



LM78G/LM79G 4-Terminal Adjustable Voltage Regulators

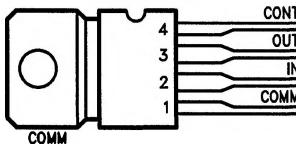
General Description

The LM78G and LM79G are 4-terminal adjustable voltage regulators. They are designed to deliver continuous load currents of up to 1.0A with a maximum input voltage of +40V for the positive regulator LM78G and -40V for the negative regulator LM79G. Output current capability can be increased to greater than 1.0A through use of one or more external transistors. The output voltage range of the LM78G positive voltage regulator is +5V to +30V and the output voltage range of the negative LM79G is -30V to -2.55V. For systems requiring both a positive and negative, the LM78G and LM79G are excellent for use as a dual tracking regulator with appropriate external circuitry.

Features

- Output current in excess of 1A
- LM78G positive output +5V to +30V
- LM79G negative output -30V to -2.55V
- Internal thermal overload protection
- Internal short circuit protection
- Output transistor safe-area protection

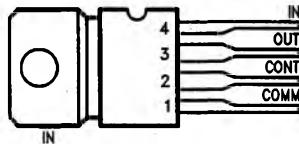
Connection Diagrams



Top View

TL/H/10054-1

Heat sink tabs connected to common through device substrate.



Top View

TL/H/10054-2

Heat sink tabs connected to input through device substrate. Not recommended for direct electrical connection.

**Order Number LM78GCP or LM79GCP
See NS Package Number P04A**

Absolute Maximum Ratings

If Military/Aerospace specified devices are required, please contact the National Semiconductor Sales Office/Distributors for availability and specifications.

Storage Temperature Range -65°C to $+150^{\circ}\text{C}$

Operating Junction Temperature Range 0°C to $+150^{\circ}\text{C}$

Lead Temperature (Soldering, 10 sec.) 265°C

Power Dissipation	Internally Limited
Input Voltage LM78G	$+40\text{V}$
LM79G	-40V
Control Lead Voltage LM78G	$0\text{V} \leq V^+ \leq V_O$
LM79G	$V_O^- \leq V^- \leq 0\text{V}$

LM78G

Electrical Characteristics

$0^{\circ}\text{C} \leq T_A \leq 125^{\circ}\text{C}$, $C_I = 0.33\ \mu\text{F}$, $C_O = 0.1\ \mu\text{F}$, $V_I = 10\text{V}$, $I_O = 500\ \text{mA}$, Test Circuit 1, unless otherwise specified

Symbol	Parameter	Conditions (Notes 1, 3)		Min	Typ	Max	Units
V_{IR}	Input Voltage Range	$T_J = 25^{\circ}\text{C}$		7.5		40	V
V_{OR}	Output Voltage Range	$V_I = V_O + 5.0\text{V}$		5.0		30	V
V_O	Output Voltage Tolerance	$(V_O + 3.0\text{V}) \leq V_I \leq (V_O + 15\text{V}),$ $5.0\ \text{mA} \leq I_O \leq 1.0\ \text{A}$ $P_D \leq 15\text{W}$, V_I Max = 38V	$T_J = 25^{\circ}\text{C}$			4.0	% V_O
						5.0	
V_{OLINE}	Line Regulation	$T_J = 25^{\circ}\text{C}, V_O \leq 10\text{V}$ $(V_O + 2.5\text{V}) \leq V_I \leq (V_O + 20\text{V})$				1.0	% V_O
$V_{O\text{ LOAD}}$	Load Regulation	$T_J = 25^{\circ}\text{C},$	$250\ \text{mA} \leq I_O \leq 750\ \text{mA}$			1.0	% V_O
		$V_I \leq V_O + 5.0\text{V}$	$5.0\ \text{mA} \leq I_O \leq 1.5\text{A}$			2.0	
I_C	Control Lead Current	$T_J = 25^{\circ}\text{C}$			1.0	5.0	μA
						8.0	
I_Q	Quiescent Current	$T_J = 25^{\circ}\text{C}$			3.2	6.0	mA
						7.0	
$\Delta V_I / \Delta V_O$	Ripple Rejection	$8.0\text{V} \leq V_I \leq 18\text{V}$, $f = 2400\ \text{Hz}$, $V_O = 5.0\text{V}$, $I_C = 350\ \text{mA}$		68	78		dB
N_O	Noise	$T_J = 25^{\circ}\text{C}$, $10\ \text{Hz} < f < 100\ \text{kHz}$, $V_O = 5.0\text{V}$, $I_O = 5.0\ \text{mA}$			8.0	40	$\mu\text{V}/V_O$
V_{DO}	Dropout Voltage (Note 2)				2.0	2.5	V
I_{OS}	Output Short Circuit Current	$T_J = 25^{\circ}\text{C}$, $V_I = 30\text{V}$			0.750	1.2	A
I_{pk}	Peak Output Current	$T_J = 25^{\circ}\text{C}$		1.3	2.2	3.3	A
$\Delta V_O / \Delta T$	Average Temperature Coefficient of Output Voltage	$V_O = 5.0\text{V}$,	$T_A = -55^{\circ}\text{C}$ to $+25^{\circ}\text{C}$			0.4	$\text{mV}/^{\circ}\text{C}/V_O$
		$I_O = 5.0\ \text{mA}$	$T_A = 25^{\circ}\text{C}$ to $+125^{\circ}\text{C}$			0.3	
V_C	Control Lead Voltage (Reference)	$T_J = 25^{\circ}\text{C}$		4.8	5.0	5.2	V
				4.75		5.25	

Note 1: V_O is defined for the LM78G as $V_O = \frac{R_1 + R_2}{R_2} (5.0)$;

Note 2: Dropout Voltage is defined as that input/output voltage differential which causes the output voltage to decrease by 5% of its initial value.

Note 3: All characteristics except noise voltage and ripple rejection ratio are measured using pulse techniques ($t_w \leq 10\ \text{ms}$, duty cycle $\leq 5\%$). Output voltage changes due to changes in internal temperature must be taken into account separately.

LM79G

Electrical Characteristics $0^\circ\text{C} \leq T_A \leq 125^\circ\text{C}$ for LM79G, $V_I = -10\text{V}$, $I_O = 500\text{ mA}$, $C_i = 2.0\text{ }\mu\text{F}$,
 $C_o = 0.1\text{ }\mu\text{F}$, Test Circuit 2 and Note 3, unless otherwise specified

Symbol	Parameter	Conditions (Note 1)		Min	Typ	Max	Units
V_{IR}	Input Voltage Range	$T_J = 25^\circ\text{C}$		-40		-7.0	V
V_{OR}	Nominal Output Voltage Range	$V_I = V_O - 5.0\text{V}$		-30		-2.55	V
V_O	Output Voltage Tolerance	$(V_O - 15\text{V}) \leq V_I \leq (V_O - 3.0\text{V})$, $5.0\text{ mA} \leq I_O \leq 1.0\text{A}$	$T_J = 25^\circ\text{C}$			4.0	% V_O
		$P_D \leq 15\text{W}$, $V_{I\ Max} = -3.8\text{V}$				5.0	
$V_{OL\ LINE}$	Line Regulation	$T_J = 25^\circ\text{C}$, $V_O \geq -10\text{V}$ $(V_O - 20\text{V}) \leq V_I \leq (V_O - 2.5\text{V})$				1.0	% V_O
$V_{OL\ LOAD}$	Load Regulation	$T_J = 25^\circ\text{C}$,	$250\text{ mA} \leq I_O \leq 750\text{ mA}$			1.0	% V_O
		$V_I = V_O = -5.0\text{V}$	$5.0\text{ mA} \leq I_O \leq 1.5\text{A}$			2.0	
I_C	Control Lead Current	$T_J = 25^\circ\text{C}$		0.4	2.0		μA
						3.0	
I_Q	Quiescent Current	$T_J = 25^\circ\text{C}$		2.5	2.0		mA
						8.0	
$\Delta V_I / \Delta V_O$	Ripple Rejection	$V_O = -8.0\text{V}$, $V_I = -13\text{V}$, $f = 2400\text{ Hz}$, $I_C = 350\text{ mA}$		50	60		dB
N_O	Noise	$T_J = 25^\circ\text{C}$, $10\text{ Hz} \leq f \leq 100\text{ kHz}$, $V_O = -8.0\text{V}$, $I_O = 5.0\text{ mA}$			25	80	$\mu\text{V}/\text{V}_O$
V_{DO}	Dropout Voltage (Note 2)				1.1	2.3	V
I_{OS}	Output Short Circuit Current	$T_J = 25^\circ\text{C}$, $V_I = -30\text{V}$			0.25	1.2	A
I_{pk}	Peak Output Current	$T_J = 25^\circ\text{C}$		1.3	2.1	3.3	A
$\Delta V_O / \Delta T$	Average Temperature Coefficient of Output Voltage	$V_O = -5.0\text{V}$, $I_O = 5.0\text{ mA}$	$T_A = -55^\circ\text{C}$ to $+25^\circ\text{C}$			0.3	$\text{mV}/^\circ\text{C}/V_O$
			$T_A = 25^\circ\text{C}$ to 125°C			0.3	
V_C	Control Lead Voltage (Reference)	$T_J = 25^\circ\text{C}$		-2.65	-2.55	-2.45	V
				-2.68		-2.43	

Note 1: V_O is defined for the LM79G as $V_O = \frac{R_1 + R_2}{R_2} (-2.55)$.

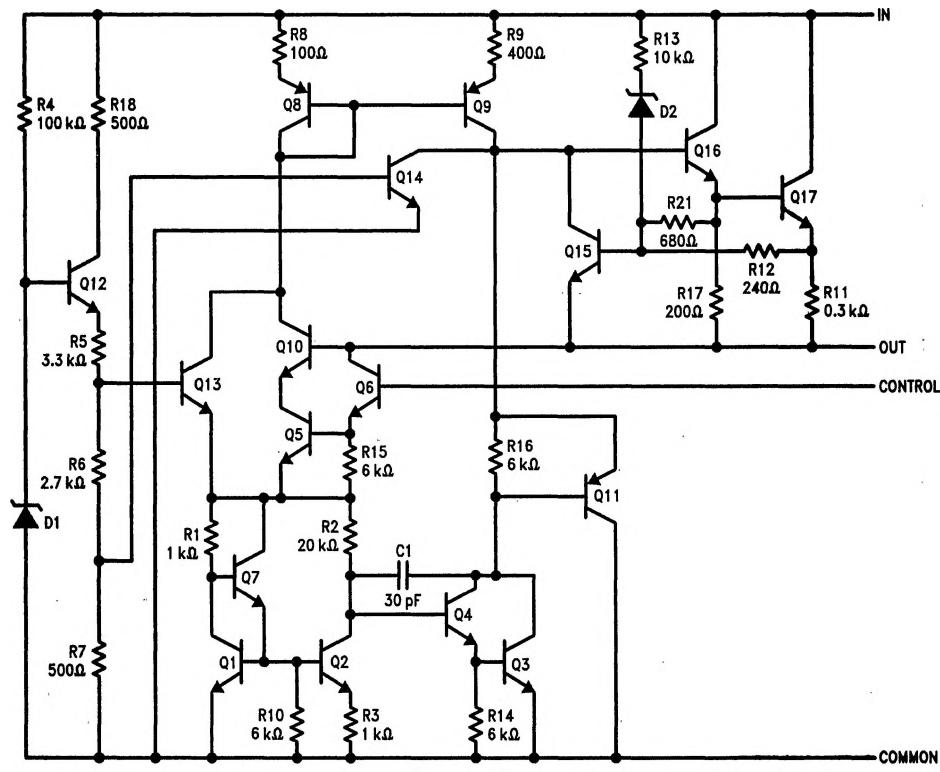
Note 2: Dropout Voltage is defined as that input/output voltage differential which causes the output voltage to decrease by 5% of its initial value.

Note 3: The convention for negative regulators is the algebraic value, thus -15V is less than -10V .

Note 4: All characteristics except noise voltage and ripple rejection ratio are measured using pulse techniques ($t_w \leq 10\text{ ms}$, duty cycle $\leq 5\%$). Output voltage changes due to changes in internal temperature must be taken into account separately.

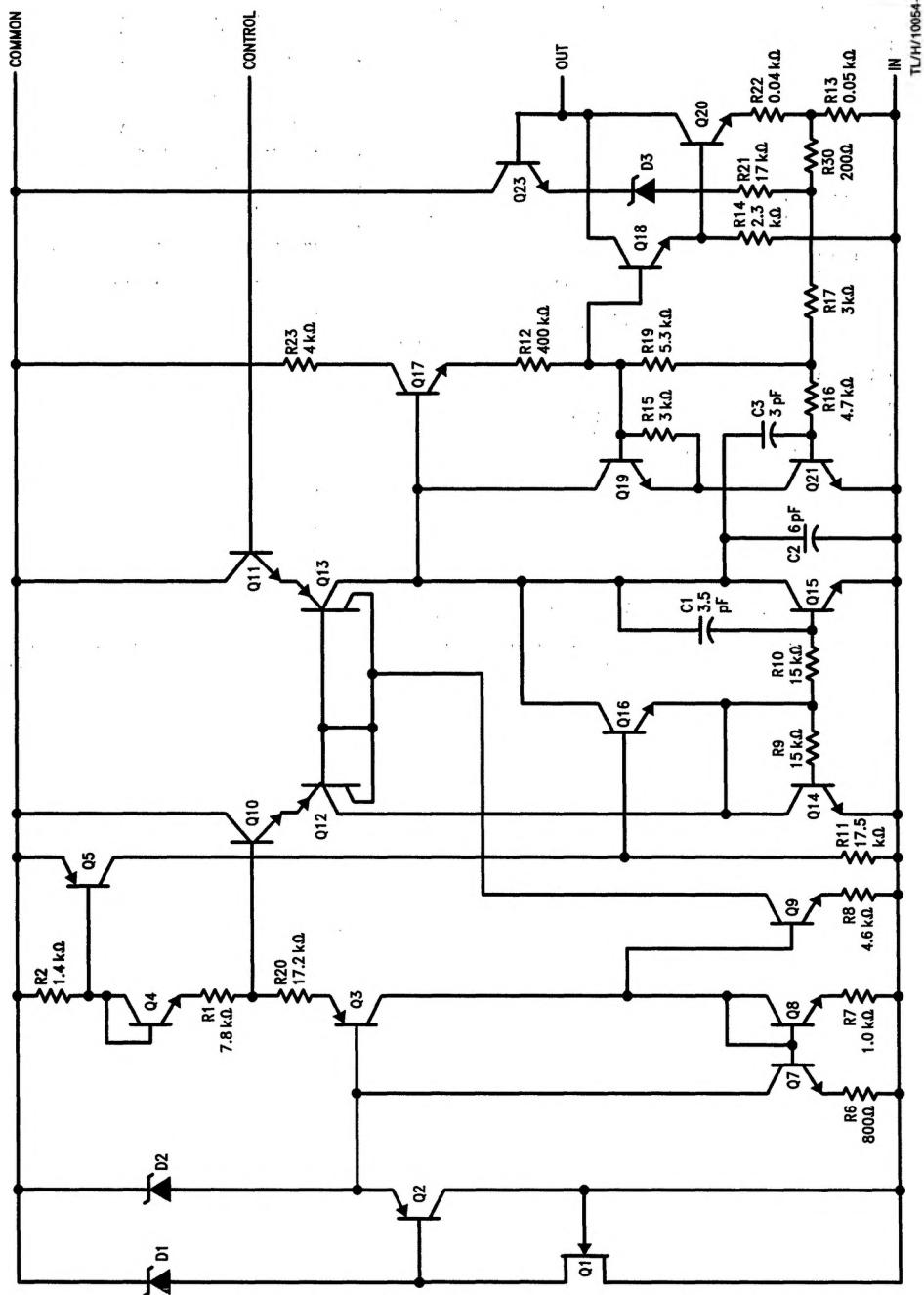
LM78G Equivalent Circuit

LM78G/LM79G

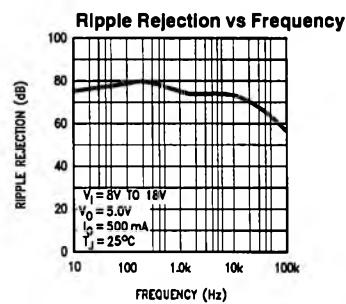
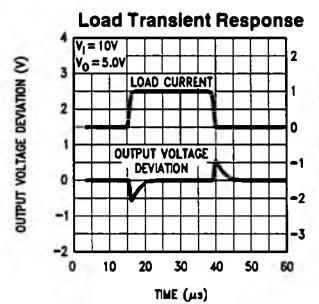
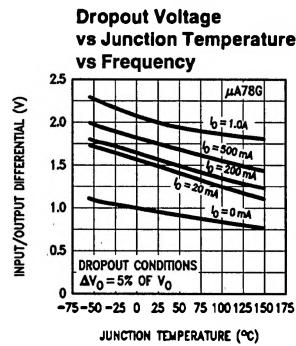
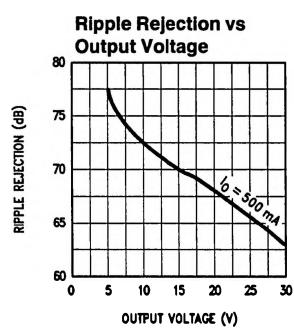
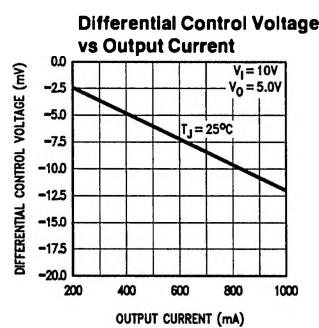
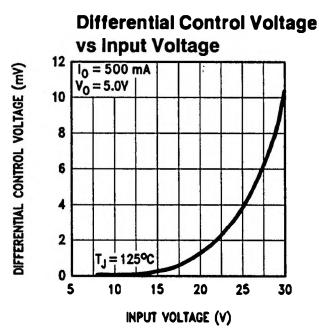
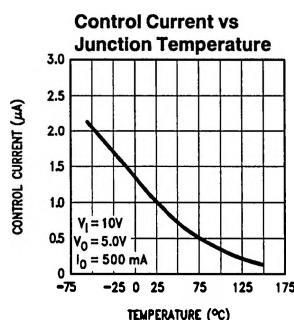
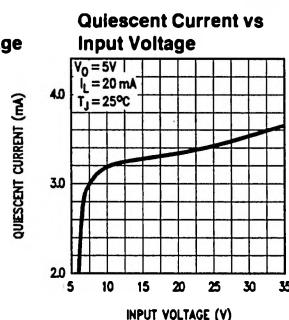
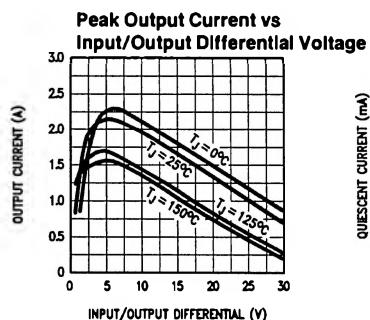


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LM79G Equivalent Circuit



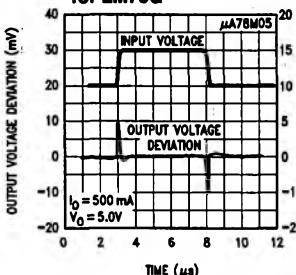
Typical Performance Curves for LM78G



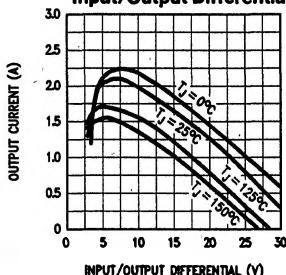
TL/H/10054-5

Typical Performance Curves for LM79G

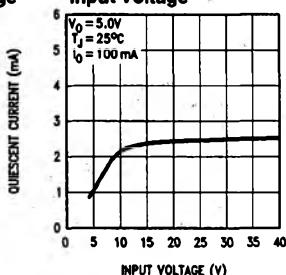
Line Transient Response for LM78G



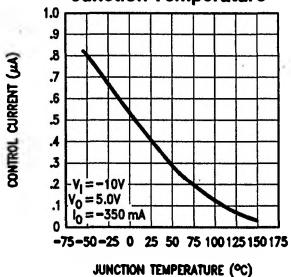
Peak Output Current vs Input/Output Differential Voltage



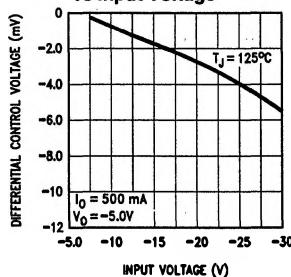
Quiescent Current vs Input Voltage



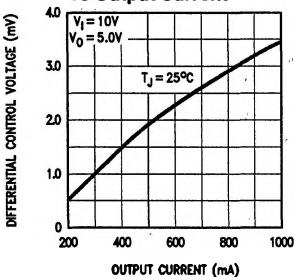
Control Current vs Junction Temperature



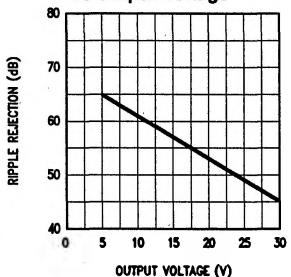
Differential Control Voltage vs Input Voltage



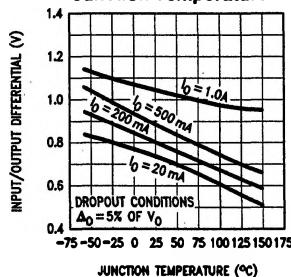
Differential Control Voltage vs Output Current



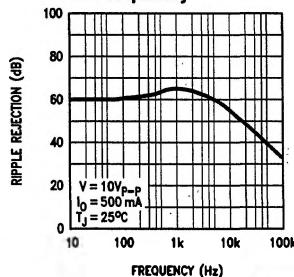
Ripple Rejection vs Output Voltage



Dropout Voltage vs Junction Temperature

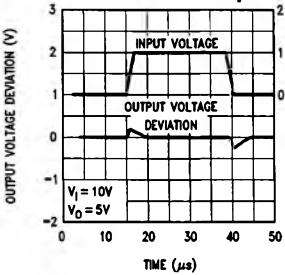


Ripple Rejection vs Frequency



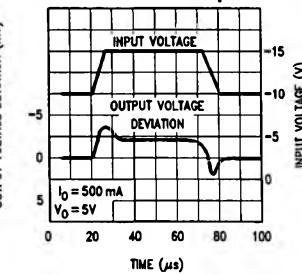
TL/H/10054-6

Load Transient Response



TL/H/10054-7

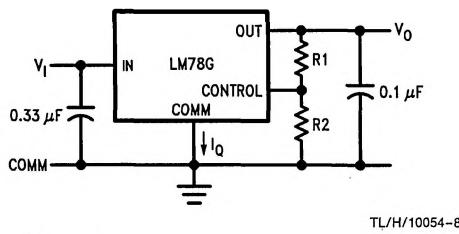
Line Transient Response



TL/H/10054-10

Test Circuits

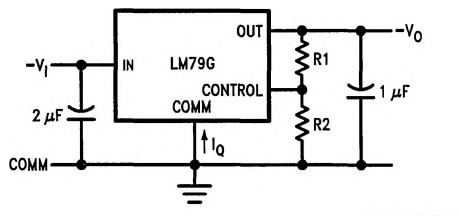
LM78G Test Circuit 1



$$V_O = \left(\frac{R_1 + R_2}{R_2} \right) V_{CONT}$$

V_{CONT} Nominal = 5.0V

LM79G Test Circuit 2



$$V_O = \left(\frac{R_1 + R_2}{R_2} \right) V_{CONT}$$

V_{CONT} Nominal = -2.55V

Recommended R2 current ≈ 1.0 mA

$\therefore R_2 = 5.0$ kΩ (LM78G)

$R_2 = 2.55$ kΩ (LM79G)

Design Considerations

The LM78G and LM79G Adjustable Voltage Regulators have an output voltage which varies from V_{CONT} to typically

$$V_I - 2.0V \text{ by } V_O = V_{CONT} \frac{R_1 + R_2}{R_2}$$

The nominal reference in the LM78G is 5.0V and LM79G is -2.55V. If we allow 1.0 mA to flow in the control string to eliminate bias current effects, we can make $R_2 = 5.0$ kΩ in the LM78G. Then, the output voltage is; $V_O = (R_1 + R_2)V$, where R_1 and R_2 are in kΩs.

Example: If $R_2 = 5.0$ kΩ and $R_1 = 10$ kΩ then

$$V_O = 15V \text{ nominal, for the LM78G}$$

$$R_2 = 2.55$$
 kΩ and $R_1 = 12.8$ kΩ then

$$V_O = -15.35 \text{ nominal, for the LM79G}$$

By proper wiring of the feedback resistors, load regulation of the device can be improved significantly.

Both LM78G and LM79G regulators have thermal overload protection from excessive power, internal short circuit protection which limits each circuit's maximum current, and out-

put transistor safe-area protection for reducing the output current as the voltage across each pass transistor is increased.

Although the internal power dissipation is limited, the junction temperature must be kept below the maximum specified temperature in order to meet data sheet specifications. To calculate the maximum junction temperature or heat sink required, the following thermal resistance values should be used:

	Typ °C/W	Max °C/W	Typ °C/W	Max °C/W
Package	θ_{JC}	θ_{JC}	θ_{JA}	θ_{JA}
Power Watt	7.5	11	75	80

$$P_D \text{ Max} = \frac{T_J \text{ Max} - T_A}{\theta_{JC} + \theta_{CA}} \text{ or}$$

$$= \frac{T_J \text{ Max} - T_A}{\theta_{JA}} \text{ (without a heat sink)}$$

$$\theta_{CA} = \theta_{CS} + \theta_{SA}$$

Solving for T_J :

$$T_J = T_A + P_D(\theta_{JC} + \theta_{CA}) \text{ or}$$

$$= T_A + P_D\theta_{JA} \text{ (without heat sink)}$$

Where:

$$T_J = \text{Junction Temperature}$$

$$T_A = \text{Ambient Temperature}$$

$$P_D = \text{Power Dissipation}$$

$$\theta_{JA} = \text{Junction to Ambient Thermal Resistance}$$

$$\theta_{JC} = \text{Junction to Case Thermal Resistance}$$

$$\theta_{CA} = \text{Case to Ambient Thermal Resistance}$$

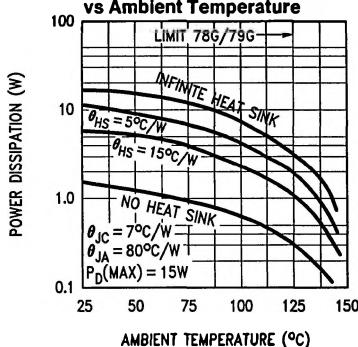
$$\theta_{CS} = \text{Case to Heat Sink Resistance}$$

$$\theta_{SA} = \text{Heat Sink to Ambient Thermal Resistance}$$

LM78G and LM79G

Power Tab (U1) Package

Worst Case Power Dissipation vs Ambient Temperature

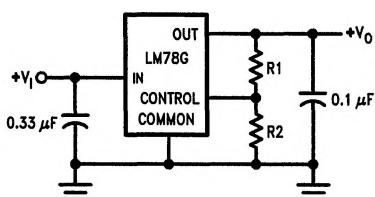


TL/H/10054-11

Typical Applications for LM78G

Bypassing of the input and output ($0.33 \mu\text{F}$ and $0.1 \mu\text{F}$, respectively) is necessary.

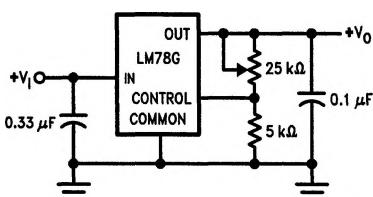
Basic Positive Regulator



TL/H/10054-12

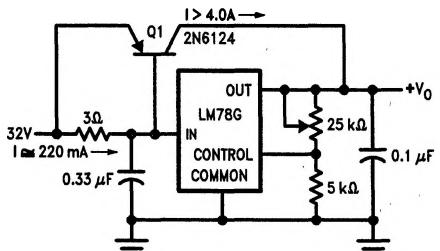
$$V_O = V_{CONT} \left(\frac{R_1 + R_2}{R_2} \right)$$

Positive 5.0V to 30V Adjustable Regulator



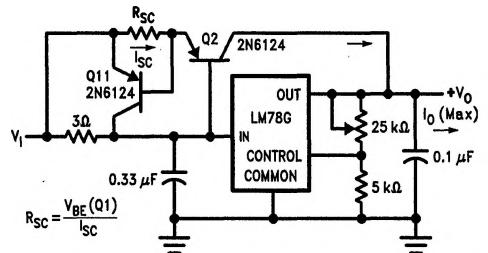
TL/H/10054-13

**Positive 5.0V to 30V
Adjustable Regulator
($I_O > 5.0A$) (Note 1)**



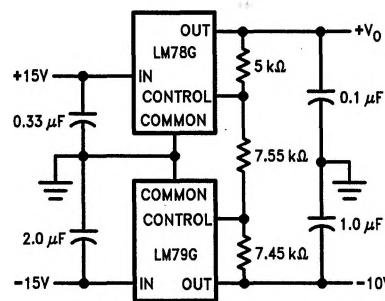
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Positive High Current Short Circuit, Protected Regulator



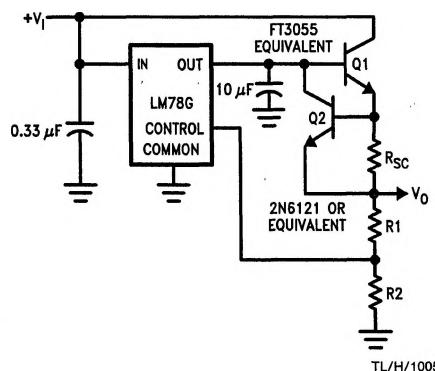
TL/H/10054-15

± 10V, 1.0A, Dual Tracking Regulator (Note 2)



TL / H/10054-16

Positive High Current, Short Circuit Protected Regulator

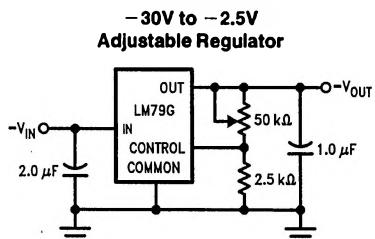


TL/H/10054-17

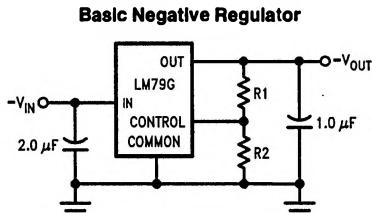
Note 2: If load is not ground referenced, connect reverse biased diodes from

Typical Applications for LM79G

All LM78G applications apply to the LM79G under the following conditions: R2 values are 2.5 kΩ, all external transistors and diodes reverse polarity.



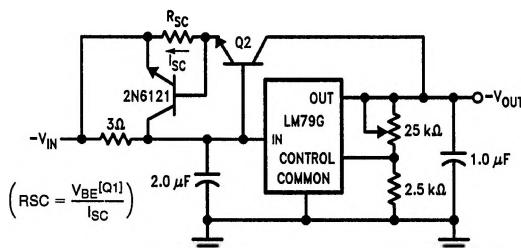
TL/H/10054-18



TL/H/10054-19

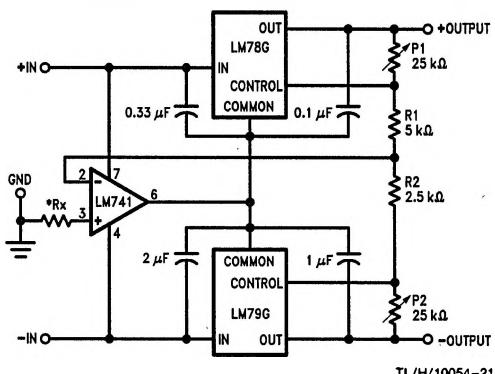
$$V_{OUT} = -V_{CONT} \left(\frac{R_1 + R_2}{R_2} \right)$$

Negative High Current Short Circuit Protected Regulator



TL/H/10054-20

Adjustable Dual Tracking Regulator



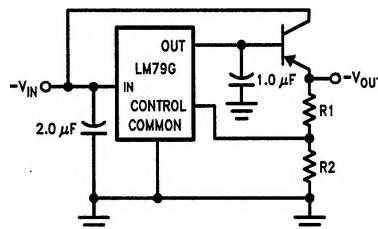
TL/H/10054-21

*Rx = Parallel combination of
(R1 + P1) and (R2 + P2).

Applications Hints

Bypass capacitors are recommended for stable operation of the LM79G series of regulators over the input voltage and output current ranges. Output bypass capacitors will improve the transient response of the regulator.

Negative High Current Voltage Regulator External Series Pass



TL/H/10054-22

The bypass capacitors (2 μF on the input, 1 μF on the output), should be ceramic or solid tantalum which have good high frequency characteristics. If aluminum electrolytics are used, their values should be 10 μF or larger. The bypass capacitors should be mounted with the shortest leads, and if possible, directly across the regulator terminals.