

# LMH6640 TFT-LCD Single, 16V Rail-to-Rail High Output Operational Amplifier

Check for Samples: LMH6640

### **FEATURES**

- (V<sub>S</sub> = 16V, R<sub>L</sub>= 2 kΩ to V\*/2, 25°C, Typical Values Unless Specified)
- Supply current (no load) 4 mA
- Output resistance (closed loop 1 MHz) 0.35Ω
- $-3 \text{ dB BW } (A_V = 1) 190 \text{ MHz}$
- Settling time (±0.1%, 2 V<sub>PP</sub>) 35 ns
- Input common mode voltage -0.3V to 15.1V
- Output voltage swing 100 mV from rails
- Linear output current ±100 mA
- Total harmonic distortion (2 V<sub>PP</sub>, 5 MHz) -64 dBc

- Fully characterized for: 5V & 16V
- No output phase reversal with CMVR exceeded
- Differential gain ( $R_L = 150\Omega$ ) 0.12%
- Differential phase  $(R_1 = 150\Omega) 0.12^\circ$

### **APPLICATIONS**

- TFT panel V<sub>COM</sub> buffer amplifier
- Active filters
- CD/DVD ROM
- ADC buffer amplifier
- Portable video
- Current sense buffer

### **DESCRIPTION**

The LMH™6640 is a voltage feedback operational amplifier with a rail-to-rail output drive capability of 100 mA. Employing National's patented VIP10 process, the LMH6640 delivers a bandwidth of 190 MHz at a current consumption of only 4mA. An input common mode voltage range extending to 0.3V below the V− and to within 0.9V of V⁺, makes the LMH6640 a true single supply op-amp. The output voltage range extends to within 100 mV of either supply rail providing the user with a dynamic range that is especially desirable in low voltage applications.

The LMH6640 offers a slew rate of 170 V/ $\mu$ s resulting in a full power bandwidth of approximately 28 MHz with 5V single supply (2 V $_{PP}$ , -1 dB). Careful attention has been paid to ensure device stability under all operating voltages and modes. The result is a very well behaved frequency response characteristic for any gain setting including +1, and excellent specifications for driving video cables including total harmonic distortion of -64 dBc @ 5 MHz, differential gain of 0.12% and differential phase of 0.12°.

### **Typical Application**

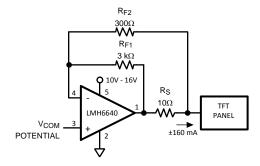


Figure 1. Typical Application as a TFT Panel V<sub>COM</sub> Driver



These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

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### Absolute Maximum Ratings (1)

ESD Tolerance (2)	
Human Body Model	2 KV
Machine Model	200V
V <sub>IN</sub> Differential	±2.5V
Input Current	±10 mA
'	18V
Supply Voltages (V <sup>+</sup> – V <sup>-</sup> )	
Voltage at Input/Output Pins	V <sup>+</sup> +0.8V, V <sup>-</sup> -0.8V
Storage Temperature Range	-65°C to +150°C
Junction Temperature (3)	+150°C
Soldering Information	
Infrared or Convection (20 sec.)	235°C
Wave Soldering (10 sec.)	260°C

- (1) Absolute maximum Ratings indicate limits beyond which damage to the device may occur. Operating Ratings indicate conditions for which the device is intended to be functional, but specific performance is not guaranteed. For guaranteed specifications and the test conditions, see the Electrical Characteristics.
- Human body model, 1.5 k $\Omega$  in series with 100 pF. Machine Model, 0 $\Omega$  in series with 200 pF.
- The maximum power dissipation is a function of  $T_{J(MAX)}$ ,  $\theta_{JA}$ , and  $T_A$ . The maximum allowable power dissipation at any ambient temperature is  $P_D = (T_{J(MAX)}, T_A) / \theta_{JA}$ . All numbers apply for packages soldered directly onto a PC board.

### Operating Ratings (1)

Supply Voltage (V <sup>+</sup> – V <sup>-</sup> )	4.5V to 16V
Operating Temperature Range (2)	−40°C to +85°C
Package Thermal Resistance (2)	
5-Pin SOT23	265°C/W

(1) Applies to both single-supply and split-supply operation. Continuous short circuit operation at elevated ambient temperature can result in exceeding the maximum allowed junction temperature of 150 °C Short circuit test is a momentary test. Output short circuit duration is infinite for  $V_S < 6V$  at room temperature and below. For  $V_S > 6V$ , allowable short circuit duration is 1.5 ms. The maximum power dissipation is a function of  $T_{J(MAX)}$ ,  $\theta_{JA}$ , and  $T_A$ . The maximum allowable power dissipation at any ambient temperature is  $P_D = (T_{J(MAX)} - T_A) / \theta_{JA}$ . All numbers apply for packages soldered directly onto a PC board.



### **5V Electrical Characteristics**

Unless otherwise specified, All limits guaranteed for  $T_J = 25$ °C,  $V^+ = 5V$ ,  $V^- = 0V$ ,  $V_O = V_{CM} = V^+/2$  and  $R_L = 2 \text{ k}\Omega$  to  $V^+/2$ . **Boldface** limits apply at temperature extremes. (1)

Symbol	Parameter	Conditions	Min (2)	<b>Тур</b> (3)	Max (2)	Units	
BW	-3 dB Bandwidth	$A_V = +1 \ (R_L = 100\Omega)$			150		
		$A_V = -1 \ (R_L = 100\Omega)$			58		MHz
BW <sub>0.1 dB</sub>	0.1 dB Gain Flatness	A <sub>V</sub> = -3			18		MHz
FPBW	Full Power Bandwidth	A <sub>V</sub> = +1, V <sub>OUT</sub> = 2 V <sub>PP</sub> , −1 dB			28		MHz
LSBW	-3 dB Bandwidth	$A_V = +1$ , $V_O = 2 V_{PP}$ ( $R_L = 1000$	2)		32		MHz
GBW	Gain Bandwidth Product	$A_V = +1$ , $(R_L = 100\Omega)$			59		MHz
SR	Slew Rate (4)	A <sub>V</sub> = -1			170		V/µs
e <sub>n</sub>	Input Referred Voltage Noise		f = 10 kHz		23		
			f = 1 MHz		15		nV/√Hz
i <sub>n</sub>	Input Referred Current Noise		f = 10 kHz		1.1		A / /II
			f = 1 MHz		0.7		pA/√Hz
THD	Total Harmonic Distortion	$f = 5 \text{ MHz}, V_O = 2 V_{PP}, A_V = +2 R_L = 1 k\Omega \text{ to } V^+/2$	!		-65		dBc
t <sub>s</sub>	Settling Time	$V_O = 2 V_{PP}, \pm 0.1\%, A_V = -1$			35		ns
V <sub>OS</sub>	Input Offset Voltage				1	5 <b>7</b>	mV
I <sub>B</sub>	Input Bias Current (5)				-1.2	-2.6 <b>-3.25</b>	μΑ
I <sub>OS</sub>	Input Offset Current				34	800 <b>1400</b>	nA
CMVR Common Mode Input Voltage Range		CMRR ≥ 50 dB		-0.3	-0.2 <b>-0.1</b>	V	
				4.0 <b>3.6</b>	4.1		V
CMRR	Common Mode Rejection Ratio	$V^{-} \le V_{CM} \le V^{+} -1.5V$		72	90		dB
A <sub>VOL</sub>	Large Signal Voltage Gain	$V_O = 4 V_{PP}$ , $R_L = 2 k\Omega$ to $V^+/2$		86 <b>82</b>	95		4D
		$V_{O} = 3.75 V_{PP}, R_{L} = 150\Omega \text{ to } V_{C}$	+/2	74 <b>70</b>	78		dB
Vo	Output Swing High	$R_L = 2 k\Omega$ to $V^+/2$		4.90	4.94		
		$R_L = 150\Omega \text{ to } V^{+}/2$		4.75	4.80		.,
	Output Swing Low	$R_L = 2 k\Omega$ to $V^+/2$			0.06	0.10	V
		$R_L = 150\Omega \text{ to } V^+/2$			0.20	0.25	
I <sub>SC</sub>	Output Short Circuit Current (6)	Sourcing to V <sup>+</sup> /2	100 <b>75</b>	130			
		Sinking from V <sup>+</sup> /2	100 <b>70</b>	130		mA	
I <sub>OUT</sub>	Output Current	V <sub>O</sub> = 0.5V from either Supply		+75/-90		mA	
PSRR	Power Supply Rejection Ratio	4V ≤ V <sup>+</sup> ≤ 6V		72	80		dB
Is	Supply Current	No Load			3.7	5.5 <b>8.0</b>	mA

<sup>(1)</sup> Electrical Table values apply only for factory testing conditions at the temperature indicated. Factory testing conditions result in very limited self-heating of the device such that T<sub>J</sub> = T<sub>A</sub>. No guarantee of parametric performance is indicated in the electrical tables under conditions of internal self-heating where T<sub>J</sub> > T<sub>A</sub>.

<sup>(2)</sup> All limits are guaranteed by testing or statistical analysis.

<sup>(3)</sup> Typical Values represent the most likely parametric norm.

<sup>(4)</sup> Slew rate is the average of the rising and falling slew rates

<sup>(5)</sup> Positive current corresponds to current flowing into the device.

<sup>(6)</sup> Applies to both single-supply and split-supply operation. Continuous short circuit operation at elevated ambient temperature can result in exceeding the maximum allowed junction temperature of 150 °C Short circuit test is a momentary test. Output short circuit duration is infinite for V<sub>S</sub> < 6V at room temperature and below. For V<sub>S</sub> > 6V, allowable short circuit duration is 1.5 ms.



## **5V Electrical Characteristics (continued)**

Unless otherwise specified, All limits guaranteed for  $T_J = 25^{\circ}C$ ,  $V^+ = 5V$ ,  $V^- = 0V$ ,  $V_O = V_{CM} = V^+/2$  and  $R_L = 2 \text{ k}\Omega$  to  $V^+/2$ . **Boldface** limits apply at temperature extremes. (1)

Symbol	Parameter	Conditions	Min (2)	Typ (3)	Max (2)	Units
R <sub>IN</sub>	Common Mode Input Resistance	$A_V = +1$ , $f = 1$ kHz, $R_S = 1$ M $\Omega$		15		ΜΩ
C <sub>IN</sub>	Common Mode Input Capacitance	$A_V = +1, R_S = 100 \text{ k}\Omega$		1.7		pF
R <sub>OUT</sub>	Output Resistance Closed Loop	$R_F = 10 \text{ k}\Omega, f = 1 \text{ kHz}, A_V = -1$		0.1		Ω
		$R_F = 10 \text{ k}\Omega$ , $f = 1 \text{ MHz}$ , $A_V = -1$		0.4		Ω
DG	Differential Gain	NTSC, $A_V = +2$ $R_L = 150\Omega$ to $V^+/2$		0.13		%
DP	Differential Phase	NTSC, $A_V = +2$ $R_L = 150\Omega$ to $V^+/2$		0.10		deg

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#### 16V Electrical Characteristics

Unless otherwise specified, All limits guaranteed for  $T_J = 25^{\circ}C$ ,  $V^+ = 16V$ ,  $V^- = 0V$ ,  $V_O = V_{CM} = V^+/2$  and  $R_L = 2 \text{ k}\Omega$  to  $V^+/2$ . **Boldface** limits apply at temperature extremes. (1)

Symbol	Parameter	Conditions		Min (2)	Тур (3)	Max (2)	Units
BW	-3 dB Bandwidth	$A_V = +1 \ (R_L = 100\Omega)$		190			
		$A_V = -1 \ (R_L = 100\Omega)$			60		MHz
BW <sub>0.1 dB</sub>	0.1 dB Gain Flatness	A <sub>V</sub> = −2.7			20		MHz
LSBW	-3 dB Bandwidth	$A_V = +1, V_O = 2 V_{PP} (R_L = 1000)$	2)		35		MHz
GBW	Gain Bandwidth Product	$A_V = +1$ , $(R_L = 100\Omega)$			62		MHz
SR	Slew Rate (4)	A <sub>V</sub> = −1			170		V/µs
e <sub>n</sub>	Input Referred Voltage Noise		f = 10 kHz		23		\ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \ \
			f = 1 MHz		15		nV/√Hz
i <sub>n</sub>	Input Referred Current Noise		f = 10 kHz		1.1		• (/!!
			f = 1 MHz		0.7		pA/√Hz
THD	Total Harmonic Distortion	f = 5 MHz, $V_O$ = 2 $V_{PP}$ , $A_V$ = +2 $R_L$ = 1 kΩ to $V^+/2$			-64		dBc
t <sub>s</sub>	Settling Time	$V_O = 2 V_{PP}, \pm 0.1\%, A_V = -1$			35		ns
V <sub>OS</sub>	Input Offset Voltage				1	5 <b>7</b>	mV
I <sub>B</sub>	Input Bias Current (5)				-1	-2.6 <b>-3.5</b>	μA
I <sub>OS</sub>	Input Offset Current				34	800 <b>1800</b>	nA
	Common Mode Input Voltage Range	CMRR ≥ 50 dB			-0.3	-0.2 - <b>0.1</b>	V
			15.0 <b>14.6</b>	15.1		V	
CMRR	Common Mode Rejection Ratio	$V^- \le V_{CM} \le V^+ -1.5V$		72	90		dB
A <sub>VOL</sub>	Large Signal Voltage Gain	$V_{O} = 15 V_{PP}, R_{L} = 2 k\Omega \text{ to } V^{+}/2$		86 <b>82</b>	95		dB
		$V_O = 14 V_{PP}, R_L = 150\Omega \text{ to } V^+/2$	74 <b>70</b>	78		ив	
Vo	Output Swing High	$R_L = 2 k\Omega$ to $V^+/2$		15.85	15.90		
		$R_L = 150\Omega$ to $V^+/2$		15.45	15.78		V
	Output Swing Low	$R_L = 2 k\Omega$ to $V^+/2$			0.10	0.15	V
		$R_L = 150\Omega$ to $V^+/2$			0.21	0.55	
I <sub>SC</sub>	Output Short Circuit Current (6)	Sourcing to V <sup>+</sup> /2		60 <b>30</b>	95		A
		Sinking from V <sup>+</sup> /2	50 <b>15</b>	75		− mA	
I <sub>OUT</sub>	Output Current	V <sub>O</sub> = 0.5V from either Supply			±100		mA
PSRR	Power Supply Rejection Ratio	15V ≤ V <sup>+</sup> ≤ 17V		72	80		dB
Is	Supply Current	No Load			4	6.5 <b>7.8</b>	mA
R <sub>IN</sub>	Common Mode Input Resistance	$A_V = +1$ , $f = 1$ kHz, $R_S = 1$ M $\Omega$			32		МΩ

<sup>(1)</sup> Electrical Table values apply only for factory testing conditions at the temperature indicated. Factory testing conditions result in very limited self-heating of the device such that T<sub>J</sub> = T<sub>A</sub>. No guarantee of parametric performance is indicated in the electrical tables under conditions of internal self-heating where T<sub>J</sub> > T<sub>A</sub>.

<sup>(2)</sup> All limits are guaranteed by testing or statistical analysis.

<sup>(3)</sup> Typical Values represent the most likely parametric norm.

<sup>(4)</sup> Slew rate is the average of the rising and falling slew rates

<sup>(5)</sup> Positive current corresponds to current flowing into the device.

<sup>(6)</sup> Applies to both single-supply and split-supply operation. Continuous short circuit operation at elevated ambient temperature can result in exceeding the maximum allowed junction temperature of 150 °C Short circuit test is a momentary test. Output short circuit duration is infinite for V<sub>S</sub> < 6V at room temperature and below. For V<sub>S</sub> > 6V, allowable short circuit duration is 1.5 ms.



## **16V Electrical Characteristics (continued)**

Unless otherwise specified, All limits guaranteed for  $T_J = 25^{\circ}C$ ,  $V^+ = 16V$ ,  $V^- = 0V$ ,  $V_O = V_{CM} = V^+/2$  and  $R_L = 2 \text{ k}\Omega$  to  $V^+/2$ . **Boldface** limits apply at temperature extremes.

Symbol	Parameter	Conditions	Min (2)	<b>Тур</b> (3)	Max (2)	Units
C <sub>IN</sub>	Common Mode Input Capacitance	$A_V = +1, R_S = 100 \text{ k}\Omega$		1.7		pF
R <sub>OUT</sub>	Output Resistance Closed Loop	$R_F = 10 \text{ k}\Omega, f = 1 \text{ kHz}, A_V = -1$		0.1		0
		$R_F = 10 \text{ k}\Omega, f = 1 \text{ MHz}, A_V = -1$		0.3		Ω
DG	Differential Gain	NTSC, $A_V = +2$ $R_L = 150\Omega$ to V <sup>+</sup> /2		0.12		%
DP	Differential Phase	NTSC, $A_V = +2$ $R_L = 150\Omega$ to $V^+/2$		0.12		deg

## **Connection Diagram**

### 5-Pin SOT23

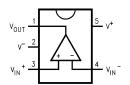
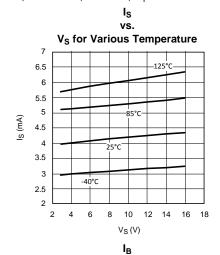


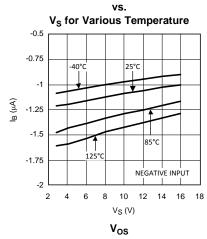
Figure 2. Top View

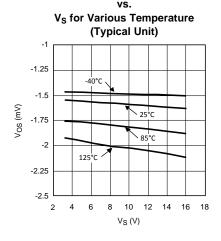


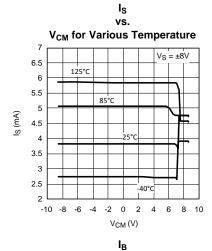
## **Typical Performance Characteristics**

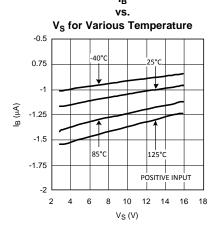
At  $T_J = 25^{\circ}C$ ,  $V^+ = 16$  V,  $V^- = 0$ V,  $R_F = 330\Omega$  for  $A_V = +2$ ,  $R_F = 1$  k $\Omega$  for  $A_V = -1$ .  $R_L$  tied to  $V^+/2$ . Unless otherwise specified.

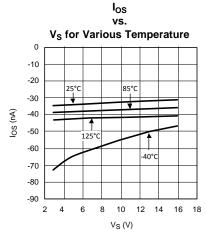








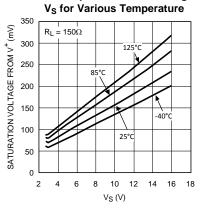




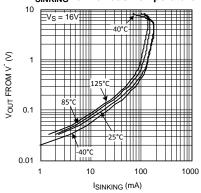


At  $T_J = 25^{\circ}C$ ,  $V^+ = 16$  V,  $V^- = 0$ V,  $R_F = 330\Omega$  for  $A_V = +2$ ,  $R_F = 1$  k $\Omega$  for  $A_V = -1$ .  $R_L$  tied to  $V^+/2$ . Unless otherwise specified.

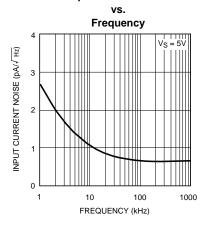
# Positive Output Saturation Voltage vs.



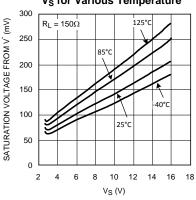
# Output Sinking Saturation Voltage vs. I<sub>SINKING</sub> for Various Temperature



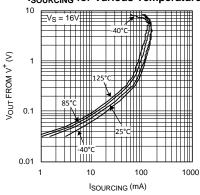
**Input Current Noise** 



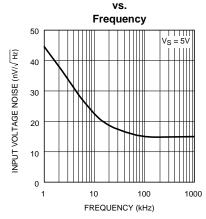
# $\label{eq:VS} \mbox{Negative Output Saturation Voltage vs.} \\ \mbox{V}_{\mbox{S}} \mbox{ for Various Temperature}$



# Output Sourcing Saturation Voltage vs. I<sub>SOURCING</sub> for Various Temperature



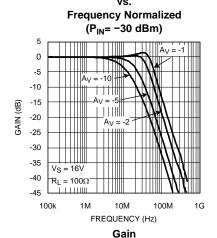
Input Voltage Noise



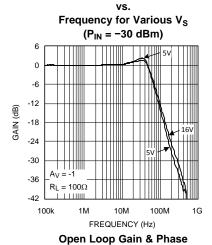
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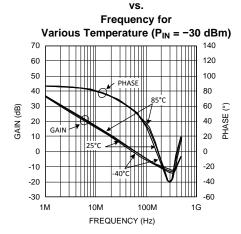


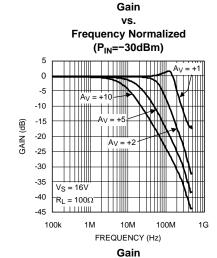
At  $T_J = 25^{\circ}C$ ,  $V^+ = 16$  V,  $V^- = 0$ V,  $R_F = 330\Omega$  for  $A_V = +2$ ,  $R_F = 1$  k $\Omega$  for  $A_V = -1$ .  $R_L$  tied to  $V^+/2$ . Unless otherwise specified.

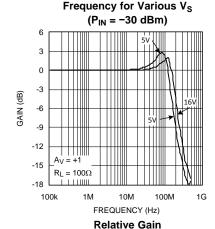


Gain

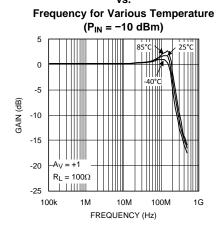








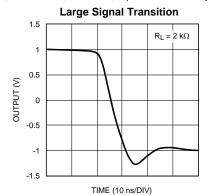
vs.

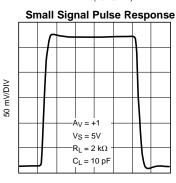


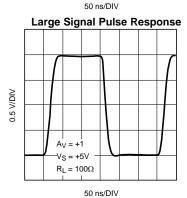
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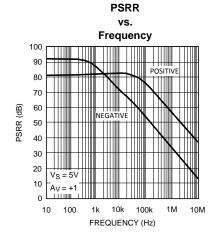


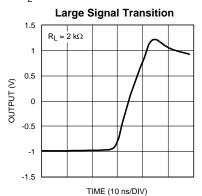
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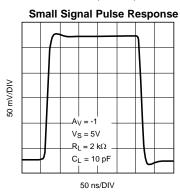


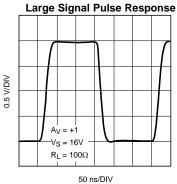


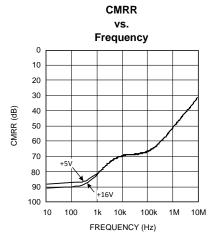










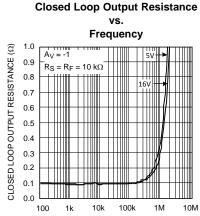


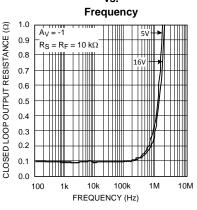
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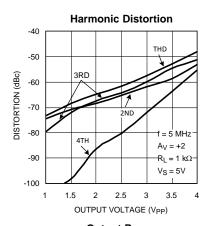
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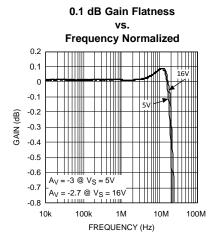


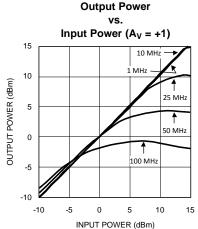
At  $T_J = 25^{\circ}$ C,  $V^+ = 16$  V,  $V^- = 0$ V,  $R_F = 330\Omega$  for  $A_V = +2$ ,  $R_F = 1$  k $\Omega$  for  $A_V = -1$ .  $R_L$  tied to  $V^+/2$ . Unless otherwise specified.



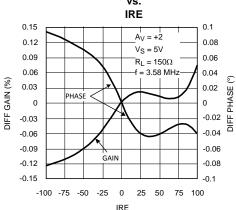












## **Application Notes**

With its high output current and speed, one of the major applications for the LMH6640 is the V<sub>COM</sub> driver in a TFT panel. This application is a specially taxing one because of the demands it places on the operational amplifier's output to drive a large amount of bi-directional current into a heavy capacitive load while operating under unity gain condition, which is a difficult challenge due to loop stability reasons. For a more detailed explanation of what a TFT panel is and what its amplifier requirements are, please see the Application Notes section of the LM6584 found on the web at: http://www.national.com/ds.cgi/LM/LM6584.pdf



Because of the complexity of the TFT  $V_{COM}$  waveform and the wide variation in characteristics between different TFT panels, it is difficult to decipher the results of circuit testing in an actual panel. The ability to make simplifying assumptions about the load in order to test the amplifier on the bench allows testing using standard equipment and provides familiar results which could be interpreted using standard loop analysis techniques. This is what has been done in this application note with regard to the LMH6640's performance when subjected to the conditions found in a TFT  $V_{COM}$  application.

Figure 3, shows a typical simplified  $V_{COM}$  application with the LMH6640 buffering the  $V_{COM}$  potential (which is usually around ½ of panel supply voltage) and looking into the simplified model of the load. The load represents the cumulative effect of all stray capacitances between the  $V_{COM}$  node and both row and column lines. Associated with the capacitances shown, is the distributed resistance of the lines to each individual transistor switch. The other end of this R-C ladder is driven by the column driver in an actual panel and here is driven with a low impedance MOSFET driver (labeled "High Current Driver") for the purposes of this bench test to simulate the effect that the column driver exerts on the  $V_{COM}$  load.

The modeled TFT  $V_{COM}$  load, shown in Figure 3, is based on the following simplifying assumptions in order to allow for easy bench testing and yet allow good matching results obtained in the actual application:

- The sum of all the capacitors and resistors in the R-C ladder is the total  $V_{COM}$  capacitance and resistance respectively. This total varies from panel to panel; capacitance could range from 50 nF-200 nF and the resistance could be anywhere from  $20\Omega-100\Omega$ .
- The number of ladder sections has been reduced to a number (4 sections in this case) which can easily be
  put together in the lab and which behaves reasonably close to the actual load.

In this example, the LMH6640 was tested under the simulated conditions of total 209 nF capacitance and  $54\Omega$  as shown in Figure 3.

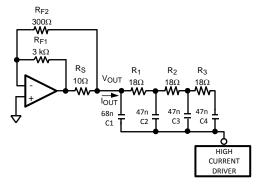


Figure 3. LMH6640 in a V<sub>COM</sub> Buffer Application with Simulated TFT Load

 $R_S$  is sometimes used in the panel to provide additional isolation from the load while  $R_{F2}$  provides a more direct feedback from the  $V_{COM}$ .  $R_{F1}$ ,  $R_{F2}$ , and  $R_S$  are trimmed in the actual circuit with settling time and stability tradeoffs considered and evaluated. When tested under simulated load conditions of Figure 3, here are the resultant voltage and current waveforms at the LMH6640 output:

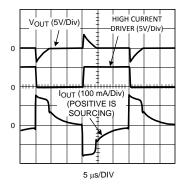


Figure 4. V<sub>COM</sub> Output, High Current Drive Waveform, & LMH6640 Output Current Waveforms



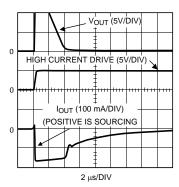


Figure 5. Expanded View of Figure 4 Waveforms showing LMH6640 Current Sinking ½ Cycle

As can be seen, the LMH6640 is capable of supplying up to 160 mA of output current and can settle the output in  $4.4 \, \mu s$ .

The LMH6640 is a cost effective amplifier for use in the TFT  $V_{COM}$  application and is made even more attractive by its large supply voltage range and high output current. The combination of all these features is not readily available in the market, especially in the space saving SOT23-5 package. All this performance is achieved at the low power consumption of 65 mW which is of utmost importance in today's battery driven TFT panels.





www.ti.com 24-Jan-2013

### PACKAGING INFORMATION

Orderable Device	Status	Package Type	U		Package Qty	Eco Plan	Lead/Ball Finish	MSL Peak Temp	Op Temp (°C)	Top-Side Markings	Samples
	(1)		Drawing			(2)		(3)		(4)	
LMH6640MF	ACTIVE	SOT-23	DBV	5	1000	TBD	CU SNPB	Level-1-260C-UNLIM	-40 to 85	AH1A	Samples
LMH6640MF/NOPB	ACTIVE	SOT-23	DBV	5	1000	Green (RoHS & no Sb/Br)	CU SN	Level-1-260C-UNLIM	-40 to 85	AH1A	Samples
LMH6640MFX	ACTIVE	SOT-23	DBV	5	3000	TBD	CU SNPB	Level-1-260C-UNLIM	-40 to 85	AH1A	Samples
LMH6640MFX/NOPB	ACTIVE	SOT-23	DBV	5	3000	Green (RoHS & no Sb/Br)	CU SN	Level-1-260C-UNLIM	-40 to 85	AH1A	Samples

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.

TBD: The Pb-Free/Green conversion plan has not been defined.

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<sup>(3)</sup> MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

<sup>&</sup>lt;sup>(4)</sup> Only one of markings shown within the brackets will appear on the physical device.

## PACKAGE MATERIALS INFORMATION

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## TAPE AND REEL INFORMATION





	Dimension designed to accommodate the component width
	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



#### \*All dimensions are nominal

All differsions are nominal												
Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
LMH6640MF	SOT-23	DBV	5	1000	178.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3
LMH6640MF/NOPB	SOT-23	DBV	5	1000	178.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3
LMH6640MFX	SOT-23	DBV	5	3000	178.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3
LMH6640MFX/NOPB	SOT-23	DBV	5	3000	178.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3

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\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
LMH6640MF	SOT-23	DBV	5	1000	203.0	190.0	41.0
LMH6640MF/NOPB	SOT-23	DBV	5	1000	203.0	190.0	41.0
LMH6640MFX	SOT-23	DBV	5	3000	206.0	191.0	90.0
LMH6640MFX/NOPB	SOT-23	DBV	5	3000	206.0	191.0	90.0

# DBV (R-PDSO-G5)

# PLASTIC SMALL-OUTLINE PACKAGE



NOTES:

- A. All linear dimensions are in millimeters.
  - B. This drawing is subject to change without notice.
  - C. Body dimensions do not include mold flash or protrusion. Mold flash and protrusion shall not exceed 0.15 per side.
  - D. Falls within JEDEC MO-178 Variation AA.



# DBV (R-PDSO-G5)

# PLASTIC SMALL OUTLINE



NOTES:

- A. All linear dimensions are in millimeters.
- B. This drawing is subject to change without notice.
- C. Customers should place a note on the circuit board fabrication drawing not to alter the center solder mask defined pad.
- D. Publication IPC-7351 is recommended for alternate designs.
- E. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Example stencil design based on a 50% volumetric metal load solder paste. Refer to IPC-7525 for other stencil recommendations.



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