INTEGRATED CIRCUITS



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FEATURES

- Double window PIP in interlaced mode at 8-bit resolution
- Internal DRAM of 1 Mbit
- Three 8-bit Analog-to-Digital Converters (ADCs) (7-bit performance) with clamp circuit for each acquisition channel
- One PLL which generates the line-locked clocks for the subchannel
- One PLL which generates the line-locked clocks for the main and display channel
- Three 8-bit Digital-to-Analog Converters (DACs)
- Linear zoom in both horizontal and vertical direction for the subchannel
- Linear zoom in horizontal direction for the main channel
- Three multi PIP modes are available.

GENERAL DESCRIPTION

The SAB9083 is an multistandard Picture-in-Picture controller which can be used in double window applications.

It inserts one or two live video signals with reduced size into another live video signal. The incoming video signals are expected to be analog baseband signals.



The conversion into the digital environment is done on chip with ADCs. Processing and storage of the video data is done entirely in the digital domain. The conversion back to the analog domain is done by means of DACs. Internal clocks are generated by PLLs which lock on to the applied horizontal and vertical syncs.

The main input channel is compressed horizontally with a factor of 2 and directly fed to the output. After compressing a horizontal expansion of 2 is possible for the main channel.

The subchannel is also compressed horizontally with a factor of 2 but stored in memory before it is fed to the outputs.

The SAB9083 can also create three multi PIP modes, one with three PIPs placed in a column (MP3) and two with two columns of three PIPs (MP6, MP6S). The reduction factors of these PIPs is horizontal 1/2 and vertical 1/3. In the first two modes the column(s) can be placed on the left or right side of the screen.

QUICK REFERENCE DATA

SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
Supply	1					
V _{DDD}	digital supply voltage		3.0	3.3	3.6	V
V _{DDA}	analog supply voltage		3.0	3.3	3.6	V
I _{DDD}	digital supply current		tbf	65	tbf	mA
I _{DDA}	analog supply current		tbf	185	tbf	mA
PLL	•		·	·	·	·
f _{sys}	system frequency	1792 × HSYNC	-	28	-	MHz
Bloop	loop bandwidth		-	4	-	kHz
t _{jitter}	short term stability	jitter during 64 μs	-	-	4	ns
ζ	damping factor		_	0.7	-	

ORDERING INFORMATION

TYPE NUMBER		PACKAGE	
	NAME	DESCRIPTION	VERSION
SAB9083	QFP100	plastic quad flat package; 100 leads (lead length 1.95 mm); body $14 \times 20 \times 2.8$ mm	SOT317-2

Philips Semiconductors

controller Multistandard Picture-In-Picture (PIP)

SAB9083





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Preliminary specification

PINNING

SYMBOL	PIN	I/O	DESCRIPTION
V _{ref(B)(MA)}	1	I/O	analog bottom reference voltage for main channel ADCs
MU	2	I	analog U input for main channel
V _{DDA(MF)}	3	S	analog supply voltage for main channel front-end buffers
V _{SSA(MA)}	4	S	analog ground for main channel ADCs
V _{DDA(MA)}	5	S	analog supply voltage for main channel ADCs
V _{DDA(DA)}	6	S	analog supply voltage for DACs
V _{SSA(DA)}	7	S	analog ground for DACs
DY	8	0	analog Y output of DAC
V _{bias(DA)}	9	I/O	input/output analog bias voltage reference for DACs
DV	10	0	analog V output of DAC
V _{ref(T)(DA)}	11	I/O	input/output analog top reference voltage for DACs
DU	12	0	analog U output of DAC
V _{ref(B)(DA)}	13	I/O	analog bottom reference voltage for DACs
V _{DDD(DA)}	14	S	digital supply voltage for DACs
V _{SSD(DA)}	15	S	digital ground for DACs
V _{SSD(P1)}	16	S	digital ground for periphery
V _{DDD(P1)}	17	S	digital supply voltage for periphery
V _{SSD(T1)}	18	_	digital ground for test
V _{SSD(T2)}	19	_	digital ground for test
V _{DDD(RP)}	20	S	digital supply voltage for memory periphery
n.c.	21 to 29	_	not connected
V _{SSD(T3)}	30	_	digital ground for test
n.c.	31	_	not connected
T5	32	I/O	test data input/output bit 5 (CMOS levels)
T4	33	I/O	test data input/output bit 4 (CMOS levels)
Т3	34	I/O	test data input/output bit 3 (CMOS levels)
T2	35	I/O	test data input/output bit 2 (CMOS levels)
T1	36	I/O	test data input/output bit 1 (CMOS levels)
Т0	37	I/O	test data input/output bit 0 (CMOS levels)
ТС	38	I	test control input (CMOS levels)
V _{DDD(RL)}	39	S	digital supply voltage for memory logic
V _{SSD(RL)}	40	S	digital ground for memory logic
V _{SSD(RM)}	41	S	digital ground for memory core
V _{DDD(RM)}	42	S	digital supply voltage for memory core
TCLK	43	I	test clock input (CMOS levels)
ТМ	44	Ι	test mode input (CMOS levels)
TCBD	45	I	test control block data input (CMOS levels)
ТСВС	46	I	test control block clock input (CMOS levels)
TCBR	47	I	test control block reset input (CMOS levels)
V _{SSD(T4-T7)}	48 to 51	_	digital ground for test

SYMBOL	PIN	I/O	DESCRIPTION
n.c.	52 to 60	_	not connected
V _{SSD(RP)}	61	S	digital ground for memory periphery
V _{SSD(T8,T9)}	62 and 63	_	digital ground for test
V _{DDD(P2)}	64	S	digital supply voltage for periphery
V _{SSD(P2)}	65	S	digital ground for periphery
V _{SSD(D)}	66	S	digital ground for digital core
V _{DDD(D)}	67	S	digital supply voltage for digital core
FBL	68	0	fast blanking control signal output (CMOS levels; +5 V tolerant)
PKOFF	69	0	peak off control signal output (CMOS levels; +5 V tolerant)
DVSYNC	70	I	vertical sync display channel input (CMOS levels; +5 V tolerant)
DCLK	71	I	test clock input (28 MHz) (CMOS levels)
SVSYNC	72	I	vertical sync for subchannel input (CMOS levels; +5 V tolerant)
SCL	73	I/O	input/output serial clock (I ² C-bus) (CMOS levels; +5 V tolerant)
SDA	74	I/O	input/output serial data/acknowledge output (I ² C-bus) (+5 V tolerant)
POR	75	I	power-on reset input (CMOS levels; pull-up resistor connected to V _{DD})
V _{DDA(SA)}	76	S	analog supply voltage for subchannel ADCs
V _{SSA(SA)}	77	S	analog ground for subchannel ADCs
V _{DDA(SF)}	78	S	analog supply voltage for subchannel front-end buffers and clamps
SU	79	Ι	analog U input for subchannel
V _{ref(B)(SA)}	80	I/O	input/output analog bottom reference voltage for subchannel ADCs
SV	81	I	analog V input for subchannel
V _{ref(T)(SA)}	82	I/O	input/output analog top reference voltage for subchannel ADCs
SY	83	I	analog Y input for subchannel
V _{bias(SA)}	84	I/O	analog bias reference voltage for subchannel ADCs
V _{SSD(SA)}	85	S	digital ground for subchannel ADCs
V _{DDD(SA)}	86	S	digital supply voltage for subchannel ADCs
SHSYNC	87	Ι	horizontal sync input for subchannel ($V_i < V_{SHSYNC}$)
Т6	88	I/O	test data input/output bit 7 (CMOS levels)
V _{DDA(SP)}	89	S	analog supply voltage for subchannel PLL
V _{SSA(SP)}	90	S	analog ground for subchannel PLL
V _{SSA(DP)}	91	S	analog ground for display channel PLL
V _{DDA(DP)}	92	S	analog supply voltage for display channel PLL
T7	93	I/O	test data input/output bit 6 (CMOS levels)
DHSYNC	94	Ι	horizontal sync display input for channel ($V_i < V_{DHSYNC}$)
V _{DDD(MA)}	95	S	digital supply voltage for main channel ADCs
V _{SSD(MA)}	96	S	digital ground for main channel ADCs
V _{bias(MA)}	97	I/O	analog bias reference voltage for main channel ADCs
MY	98		analog Y input for main channel
V _{ref(T)(MA)}	99	I/O	analog top reference voltage for main channel ADCs
MV	100	I	analog V input for main channel

SAB9083

Multistandard Picture-In-Picture (PIP) controller



SAB9083

FUNCTIONAL DESCRIPTION

Acquisition

The internal pixel rate is 28 MHz for the Y, U and V channels. It is expected that the bandwidth of the input signals is limited to 4.5 MHz for the Y input and 1.125 MHz for the U and V input. Inset synchronisation is achieved via the acquisition HSYNC and VSYNC pins of the main channel. The display is driven by the main channel clock. With the acquisition fine positioning added to a system constant the starting point of the acquisition can be controlled. With a nominal input HSYNC frequency, resulting in a nominal system frequency of 1792 × HSYNC (approximately 28 MHz) and standard NTSC signals 1408 samples are acquired and processed by the SAB9083.

PIP modes



SAB9083

Multistandard Picture-In-Picture (PIP) controller



I²C-bus description

The I²C-bus provides bidirectional 2-line communication between different ICs. The SDA line is the serial data line and the SCL serves as serial clock line. Both lines must be connected to a positive supply via a pull-up resistor when connected to the output stages of a device. Data transfer may be initiated only when the bus is not busy. The SAB9083 has the I²C-bus addresses 2C. Valid subaddresses are 00H to 18H, registers 15H to 18H are reserved for future extensions. I²C-bus control is according to the I²C-bus protocol: First a START sequence must be put on the I²C-bus, then the I²C-bus address of the circuit must be sent, then a subaddress. After this sequence the data of the subaddresses must be sent. An auto increment function gives the option to send data of the incremented subaddresses until a STOP sequence is sent. Table 1 gives an overview of the I²C-bus addresses. The data bits which are not used should be set to zero.

SUB				DATA	BYTES			
ADD	BIT 7	BIT 6	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0
00H	MPIPON	SPIPON	S1FLD	SFreeze	DNonint	PipMode2	PipMode1	PipMode0
01H	SHBlow1	SHBlow0	SHRed5	SHRed4	SHRed3	SHRed2	SHRed1	SHRed0
02H	SVBlow	SVRed6	SVRed5	SVRed4	SVRed3	SVRed2	SVRed1	SVRed0
03H	BGVfp3	BGVfp2	BGVfp1	BGVfp0	BGHfp3	BGHfp2	BGHfp1	BGHfp0
04H	SDHfp7	SDHfp6	SDHfp5	SDHfp4	SDHfp3	SDHfp2	SDHfp1	SDHfp0
05H	SDVfp7	SDVfp6	SDVfp5	SDVfp4	SDVfp3	SDVfp2	SDVfp1	SDVfp0
06H	SHPic7	SHPic6	SHPic5	SHPic4	SHPic3	SHPic2	SHPic1	SHPic0
07H	SVPic7	SVPic6	SVPic5	SVPic4	SVPic3	SVPic2	SVPic1	SVPic0
08H	MAHfp3	MAHfp2	MAHfp1	MAHfp0	SAHfp3	SAHfp2	SAHfp1	SAHfp0
09H	SAVfp7	SAVfp6	SAVfp5	SAVfp4	SAVfp3	SAVfp2	SAVfp1	SAVfp0
0AH	DUVPol	DVSPol	DFPol	DHsync	SUVPol	SVSPol	SFPol	SHsync
0BH	MainFidPos7	MainFidPos6	MainFidPos5	MainFidPos4	MainFidPos3	MainFidPos2	MainFidPos1	MainFidPos0
0CH	SubFidPos7	SubFidPos6	SubFidPos5	SubFidPos4	SubFidPos3	SubFidPos2	SubFidPos1	SubFidPos0
0DH	BGon	Bon	MFidPOn	SFidPOn	Prio	AlgOff	SFblkPkff1	SFblkPkff0
0EH	BSel1	BSel0	SBBrt1	SBBrt0	-	SBCol2	SBCol1	SBCol0
0FH	DPal	SPal	SLSel5	SLSel4	SLSel3	SLSel2	SLSel1	SLSel0
10H	I2CHold	SV	SDSel5	SDSel4	SDSel3	SDSel2	SDSel1	SDSel0
11H	MDHfp7	MDHfp6	MDHfp5	MDHfp4	MDHfp3	MDHfp2	MDHfp1	MDHfp0
12H	MDVfp7	MDVfp6	MDVfp5	MDVfp4	MDVfp3	MDVfp2	MDVfp1	MDVfp0
13H	MHBlow	-	MHRED5	MHRED4	MHRED3	MHRED2	MHRED1	MHRED0
14H	-	VBwidth2	VBwidth1	VBwidth0	_	HBwidth2	HBwidth1	HBwidth0
15H to 18H			•	all bits a	re reserved			

 Table 1
 Overview of I²C-bus addresses

controller

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MPIPON (DOUBLE WINDOW)

With MPIPON the main channel PIP is switched on (logic 1) or off (logic 0).

SPIPON

With SPIPON the sub PIPs are switched on (logic 1) or off (logic 0).

S1FLD

If S1FLD is set to logic 0 two fields are used for the live PIP. When a 50/60 Hz or a 60/50 Hz mode is detected the SAB908x automatically switches to the 1 Field mode (1 Field resolution vertically).

If S1FLD is set to logic 1 only one field is used. This causes joint line errors but saves memory. In normal modes this bit should not be set.

SFREEZE

With SFreeze set to logic 1 the current live sub PIP will be frozen. If set to logic 0 it is unfrozen.

ALGOFF

In double window mode precautions are taken to prevent a joint line error. Under some conditions this feature should be switched off. This can be realized by setting this bit to logic 1. AlgOff should be set to logic 0.

DNONINT

In normal mode (this bit is logic 0) the SAB9083 calculates whether a signal is non-interlaced or not and reacts accordingly. With the DNonint bit set to logic 1 the display channel is forced into non-interlaced mode. In the non-interlaced mode only one field is used during processing the PIPs.

PIP MODE AND REPLAY

The PIP modes for the SAB9083 are shown in Table 2.

Table 2 PIP modes

BITS	MODE
000	double window mode
001	replay PIP
010	multi PIP 3
011	multi PIP 6
100	reserved
111	multi PIP 6 split

SHRED AND SVRED (DOUBLE WINDOW)

SHRed and SVRed determine the reduction factor in the double window mode. The horizontal reduction is equal to SHRed/96 and the vertical reduction is equal to SVRed/96. SHRed should lie in the range from 0 to 48, if set to logic 0 the PIP is off. SVRed should lie in the range from 0 to 96, if set to logic 0 the PIP is off.

For the horizontal reduction factor, when the reduction factor is 48/96, 704 samples are processed. The HRed is linear so when HRed is e.g 24/96 352 samples are processed. For the vertical reduction factor the same holds but then with the number of lines. For NTSC the number of processed lines can be calculated from VRed/96 × 228 lines and for PAL this is VRed/96 × 276 lines.

SHPIC AND SVPIC (MULTI PIP MODES)

SHPic and SVPic control the picture size in the multi PIP modes. The horizontal range is 256 steps of four 28 MHz clocks. The vertical range is 256 steps of one line/field.

In the double window and replay PIP modes the picture size is determined by the reduction factors (SHRed and SVRed) and HBlow and VBlow.

BGHFP AND BGVFP

These bits control the horizontal and vertical positioning of the PIP configuration on the screen. The horizontal range is adjustable in 16 steps of four 28 MHz clocks. The vertical range is 16 steps of 1 line/field. The background colour can be adjusted with bits Bsel, SBBrt and SBCol.

SDHFP AND SDVFP

These bytes control the horizontal and vertical positioning of the sub PIPs on the screen. The horizontal range is 256 steps of eight 28 MHz clocks. The vertical range is 256 steps of 1 line/field.

MAHFP, SAHFP AND SAVFP

These bytes control the horizontal and vertical inset starting point of the acquired data. The horizontal range is 16 steps of eight 28 MHz clocks. The vertical range is 256 steps of 1 line/field.

DUVPOL, DVSPOL, DFPOL AND DHSYNC

These bits control the PLL/deflection settings. With DUVPol the polarity of the border UV signals can be inverted in case the deflection circuit behind the SAB9083 expects inverted signals. With DVSPol set to logic 0 the SAB9083 triggers on positive edges of the DVSYNC. If it is set to logic 1 it triggers on negative edges. DHSYNC determines the timing of the DHSYNC pulse. If it is set to logic 0 a burstkey is expected and if it is set to logic 1 a H-sync is expected. DFPol can invert the field ID of the incoming fields.

SUVPOL, SVSPOL, SFPOL AND SHSYNC

These bits control the PLL/Decoder settings. With SUVPol the polarity of the video UV signals can be inverted in case the decoder circuit before the SAB9083 gives inverted signals out. With SVSPol set to logic 0 the SAB9083 triggers on positive edges of the SVSYNC. If it is set to logic 1 it triggers on the negative edges. SHSYNC determines the timing of the SHSYNC pulse. If it is set to logic 0 a burstkey is expected and if it is set to logic 1 a H-sync is expected. SFPol can invert the field ID of the incoming fields.

MAIN, SUB FID POS ON (MFIDPON AND SFIDPON)

MFidPon (main) and SFidPon (sub) enable the field identification position fine tuning. The default value is off (logic 0), no fine positioning; when on (logic 1) the field identification position is determined by the value of M/S FIDPos.

BGON

BGOn determines whether a the background is visible. The background has a size of 720 pixels and 240 lines for NTSC and 720 pixels and 288 lines for PAL. The background colour can be adjusted with bits Bsel, SBBrt and SBCol.

BON, SBBRT, SBCOL AND BSEL

Bon can switch the sub-borders on (logic 1) or off (logic 0). SBBrt and SBCol set the brightness and colour type of the selected border. The brightness is set in 4 levels of 30%, 50%, 70% and 100% IRE. The colour type is one of black (grey), blue, red, magenta, green, cyan, yellow or white (gray). Bsel selects which colour is set, background or border.

Table 3 Bsel modes

BSEL	BORDER COLOUR SET
00	main
01	sub
10	background
11	sub-border select

MDHFP AND MDVFP

These bytes control the horizontal and vertical positioning of the main PIP on the screen. The horizontal range is 256 steps size of eight 28 MHz clocks. The vertical range is 256 steps of 1 line/field.

MHRED

MHRed can set the horizontal reduction factor, equal to MHRed/96, in a range from 0 to 48. If it is set to logic 0 the PIP is off, if it is set to 48 (the maximum value of MHRed) the horizontal reduction factor is 0.5.

SHBLOW AND SVBLOW (REPLAY MODE)

SHBlow and SVBlow are used in the replay mode. These bits can expand a pixel on the display side by a factor two (01) or four (11) in the horizontal direction (SHBlow) and a factor of two (1) in the vertical direction (SVBlow). Zero values indicate no expansion.

MHBLOW

MHBlow can expand the main picture by a factor of two in the horizontal direction.

SLSEL (REPLAY MODE)

In the replay PIP mode SLSel determines at which memory location the PIP data is written, the range depends on the memory usage for each PIP. The maximum number of PIPs that can be stored in NTSC mode is 42.

SLSEL (MULTI PIP MODES)

SLSel selects which of the PIPs in a multi PIP mode is live. In MP3 modes SLSel must be in the range from 0 to 2. In all MP6 modes SLSel must be in the range from 0 to 5.

SDSEL (REPLAY MODE)

SDSel selects which PIP is read from memory. Valid numbers are dependent on the maximum value of SLSel.

DPAL AND SPAL

In normal operation (DPal and SPal are logic 0) the SAB9083 calculates from the number of incoming lines whether the signal is NTSC (< 288 lines) or PAL (=> 288 lines). If SPal is set to logic 1 the channel is forced in PAL mode and 276 lines are acquired in stead of 228 in NTSC mode. If Dpal is set to logic 1 the main window is sized to 276 lines. If DPal is set to logic 1 and the subchannel is still NTSC the subchannel picture will be smaller than the main channel picture (difference of approximately 40 lines).

SFBLKPKOFF 1:0

SFBlkPkoff shifts the signals Fblk and Pkoff with respect to the YUV output, by half pixels.

Table 4 Shifts of FBLK and PKOff

SFBLKPKOFF	SHIFT OF FBLK AND PKOff
00	+0.5 pixel
01	no shift
10	–0.5 pixel
11	–1 pixel

I2CHOLD

I2CHOLD controls the updating of the I²C-bus controlled function towards the PIP. If set to logic 1 some register updates are on hold until the bit is set to logic 0. At the next main Vsync all settings are passed to the PIP functions.

The registers which are on hold when the I2CHold bit is set to logic 1 are:

- 1. MPIPON and SPIPON.
- 2. SHBlow and SVBlow.
- 3. SHRed and SVRed.
- 4. BGHfp and BGVfp.
- 5. BGOn and Bon.
- 6. SBBrt and SBCol.
- 7. MDHfp and MDVfp.
- 8. HBWidth and VBWidth.
- 9. DNonint.
- 10. BSel
- 11. SHPic and SVPic
- 12. SLSel.

SV

SV controls the internal horizontal offset of the background. When set to logic 0 the offset is 0.86 μ s, when set to logic 1 the offset is 4.56 μ s.

HBWIDTH AND VBWIDTH

These bits control the horizontal and vertical border size in steps of 2 pixels and 1 line. The default horizontal border size is 4 pixels and the vertical border size is 2 lines. Default means after power-up and no I^2C -bus data sent to the Picture-in-Picture controller.

In MP6 mode the minimum value of HBWidth is two.

NOTES

- 1. When the SAB9083 is set in the 1 field mode, joint line errors can occur. The SAB9083 is set in the 1 field mode by setting the 1Fld bit to logic 1.
- 2. When the SAB9083 is set in the non-interlace mode, joint line errors can occur. The SAB9083 is set in the non-interlace mode by setting the NonInt bit to logic 1.
- 3. When the input signals for the main and/or subchannel are non-interlaced signals, joint line errors can occur. When non-interlaced signals are input the SAB9083 switches automatically to the non-interlaced mode.
- 4. When the prevent joint line error algorithm is switched off (AlgOff is set to logic 1) joint line errors can still occur in the 2 field mode.
- 5. When a PAL signal is applied to the main channel and an NTSC signal is applied to the subchannel, the subchannel will automatically enter the 1 Field mode. Now a joint line error can occur. In the PAL/NTSC mode the subpicture will be smaller than the main picture (difference of approximately 40 lines).
- 6. When an NTSC signal is applied to the main channel and an PAL signal is applied to the subchannel, the subchannel will automatically enter the 1 Field mode. Now a joint line error can occur. In the NTSC/PAL mode the sub picture will be larger than the main picture (difference of approximately 40 lines):
- 7. The multi PIP modes are not meant for mixing PAL and NTSC PIPs.
- 8. In the MP6 mode the live PIP is displayed in the 1 Field mode when the input signal is PAL. This means that joint line errors can occur in the live PIP when the input signal is PAL.

Acquisition channel ADCs and clamping

The analog input signals are converted to digital signals by means of three ADCs. The resolution of the ADCs is 8-bit (DNL is 7-bit and INL is 6-bit) and the sampling is performed at the system frequency of 28 MHz for the Y input. A bias voltage (V_{bias}) is used for decoupling the AC components on internal references. The inputs should be AC-coupled and an internal clamp circuit, using external clamp capacitors, will clamp the input to $V_{ref(B)(DA)}$ for the luminance channels and to

 $(V_{ref(T)(DA)} - V_{ref(B)(DA)})/2 + LSB/2$ for the chrominance channels. The clamping starts at the active edge of the burst key. Internal video buffers amplify the standard input signals Y, U and V to the correct ADC levels.

PLL

The PLL generates an internal system clock of 1792 \times HSYNC, from the HSYNC, which is approximately 28 MHz.

DACs and video buffers

The 28 MHz digital video signals are fed to the 8-bit DACs which produce the required analog video signals. The video buffers amplify these signals prior to being fed to the output to drive another device.

LIMITING VALUES

In accordance with the Absolute Maximum Rating System (IEC 134).

SYMBOL	PARAMETER	MIN.	MAX.	UNIT
V _{DD}	supply voltage range	-0.5	tbf	V
T _{stg}	storage temperature	-25	+150	°C
T _{amb}	operating ambient temperature	0	70	°C
V _{esd}	electrostatic discharge handling	-	3	kV
R _{thj-a}	thermal resistance	-	45	K/W
P _{max}	maximum power dissipation	-	1.0	W

QUALITY SPECIFICATION

In accordance with "SNW-FQ-611, Part E", dated 14 december 1992.

ESD LEVELS

Table 5 ESD performance

PIN	SYMBOL	HUMAN BODY MODEL (V)	MACHINE MODEL (V)
20	V _{DD(RP)}	> 500	> 150
32	T5	400	> 150
33	T4	400	< 150 (estimated 50)
34	Т3	400	< 150 (estimated 50)
35	T2	400	> 150
36	T1	400	> 150
37	Т0	400	> 150
38	TC	> 500	> 150
39	V _{DDD(RL)}	> 500	> 150
40	V _{SSD(RL)}	> 500	> 150
41	V _{SSD(RM)}	> 500	> 150
42	V _{DDD(RM)}	> 500	> 150
43	TCLK	> 500	< 150 (estimated 50)
44	ТМ	> 500	> 150
45	TCBD	> 500	> 150
46	TCBC	> 500	> 150
47	TCBR	> 500	< 150 (estimated 50)
61	V _{SSD(RP)}	> 500	> 150
68	FBL	1000	standard specification
69	PKOFF	1000	standard specification
70	DVSYNC	1000	standard specification
72	SVSYNC	1000	standard specification
73	SCL	1000	standard specification
81	SV	> 3000	150
93	T7	> 3000	> 325
rest in range 1 to 17	all other pins	> 3000	> 325
rest in range 64 to 100	all other pins	> 3000	> 325

SAB9083

ANALOG CHARACTERISTICS

 V_{DDA} = 3.3 V; T_{amb} = 25 °C; unless otherwise specified.

SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
Supplies		•	-	•	•	-
V _{DD}	positive supply voltage		3.0	3.3	3.6	V
V _{SS}	ground voltage		_	0	-	V
$\Delta V_{DD(max)}$	maximum DC difference between supply voltages		-	0	100	mV
$\Delta V_{SS(max)}$	maximum DC difference between ground voltages		-	0	100	mV
I _{DDD(q)}	quiescent current of digital supply voltages	note 1	-	0	50	μA
I _{DDA(DP)}	display PLL supply current		_	1	-	mA
I _{DDA(SP}	sub PLL supply current		-	1	-	mA
I _{DDA(MA)}	main ADCs supply current	note 2	-	65	—	mA
I _{DDA(SA)}	sub ADCs supply current	note 2	-	65	-	mA
I _{DDA(DA)}	DACs supply current		-	tbf	-	mA
I _{DDA(tot)}	total analog supply current	note 2	-	tbf	-	mA
I _{DDD(tot)}	total digital supply current		-	tbf	-	mA
Analog-to-dig	ital converter and clamping	•			•	·
V _{ref(T)}	top reference voltage	note 3	tbf	2.90	tbf	V
V _{ref(B)}	bottom reference voltage	note 3	tbf	1.10	tbf	V
V _{iY(p-p)}	Y input signal amplitude (peak-to-peak value)	note 4	tbf	1.00	tbf	V
V _{i(V)(p-p)}	V input signal amplitude (peak-to-peak value)	note 4	tbf	1.05	tbf	V
V _{i(U)(p-p)}	U input signal amplitude (peak-to-peak value)	note 4	tbf	1.33	tbf	V
li	input current	clamping off	-	0.1	—	μA
		clamping on	-	tbf	-	μA
Ci	input capacitance		-	5	-	pF
f _s	sample frequency	note 5	-	1792 × HSYNC	-	kHz
RES	resolution		8	8	8	bit
DNL	differential non-linearity		-1.1	_	+1.1	LSB
INL	integral non-linearity		-2.0	_	+2.0	LSB
α_{cs}	channel separation		_	48	-	dB
PSRR	power supply rejection ratio		_	48	-	dB
V _{clamp(Y)}	Y clamping voltage level	note 6	_	1.28	-	V
V _{clamp(U,V)}	U/V clamping voltage level	note 6	_	2.00	_	V

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SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
Digital-to-ana	log converter and output stag	e	-	1		-
V _{ref(T)}	top reference voltage	note 2	tbf	1.19	tbf	V
V _{ref(B)}	bottom reference voltage		tbf	0.19	tbf	V
RL	load resistance		1	-	1000	kΩ
CL	load capacitance		0	-	5	pF
f _s	sample frequency	1FH, note 5	_	1792 × HSYNC	-	kHz
RES	resolution		8	8	8	bit
DNL	differential non-linearity		-1.0	-	+1.0	LSB
INL	integral non-linearity		-1.0	-	+1.0	LSB
α_{cs}	channel separation		-	48	-	dB
PSRR	power supply rejection ratio		-	48	-	dB
Display PLL a	nd clock generation	•	•	•		
f _{i(PLL)}	input frequency	note 1	14	15.75 (NTSC)	17	kHz
			14	15.625 (PAL)	17	kHz
Sub PLL and	clock generation		*		*	
f _{i(subPLL)}	input frequency	note 1	14	15.75 (NTSC)	17	kHz
·			14	15.625 (PAL)	17	kHz

Notes

1. Digital clocks are silent, POR and TM are connected to V_{DDA} .

- 2. This value is measured with an external bias resistor of 39 k Ω resulting in a bias current of 50 $\mu A.$
- 3. The V_{ref(T)} and V_{ref(B)} are made by a resistor division of the V_{DD}. They can be calculated with the formulas:

$$V_{ref(T)} = V_{DDA} \times \frac{2.90}{V_{DDA(nom)}} V$$
 and $V_{ref(B)} = V_{DDA} \times \frac{1.10}{V_{DDA(nom)}} V$

- 4. The input signal are amplified to meet an internal peak-to-peak voltage level of $0.8 \times [V_{ref(T)} V_{ref(B)}]$.
- 5. The internal system frequency is 1792 times the HSYNC input frequency for the subchannel.
- 6. The clamp level is not necessarily equal to the $V_{ref(B)}$ of the ADCs.
- 7. The UV-channels are clamped to: $\frac{(V_{ref(B)} + V_{ref(T)} + V_{LSB})}{2}$

DIGITAL CHARACTERISTICS

 V_{DDD} = 3.0 to 3.6 V; T_{amb} = 0 to 70 °C; unless otherwise specified.

SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
DC character	istics					-
V _{IH}	HIGH-level input voltage	default	0.8V _{DDD}	-	V _{DDD} + 0.5	V
		pin 74	0.8V _{DDD}	-	tbf	V
		5 V tolerant pins 68, 69, 70, 72, 73	0.8V _{DDD}	-	tbf	V
V _{IL}	LOW-level input voltage	default	-0.5	_	0.2V _{DDD}	V
		pins tbf	-0.5	-	0.2V _{DDD}	V
V _{hys}	hysteresis voltage		0.8	-	_	V
V _{OH}	HIGH-level output voltage	I _{OL} = -X mA; V _{DDD} = 3.0 V; note 1	0.85V _{DDD}	-	-	V
V _{OL}	LOW-level output voltage	I _{OL} =X mA; V _{DDD} = 3.0 V; note 1	-	-	0.4	V
		I _{OL} = 2 mA; V _{DDD} = 3.0 V	-	-	0.4	V
I _{LI}	input leakage current	V ₁ = 0	-	-	1	μA
		$V_{I} = V_{DDD}$	-	-	1	μA
I _{OZ}	3-state output leakage current	$V_{O} = 0 V \text{ or}$ $V_{O} = V_{DDD}$	-	-	1	μA
I _{lu(I/O)}	I/O latch-up current	$V < 0V; V > V_{DDD}$	200	-	_	mA
R _{pu}	internal pull-up resistor		16	33	78	kΩ
AC character	istics		•	•		
f _{sys}	system frequency	note 2	-	1792 × HSYNC	-	kHz
t _r	rise time		-	6	25	ns
t _f	fall time		-	6	25	ns

Note

1. X is the source/sink current under worst case conditions. X is reflected in the name of the I/O cell according to the drive capability. Minimum value of X is 1 mA.

2. The internal system frequency is 1792 times the HSYNC input frequency for the subchannel.

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Preliminary specification

TEST AND APPLICATION INFORMATION

The application diagram in a standard configuration is shown in Figure 5. The input signals main CVBS and sub CVBS of different video sources are processed by the SAB9083 and inserted by the YUV/RGB switch.



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PACKAGE OUTLINE



SOLDERING

Introduction to soldering surface mount packages

This text gives a very brief insight to a complex technology. A more in-depth account of soldering ICs can be found in our *"Data Handbook IC26; Integrated Circuit Packages"* (document order number 9398 652 90011).

There is no soldering method that is ideal for all surface mount IC packages. Wave soldering is not always suitable for surface mount ICs, or for printed-circuit boards with high population densities. In these situations reflow soldering is often used.

Reflow soldering

Reflow soldering requires solder paste (a suspension of fine solder particles, flux and binding agent) to be applied to the printed-circuit board by screen printing, stencilling or pressure-syringe dispensing before package placement.

Several methods exist for reflowing; for example, infrared/convection heating in a conveyor type oven. Throughput times (preheating, soldering and cooling) vary between 100 and 200 seconds depending on heating method.

Typical reflow peak temperatures range from 215 to 250 °C. The top-surface temperature of the packages should preferable be kept below 230 °C.

Wave soldering

Conventional single wave soldering is not recommended for surface mount devices (SMDs) or printed-circuit boards with a high component density, as solder bridging and non-wetting can present major problems.

To overcome these problems the double-wave soldering method was specifically developed.

If wave soldering is used the following conditions must be observed for optimal results:

- Use a double-wave soldering method comprising a turbulent wave with high upward pressure followed by a smooth laminar wave.
- For packages with leads on two sides and a pitch (e):
 - larger than or equal to 1.27 mm, the footprint longitudinal axis is preferred to be parallel to the transport direction of the printed-circuit board;
 - smaller than 1.27 mm, the footprint longitudinal axis must be parallel to the transport direction of the printed-circuit board.

The footprint must incorporate solder thieves at the downstream end.

• For packages with leads on four sides, the footprint must be placed at a 45° angle to the transport direction of the printed-circuit board. The footprint must incorporate solder thieves downstream and at the side corners.

During placement and before soldering, the package must be fixed with a droplet of adhesive. The adhesive can be applied by screen printing, pin transfer or syringe dispensing. The package can be soldered after the adhesive is cured.

Typical dwell time is 4 seconds at 250 °C. A mildly-activated flux will eliminate the need for removal of corrosive residues in most applications.

Manual soldering

Fix the component by first soldering two diagonally-opposite end leads. Use a low voltage (24 V or less) soldering iron applied to the flat part of the lead. Contact time must be limited to 10 seconds at up to 300 °C.

When using a dedicated tool, all other leads can be soldered in one operation within 2 to 5 seconds between 270 and 320 $^\circ\text{C}.$

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Suitability of surface mount IC packages for wave and reflow soldering methods

PACKAGE	SOLDERING METHOD		
PACKAGE	WAVE	REFLOW ⁽¹⁾	
BGA, SQFP	not suitable	suitable	
HLQFP, HSQFP, HSOP, HTSSOP, SMS	not suitable ⁽²⁾	suitable	
PLCC ⁽³⁾ , SO, SOJ	suitable	suitable	
LQFP, QFP, TQFP	not recommended ⁽³⁾⁽⁴⁾	suitable	
SSOP, TSSOP, VSO	not recommended ⁽⁵⁾	suitable	

Notes

- 1. All surface mount (SMD) packages are moisture sensitive. Depending upon the moisture content, the maximum temperature (with respect to time) and body size of the package, there is a risk that internal or external package cracks may occur due to vaporization of the moisture in them (the so called popcorn effect). For details, refer to the Drypack information in the "Data Handbook IC26; Integrated Circuit Packages; Section: Packing Methods".
- 2. These packages are not suitable for wave soldering as a solder joint between the printed-circuit board and heatsink (at bottom version) can not be achieved, and as solder may stick to the heatsink (on top version).
- 3. If wave soldering is considered, then the package must be placed at a 45° angle to the solder wave direction. The package footprint must incorporate solder thieves downstream and at the side corners.
- 4. Wave soldering is only suitable for LQFP, TQFP and QFP packages with a pitch (e) equal to or larger than 0.8 mm; it is definitely not suitable for packages with a pitch (e) equal to or smaller than 0.65 mm.
- 5. Wave soldering is only suitable for SSOP and TSSOP packages with a pitch (e) equal to or larger than 0.65 mm; it is definitely not suitable for packages with a pitch (e) equal to or smaller than 0.5 mm.

DEFINITIONS

Data sheet status				
Objective specification	This data sheet contains target or goal specifications for product development.			
Preliminary specification	This data sheet contains preliminary data; supplementary data may be published later.			
Product specification	This data sheet contains final product specifications.			
Limiting values				
Limiting values given are in accordance with the Absolute Maximum Rating System (IEC 134). Stress above one or more of the limiting values may cause permanent damage to the device. These are stress ratings only and operation of the device at these or at any other conditions above those given in the Characteristics sections of the specification is not implied. Exposure to limiting values for extended periods may affect device reliability.				
Application information				

Where application information is given, it is advisory and does not form part of the specification.

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