are adjacent to outputs for all of the amplifiers and the outputs have also been placed at the corners of the package (pins 1, 7, 8, and 14).

Precautions should be taken to insure that the power supply for the integrated circuit never becomes reversed in polarity or that the unit is not inadvertently installed backwards in a test socket as an unlimited current surge through the resulting forward diode within the IC could cause fusing of the internal conductors and result in a destroyed unit.

Large differential input voltages can be easily accommodated and, as input differential voltage protection diodes are not needed, no large input currents result from large differential input voltages. The differential input voltage may be larger than V+ without damaging the device. Protection should be provided to prevent the input voltages from going negative more than $-0.3~\rm V_{DC}$ (at 25°C). An input clamp diode with a resistor to the IC input terminal can be used.

To reduce the power supply drain, the amplifiers have a class A output stage for small signal levels which converts to class B in a large signal mode. This allows the amplifiers to both source and sink large output currents. Therefore both NPN and PNP external current boost transistors can be used to extend the power capability of the basic amplifiers. The output voltage needs to raise approximately 1 diode drop above ground to bias the on-chip vertical PNP transistor for output current sinking applications.

For ac applications, where the load is capacitively coupled to the output of the amplifier, a resistor should be used, from the output of the amplifier to ground to increase the class A bias current and prevent crossover distortion.

Where the load is directly coupled, as in dc applications, there is no crossover distortion.

Capacitive loads which are applied directly to the output of the amplifier reduce the loop stability margin. Values of 50 pF can be accommodated using the worst-case non-inverting unity gain connection. Large closed loop gains or resistive isolation should be used if larger load capacitance must be driven by the amplifier.

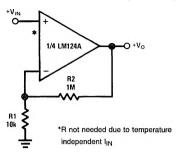
The bias network of the LM124 establishes a drain current which is independent of the magnitude of the power supply voltage over the range of from 3 V_{DC} to 30 V_{DC} .

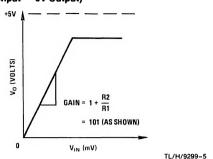
Output short circuits either to ground or to the positive power supply should be of short time duration. Units can be destroyed, not as a result of the short circuit current causing metal fusing, but rather due to the large increase in IC chip dissipation which will cause eventual failure due to excessive junction temperatures. Putting direct short-circuits on more than one amplifier at a time will increase the total IC power dissipation to destructive levels, if not properly protected with external dissipation limiting resistors in series with the output leads of the amplifiers. The larger value of output source current which is available at 25°C provides a larger output current capability at elevated temperatures (see typical performance characteristics) than a standard IC op amp.

The circuits presented in the section on typical applications emphasize operation on only a single power supply voltage. If complementary power supplies are available, all of the standard op amp circuits can be used. In general, introducing a pseudo-ground (a bias voltage reference of V $^+$ /2) will allow operation above and below this value in single power supply systems. Many application circuits are shown which take advantage of the wide input common-mode voltage range which includes ground. In most cases, input biasing is not required and input voltages which range to ground can easily be accommodated.

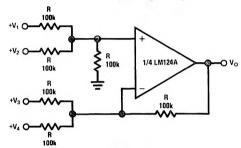
Typical Single-Supply Applications ($V^+ = 5.0 V_{DC}$)

Non-Inverting DC Gain (0V Input = 0V Output)

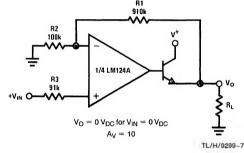




DC Summing Amplifier $(V_{IN'S} \ge 0 V_{DC} \text{ and } V_O \ge V_{DC})$

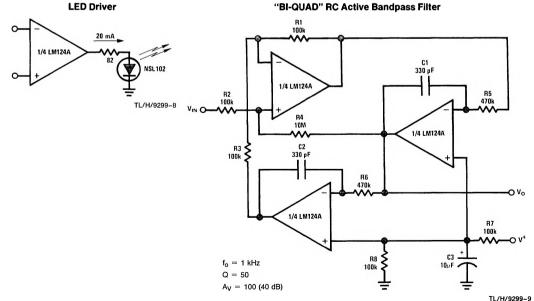


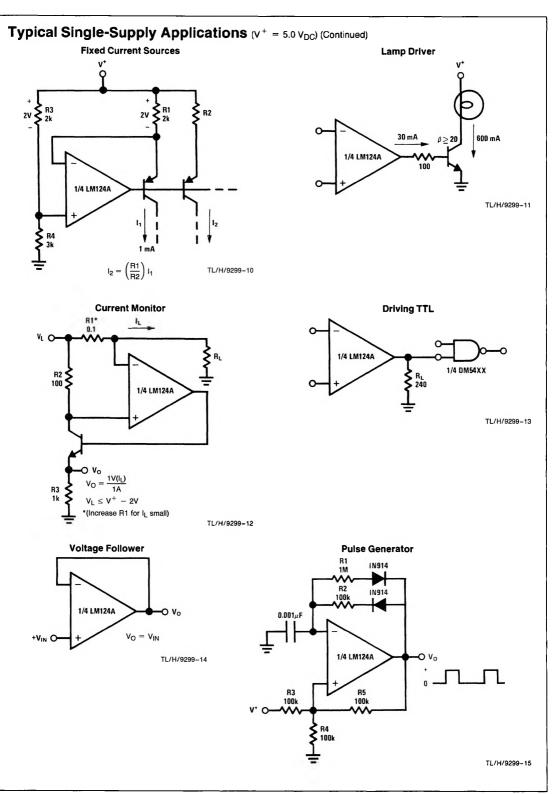
Power Amplifier



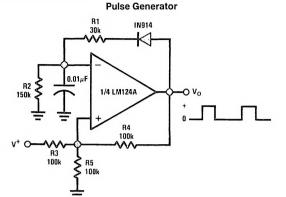
TL/H/9299-6 Where: $V_O = V_1 + V_2 - V_3 - V_4$ $(V_1 + V_2) \ge (V_3 + V_4)$ to keep $V_O > 0$ V_{DC}

"BI-QUAD" RC Active Bandpass Filter





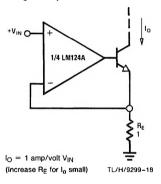
Squarewave Oscillator R1 100k 1/4 LM124A 100k R3 100k



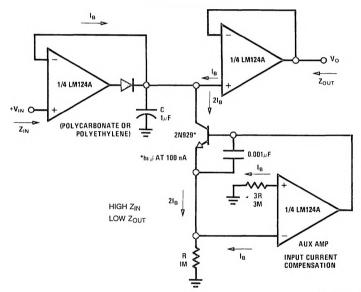
TL/H/9299-16

TL/H/9299-17

High Compliance Current Sink

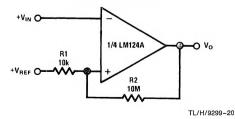


Low Drift Peak Detector

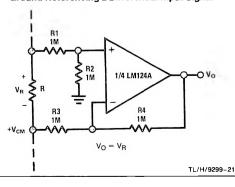


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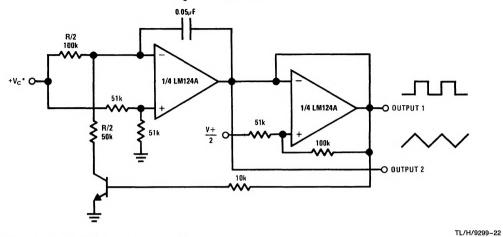
Comparator with Hysteresis



Ground Referencing a Differential Input Signal

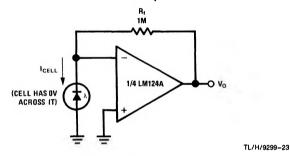


Voltage Controlled Oscillator Circuit

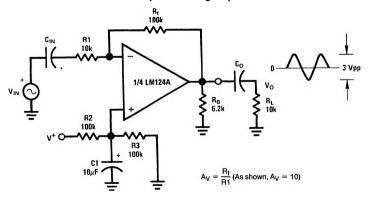


*Wide control voltage range: 0 $V_{DC} \le V_{C} \le$ 2 (V $^+$ -1.5 V_{DC})

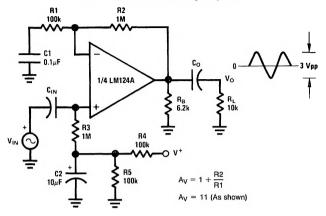
Photo Voltaic-Cell Amplifier



AC Coupled Inverting Amplifier

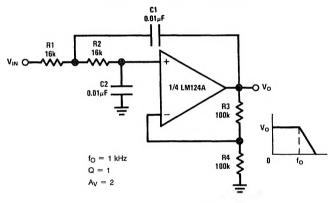


AC Coupled Non-Inverting Amplifier



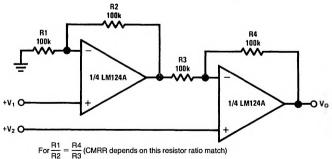
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DC Coupled Low-Pass RC Active Filter



TL/H/9299-26

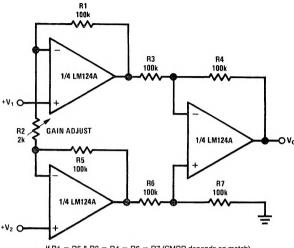
High Input Z, DC Differential Amplifier



$$V_0 = 1 + \frac{R4}{R3}(V_2 - V_1)$$

As shown: $V_0 = 2(V_2 - V_1)$

High Input Z Adjustable-Gain **DC Instrumentation Amplifier**



If R1 = R5 & R3 = R4 = R6 = R7 (CMRR depends on match)

$$V_0 = 1 + \frac{2R1}{R2}(V_2 - V_1)$$

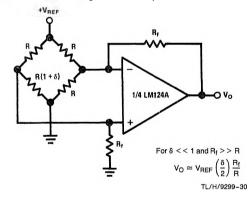
As shown $V_0 = 101 (V_2 - V_1)$

Using Symmetrical Amplifiers to Reduce Input Current (General Concept)

1/4 LM124A 2N929 0.001µF *hi β AT 50 nA AUX AMP INPUT CURRENT COMPENSATION

TL/H/9299-29

Bridge Current Amplifier



Typical Single-Supply Applications (V⁺ = 5.0 V_{DC}) (Continued)

Bandpass Active Filter

