

16/32-BIT RISC FLASH MICROCONTROLLER

FEATURES

- High-Performance Static CMOS Technology
- SM470R1x 16/32-Bit RISC Core (ARM7TDMI™)
 - 60-MHz System Clock (Pipeline Mode)
 - Independent 16/32-Bit Instruction Set
 - Open Architecture With Third-Party Support
 - Built-In Debug Module
- Integrated Memory
 - 1M-Byte Program Flash
 - Two Banks With 16 Contiguous Sectors
 - 64K-Byte Static RAM (SRAM)
 - Memory Security Module (MSM)
 - JTAG Security Module
- Operating Features
 - Low-Power Modes: STANDBY and HALT
 - Industrial Temperature Range
- 470+ System Module
 - 32-Bit Address Space Decoding
 - Bus Supervision for Memory/Peripherals
 - Digital Watchdog (DWD) Timer
 - Analog Watchdog (AWD) Timer
 - Enhanced Real-Time Interrupt (RTI)
 - Interrupt Expansion Module (IEM)
 - System Integrity and Failure Detection
 - ICE Breaker
- Direct Memory Access (DMA) Controller
 - 32 Control Packets and 16 Channels
- Zero-Pin Phase-Locked Loop (ZPLL)-Based Clock Module With Prescaler
 - Multiply-by-4 or -8 Internal ZPLL Option
 - ZPLL Bypass Mode
- Twelve Communication Interfaces:
 - Two Serial Peripheral Interfaces (SPIs)
 - 255 Programmable Baud Rates
 - Three Serial Communication Interfaces (SCIs)
 - 2²⁴ Selectable Baud Rates
- Asynchronous/Isosynchronous Modes
- Two High-End CAN Controllers (HECC)
 - 32-Mailbox Capacity
 - Fully Compliant With CAN Protocol, Version 2.0B
- Five Inter-Integrated Circuit (I2C) Modules
 - Multi-Master and Slave Interfaces
 - Up to 400 Kbps (Fast Mode)
 - 7- and 10-Bit Address Capability
- High-End Timer Lite (HET)
 - 12 Programmable I/O Channels:
 - 12 High-Resolution Pins
 - High-Resolution Share Feature (XOR)
 - High-End Timer RAM
 - 64-Instruction Capacity
- External Clock Prescale (ECP) Module
 - Programmable Low-Frequency External Clock (CLK)
- 12-Channel, 10-Bit Multi-Buffered ADC (MibADC)
 - 64-Word FIFO Buffer
 - Single- or Continuous-Conversion Modes
 - 1.55-μs Minimum Sample and Conversion Time
 - Calibration Mode and Self-Test Features
- Flexible Interrupt Handling
- Expansion Bus Module (EBM)
 - Supports 8- and 16-Bit Expansion Bus Memory Interface Mappings
 - 42 I/O Expansion Bus Pins
- 46 Dedicated General-Purpose I/O (GIO) Pins and 47 Additional Peripheral I/Os
- Sixteen External Interrupts
- On-Chip Scan-Base Emulation Logic, IEEE Standard 1149.1 ⁽¹⁾ (JTAG) Test-Access Port
- Available in KGD, HFQ and HKP Packages

(1) The test-access port is compatible with the IEEE Standard 1149.1-1990, IEEE Standard Test-Access Port and Boundary Scan Architecture specification. Boundary scan is not supported on this device.



Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.

ARM7TDMI is a trademark of Advanced RISC Machines Limited (ARM).

SUPPORTS EXTREME TEMPERATURE APPLICATIONS

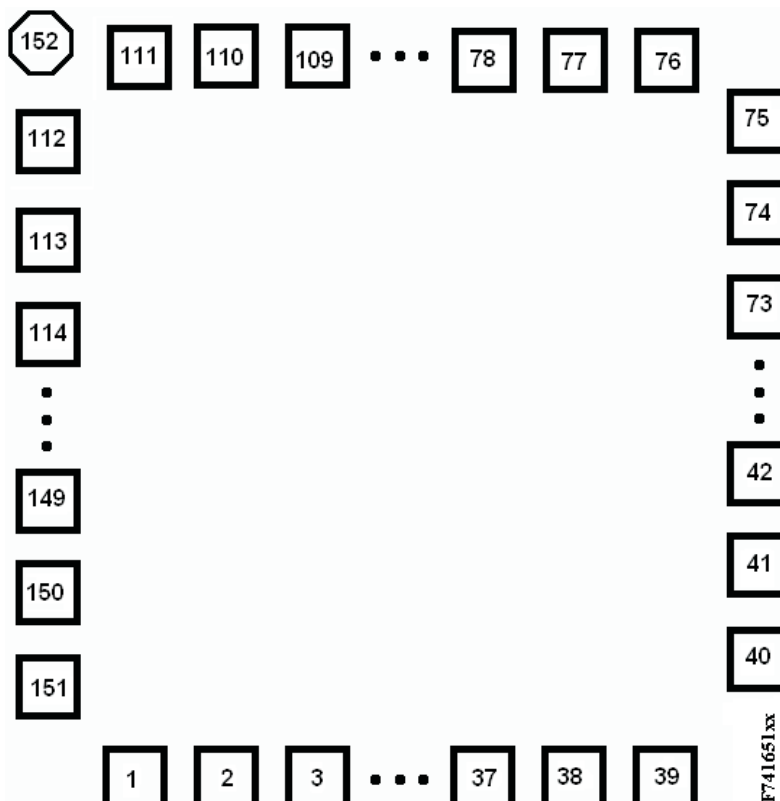
- **Controlled Baseline**
- **One Assembly/Test Site**
- **One Fabrication Site**
- **Available in Extreme (–55°C/220°C) Temperature Range ⁽²⁾**
- **Extended Product Life Cycle**
- **Extended Product-Change Notification**
- **Product Traceability**
- **Texas Instruments high temperature products utilize highly optimized silicon (die) solutions with design and process enhancements to maximize performance over extended temperatures.**

(2) Custom temperature ranges available

ORDERING INFORMATION⁽¹⁾

T _A	PACKAGE ⁽²⁾	ORDERABLE PART NUMBER	TOP-SIDE MARKING
–55°C to 220°C	KGD	SM470R1B1MKGDS1	SM470R1B1MKGDS1
	CQFP-HFQ	SM470R1B1MHFQS	SM470R1B1MHFQS
	CQFP-HKP	SM470R1B1MHKPS	SM470R1B1MHKPS

- (1) For the most current package and ordering information, see the Package Option Addendum at the end of this document, or see the TI Web site at www.ti.com.
- (2) Package drawings, standard packing quantities, thermal data, symbolization, and PCB design guidelines are available at www.ti.com/sc/package.

DIE LAYOUT

Table 1. Bare Die Information

DIE SIZE	DIE PAD SIZE	DIE PAD COORDINATES ⁽¹⁾	DIE THICKNESS	DIE PAD COMPOSITION	BACKSIDE FINISH	BACKSIDE POTENTIAL
208.858 x 211.890 mils/ 5304.99 x 5382.01 μm	65.1 x 65.1(μm)	See Table 2	11 mils	AlCu	Silicon with backgrind	Ground

- (1) Pads 12, 22, 26, 136, 143, 146, 149 and 152 are test pads, no connections required. It is highly recommended to leave them open.

Table 2. Bond Pad Coordinates

PAD NO.	BOND PAD COORDINATES (μm)				PAD SIZE (μm)	
	X MIN	Y MIN	X MAX	Y MAX	X	Y
1	178.955	10.08	244.055	75.18	65.1	65.1
2	368.2	10.08	433.3	75.18	65.1	65.1
3	557.445	10.08	622.545	75.18	65.1	65.1
4	664.335	10.08	729.435	75.18	65.1	65.1
5	853.58	10.08	918.68	75.18	65.1	65.1
6	1042.825	10.08	1107.925	75.18	65.1	65.1
7	1149.715	10.08	1214.815	75.18	65.1	65.1
8	1338.96	10.08	1404.06	75.18	65.1	65.1
9	1445.85	10.08	1510.95	75.18	65.1	65.1
10	1552.74	10.08	1617.84	75.18	65.1	65.1
11	1659.63	10.08	1724.73	75.18	65.1	65.1
12	1766.52	10.08	1831.62	75.18	65.1	65.1
13	1866.2	10.08	1931.3	75.18	65.1	65.1
14	1965.88	10.08	2030.98	75.18	65.1	65.1
15	2065.56	10.08	2130.66	75.18	65.1	65.1
16	2165.24	10.08	2230.34	75.18	65.1	65.1
17	2272.13	10.08	2337.23	75.18	65.1	65.1
18	2396.1	10.08	2461.2	75.18	65.1	65.1
19	2520.07	10.08	2585.17	75.18	65.1	65.1
20	2626.96	10.08	2692.06	75.18	65.1	65.1
21	2733.85	10.08	2798.95	75.18	65.1	65.1
22	2840.74	10.08	2905.84	75.18	65.1	65.1
23	2947.63	10.08	3012.73	75.18	65.1	65.1
24	3054.52	10.08	3119.62	75.18	65.1	65.1
25	3243.765	10.08	3308.865	75.18	65.1	65.1
26	3350.655	10.08	3415.755	75.18	65.1	65.1
27	3539.9	10.08	3605	75.18	65.1	65.1
28	3646.79	10.08	3711.89	75.18	65.1	65.1
29	3746.47	10.08	3811.57	75.18	65.1	65.1
30	3846.15	10.08	3911.25	75.18	65.1	65.1
31	3953.04	10.08	4018.14	75.18	65.1	65.1
32	4142.285	10.08	4207.385	75.18	65.1	65.1
33	4331.53	10.08	4396.63	75.18	65.1	65.1
34	4431.21	10.08	4496.31	75.18	65.1	65.1
35	4530.89	10.08	4595.99	75.18	65.1	65.1
36	4630.57	10.08	4695.67	75.18	65.1	65.1
37	4730.25	10.08	4795.35	75.18	65.1	65.1
38	4829.93	10.08	4895.03	75.18	65.1	65.1
39	4936.82	10.08	5001.92	75.18	65.1	65.1
40	5150.04	178.955	5215.14	244.055	65.1	65.1
41	5150.04	368.2	5215.14	433.3	65.1	65.1
42	5150.04	557.445	5215.14	622.545	65.1	65.1
43	5150.04	666.26	5215.14	731.36	65.1	65.1
44	5150.04	855.505	5215.14	920.605	65.1	65.1
45	5150.04	1044.75	5215.14	1109.85	65.1	65.1
46	5150.04	1153.565	5215.14	1218.665	65.1	65.1

Table 2. Bond Pad Coordinates (continued)

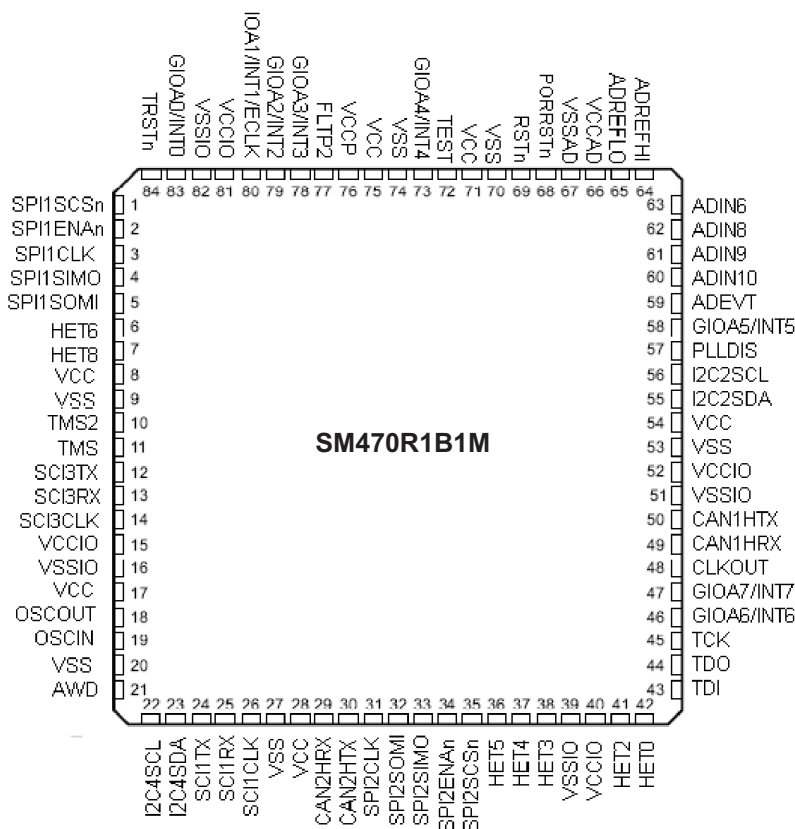
PAD NO.	BOND PAD COORDINATES (μm)				PAD SIZE (μm)	
	X MIN	Y MIN	X MAX	Y MAX	X	Y
47	5150.04	1262.38	5215.14	1327.48	65.1	65.1
48	5150.04	1371.195	5215.14	1436.295	65.1	65.1
49	5150.04	1480.01	5215.14	1545.11	65.1	65.1
50	5150.04	1669.255	5215.14	1734.355	65.1	65.1
51	5150.04	1778.07	5215.14	1843.17	65.1	65.1
52	5150.04	1886.885	5215.14	1951.985	65.1	65.1
53	5150.04	1995.7	5215.14	2060.8	65.1	65.1
54	5150.04	2184.945	5215.14	2250.045	65.1	65.1
55	5150.04	2293.76	5215.14	2358.86	65.1	65.1
56	5150.04	2402.575	5215.14	2467.675	65.1	65.1
57	5150.04	2511.39	5215.14	2576.49	65.1	65.1
58	5150.04	2700.635	5215.14	2765.735	65.1	65.1
59	5150.04	2809.45	5215.14	2874.55	65.1	65.1
60	5150.04	2998.695	5215.14	3063.795	65.1	65.1
61	5150.04	3187.94	5215.14	3253.04	65.1	65.1
62	5150.04	3296.755	5215.14	3361.855	65.1	65.1
63	5150.04	3486	5215.14	3551.1	65.1	65.1
64	5150.04	3675.245	5215.14	3740.345	65.1	65.1
65	5150.04	3784.06	5215.14	3849.16	65.1	65.1
66	5150.04	3973.305	5215.14	4038.405	65.1	65.1
67	5150.04	4082.12	5215.14	4147.22	65.1	65.1
68	5150.04	4190.935	5215.14	4256.035	65.1	65.1
69	5150.04	4299.75	5215.14	4364.85	65.1	65.1
70	5150.04	4408.565	5215.14	4473.665	65.1	65.1
71	5150.04	4517.38	5215.14	4582.48	65.1	65.1
72	5150.04	4626.195	5215.14	4691.295	65.1	65.1
73	5150.04	4735.01	5215.14	4800.11	65.1	65.1
74	5150.04	4843.825	5215.14	4908.925	65.1	65.1
75	5150.04	4952.64	5215.14	5017.74	65.1	65.1
76	4981.165	5148.85	5046.265	5213.95	65.1	65.1
77	4862.935	5148.85	4928.035	5213.95	65.1	65.1
78	4738.965	5148.85	4804.065	5213.95	65.1	65.1
79	4614.995	5148.85	4680.095	5213.95	65.1	65.1
80	4496.765	5148.85	4561.865	5213.95	65.1	65.1
81	4378.535	5148.85	4443.635	5213.95	65.1	65.1
82	4189.29	5148.85	4254.39	5213.95	65.1	65.1
83	4000.045	5148.85	4065.145	5213.95	65.1	65.1
84	3881.815	5148.85	3946.915	5213.95	65.1	65.1
85	3757.845	5148.85	3822.945	5213.95	65.1	65.1
86	3639.615	5148.85	3704.715	5213.95	65.1	65.1
87	3450.37	5148.85	3515.47	5213.95	65.1	65.1
88	3332.14	5148.85	3397.24	5213.95	65.1	65.1
89	3213.91	5148.85	3279.01	5213.95	65.1	65.1
90	3095.68	5148.85	3160.78	5213.95	65.1	65.1
91	2906.435	5148.85	2971.535	5213.95	65.1	65.1
92	2717.19	5148.85	2782.29	5213.95	65.1	65.1

Table 2. Bond Pad Coordinates (continued)

PAD NO.	BOND PAD COORDINATES (μm)				PAD SIZE (μm)	
	X MIN	Y MIN	X MAX	Y MAX	X	Y
93	2598.96	5148.85	2664.06	5213.95	65.1	65.1
94	2480.73	5148.85	2545.83	5213.95	65.1	65.1
95	2362.5	5148.85	2427.6	5213.95	65.1	65.1
96	2244.27	5148.85	2309.37	5213.95	65.1	65.1
97	2126.04	5148.85	2191.14	5213.95	65.1	65.1
98	1936.795	5148.85	2001.895	5213.95	65.1	65.1
99	1747.55	5148.85	1812.65	5213.95	65.1	65.1
100	1629.32	5148.85	1694.42	5213.95	65.1	65.1
101	1511.09	5148.85	1576.19	5213.95	65.1	65.1
102	1321.845	5148.85	1386.945	5213.95	65.1	65.1
103	1203.615	5148.85	1268.715	5213.95	65.1	65.1
104	1085.385	5148.85	1150.485	5213.95	65.1	65.1
105	967.155	5148.85	1032.255	5213.95	65.1	65.1
106	843.185	5148.85	908.285	5213.95	65.1	65.1
107	719.215	5148.85	784.315	5213.95	65.1	65.1
108	595.245	5148.85	660.345	5213.95	65.1	65.1
109	471.275	5148.85	536.375	5213.95	65.1	65.1
110	347.305	5148.85	412.405	5213.95	65.1	65.1
111	223.335	5148.85	288.435	5213.95	65.1	65.1
112	10.08	4979.975	75.18	5045.075	65.1	65.1
113	10.08	4868.5	75.18	4933.6	65.1	65.1
114	10.08	4757.025	75.18	4822.125	65.1	65.1
115	10.08	4645.55	75.18	4710.65	65.1	65.1
116	10.08	4534.075	75.18	4599.175	65.1	65.1
117	10.08	4410.105	75.18	4475.205	65.1	65.1
118	10.08	4286.135	75.18	4351.235	65.1	65.1
119	10.08	4162.165	75.18	4227.265	65.1	65.1
120	10.08	4038.195	75.18	4103.295	65.1	65.1
121	10.08	3912.825	75.18	3977.925	65.1	65.1
122	10.08	3801.35	75.18	3866.45	65.1	65.1
123	10.08	3689.875	75.18	3754.975	65.1	65.1
124	10.08	3578.4	75.18	3643.5	65.1	65.1
125	10.08	3466.925	75.18	3532.025	65.1	65.1
126	10.08	3355.45	75.18	3420.55	65.1	65.1
127	10.08	3243.975	75.18	3309.075	65.1	65.1
128	10.08	3132.5	75.18	3197.6	65.1	65.1
129	10.08	3021.025	75.18	3086.125	65.1	65.1
130	10.08	2909.55	75.18	2974.65	65.1	65.1
131	10.08	2720.305	75.18	2785.405	65.1	65.1
132	10.08	2608.83	75.18	2673.93	65.1	65.1
133	10.08	2497.355	75.18	2562.455	65.1	65.1
134	10.08	2385.88	75.18	2450.98	65.1	65.1
135	10.08	2274.405	75.18	2339.505	65.1	65.1
136	10.08	2162.93	75.18	2228.03	65.1	65.1
137	10.08	2051.455	75.18	2116.555	65.1	65.1
138	10.08	1862.21	75.18	1927.31	65.1	65.1

Table 2. Bond Pad Coordinates (continued)

PAD NO.	BOND PAD COORDINATES (μm)				PAD SIZE (μm)	
	X MIN	Y MIN	X MAX	Y MAX	X	Y
139	10.08	1672.965	75.18	1738.065	65.1	65.1
140	10.08	1561.49	75.18	1626.59	65.1	65.1
141	10.08	1372.245	75.18	1437.345	65.1	65.1
142	10.08	1260.77	75.18	1325.87	65.1	65.1
143	10.08	1149.295	75.18	1214.395	65.1	65.1
144	10.08	1037.82	75.18	1102.92	65.1	65.1
145	10.08	926.345	75.18	991.445	65.1	65.1
146	10.08	814.87	75.18	879.97	65.1	65.1
147	10.08	703.395	75.18	768.495	65.1	65.1
148	10.08	514.15	75.18	579.25	65.1	65.1
149	10.08	402.675	75.18	467.775	65.1	65.1
150	10.08	291.2	75.18	356.3	65.1	65.1
151	10.08	179.725	75.18	244.825	65.1	65.1
152	4.9	5154.1	69.93	5219.13	65.03	65.03

HFQ OR HKP PACKAGE**Figure 1. SM470R1B1M 84-Pin CQFP-HFQ/HKP (Top View)****Pin Assignments**

The SM470R1B1M 84-pin HFQ ceramic quad flatpack (CQFP) pin assignments are shown in [Figure 1](#).

Features

The reduced pin count version of SM470R1B1M has the following features.

- Communication interfaces
 - Two serial peripheral interface
 - Two serial communication interface
 - Two high-end CAN controllers
 - Two inter- integrated circuit (I2C) modules
- 4-channel, 10-bit multi-buffered ADC
- High-end timer lite (HET) controlling seven programmable I/O channels
- Eight general purpose I/O's

DESCRIPTION

The SM470R1B1M ⁽¹⁾ devices are members of the Texas Instruments SM470R1x family of general-purpose 16/32-bit reduced instruction set computer (RISC) microcontrollers. The B1M microcontroller offers high performance utilizing the high-speed ARM7TDMI 16/32-bit RISC central processing unit (CPU), resulting in a high instruction throughput while maintaining greater code efficiency. The ARM7TDMI 16/32-bit RISC CPU views memory as a linear collection of bytes numbered upwards from zero. The SM470R1B1M utilizes the big-endian format where the most significant byte of a word is stored at the lowest numbered byte and the least significant byte at the highest numbered byte.

High-end embedded control applications demand more performance from their controllers while maintaining low costs. The B1M RISC core architecture offers solutions to these performance and cost demands while maintaining low power consumption.

The B1M devices contain the following:

- ARM7TDMI 16/32-Bit RISC CPU
- SM470R1x system module (SYS) with 470+ enhancements
- 1M-byte flash
- 64K-byte SRAM
- Zero-pin phase-locked loop (ZPLL) clock module
- Digital watchdog (DWD) timer
- Analog watchdog (AWD) timer
- Enhanced real-time interrupt (RTI) module
- Interrupt expansion module (IEM)
- Memory security module (MSM)
- JTAG security module
- Two serial peripheral interface (SPI) modules
- Three serial communications interface (SCI) modules
- Two high-end CAN controllers (HECC)
- Five inter-integrated circuit (I2C) modules
- 10-bit multi-buffered analog-to-digital converter (MibADC), with 12 input channels
- High-end timer lite (HET) controlling 12 I/Os
- External clock prescale (ECP)
- Expansion bus module (EBM)
- Up to 93 I/O pins

The functions performed by the 470+ system module (SYS) include:

- Address decoding
- Memory protection
- Memory and peripherals bus supervision
- Reset and abort exception management
- Prioritization for all internal interrupt sources
- Device clock control
- Parallel signature analysis (PSA)

The enhanced real-time interrupt (RTI) module on the B1M has the option to be driven by the oscillator clock. The digital watchdog (DWD) is a 25-bit resettable decremmenting counter that provides a system reset when the watchdog counter expires. This data sheet includes device-specific information such as memory and peripheral select assignment, interrupt priority, and a device memory map. For a more detailed functional description of the SYS module, see the *TMS470R1x System Module Reference Guide* (literature number SPNU189).

The B1M memory includes general-purpose SRAM supporting single-cycle read/write accesses in byte, half-word, and word modes.

(1) Throughout the remainder of this document, the SM470R1B1M will be referred to as either the full device name or as B1M.

The flash memory on this device is a nonvolatile, electrically erasable and programmable memory implemented with a 32-bit-wide data bus interface. The flash operates with a system clock frequency of up to 24 MHz or 30 MHz, depending on the input voltage. When in pipeline mode, the flash operates with a system clock frequency of up to 48 MHz or 60 MHz, depending on the input voltage. For more detailed information on the flash, see the *F05 Flash* section of this data sheet.

The memory security module (MSM) and the JTAG security module prevent unauthorized access and visibility to on-chip memory, thereby preventing reverse engineering or manipulation of proprietary code.

The B1M device has twelve communication interfaces: two SPIs, three SCIs, two HECCs, and five I2Cs. The SPI provides a convenient method of serial interaction for high-speed communications between similar shift-register type devices. The SCI is a full-duplex, serial I/O interface intended for asynchronous communication between the CPU and other peripherals using the standard non-return-to-zero (NRZ) format. The HECC uses a serial, multimaster communication protocol that efficiently supports distributed real-time control with robust communication rates of up to 1 megabit per second (Mbps). These CAN peripherals are ideal for applications operating in noisy and harsh environments (e.g., industrial fields) that require reliable serial communication or multiplexed wiring. The I2C module is a multi-master communication module providing an interface between the B1M microcontroller and an I2C-compatible device via the I2C serial bus. The I2C supports both 100 Kbps and 400 Kbps speeds. For more detailed functional information on the SPI, SCI, and CAN peripherals, see the specific reference guides (literature numbers SPNU195, SPNU196, and SPNU197). For more detailed functional information on the I2C, see the *TMS470R1x Inter-Integrated Circuit (I2C) Reference Guide* (literature number SPNU223).

The HET is an advanced intelligent timer that provides sophisticated timing functions for real-time applications. The timer is software-controlled, using a reduced instruction set, with a specialized timer micromachine and an attached I/O port. The HET can be used for compare, capture, or general-purpose I/O. It is especially well suited for applications requiring multiple sensor information and drive actuators with complex and accurate time pulses. The HET used in this device is the high-end timer lite. It has fewer I/Os than the usual 32 in a standard HET. For more detailed functional information on the HET, see the *TMS470R1x High-End Timer (HET) Reference Guide* (literature number SPNU199).

The B1M HET peripheral contains the XOR-share feature. This feature allows two adjacent HET high-resolution channels to be XORed together, making it possible to output smaller pulses than a standard HET. For more detailed information on the HET XOR-share feature, see the *TMS470R1x High-End Timer (HET) Reference Guide* (literature number SPNU199).

The B1M device has one 10-bit-resolution, sample-and-hold MibADC. Each of the MibADC channels can be converted individually or can be grouped by software for sequential conversion sequences. There are three separate groupings, two of which can be triggered by an external event. Each sequence can be converted once when triggered or configured for continuous conversion mode. For more detailed functional information on the MibADC, see the *TMS470R1x Multi-Buffered Analog-to-Digital Converter (MibADC) Reference Guide* (literature number SPNU206).

The zero-pin phase-locked loop (ZPLL) clock module contains a phase-locked loop, a clock-monitor circuit, a clock-enable circuit, and a prescaler (with prescale values of 1–8). The function of the ZPLL is to multiply the external frequency reference to a higher frequency for internal use. The ZPLL provides ACLK to the system (SYS) module. The SYS module subsequently provides system clock (SYSCLK), real-time interrupt clock (RTICLK), CPU clock (MCLK), and peripheral interface clock (ICLK) to all other B1M device modules. For more detailed functional information on the ZPLL, see the *TMS470R1x Zero-Pin Phase-Locked Loop (ZPLL) Clock Module Reference Guide* (literature number SPNU212).

NOTE

ACLK should not be confused with the MibADC internal clock, ADCLK. ACLK is the continuous system clock from an external resonator/crystal reference.

The expansion bus module (EBM) is a standalone module that supports the multiplexing of the GIO functions and the expansion bus interface. For more information on the EBM, see the *TMS470R1x Expansion Bus Module (EBM) Reference Guide* (literature number SPNU222).

The B1M device also has an external clock prescaler (ECP) module that when enabled, outputs a continuous external clock (ECLK) on a specified GIO pin. The ECLK frequency is a user-programmable ratio of the peripheral interface clock (ICLK) frequency. For more detailed functional information on the ECP, see the *TMS470R1x External Clock Prescaler (ECP) Reference Guide* (literature number SPNU202).

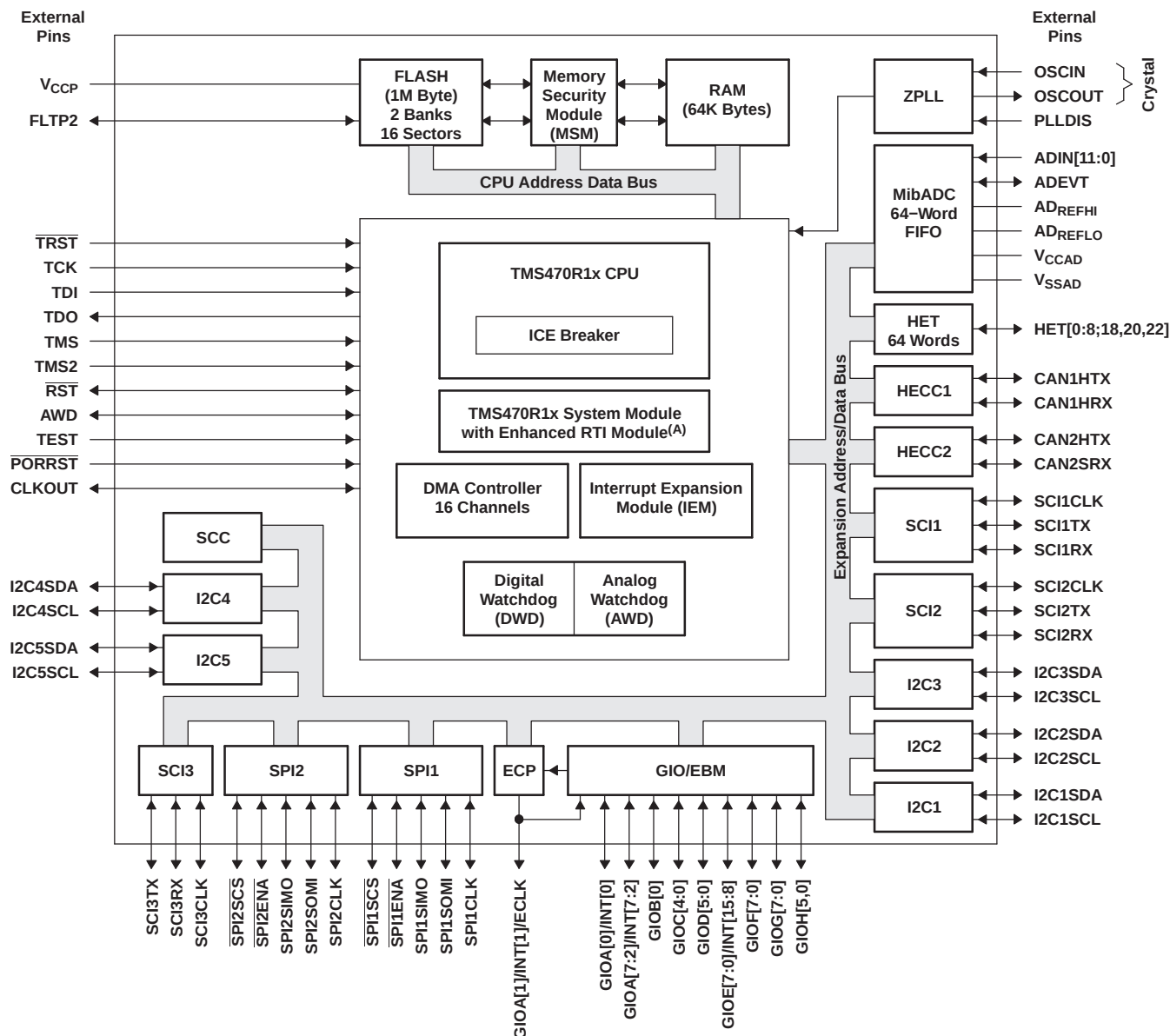
Device Characteristics

[Table 3](#) identifies all the characteristics of the B1M device except the SYSTEM and CPU, which are generic.

Table 3. Device Characteristics

CHARACTERISTICS	DEVICE DESCRIPTION SM470R1B1M	COMMENTS
MEMORY		
For the number of memory selects on this device, see Table 5 , <i>SM470R1B1M Memory Selection Assignment</i> .		
INTERNAL MEMORY	Pipeline/Non-Pipeline 1M-Byte flash 64K-Byte SRAM Memory Security Module (MSM) JTAG Security Module	Flash is pipeline-capable. The B1M RAM is implemented in one 64K array selected by two memory-select signals (see Table 5 , <i>SM470R1B1M Memory Selection Assignment</i>).
PERIPHERALS		
For the device-specific interrupt priority configurations, see Table 8 , <i>Interrupt Priority</i> . And for the 1K peripheral address ranges and their peripheral selects, see Table 6 , <i>B1M Peripherals, System Module, and Flash Base Addresses</i> .		
CLOCK	ZPLL	Zero-pin PLL has no external loop filter pins.
Expansion Bus	EBM	Expansion bus module with 42 pins. Supports 8- and 16-bit memories. See Table 9 for details.
GENERAL-PURPOSE I/Os	46 I/O	Port A has 8 external pins; Port B has only 1 external pin; Port C has 5 external pins; Port D has 6 external pins; Ports E, F, and G each have 8 external pins; and Port H has 2 external pins.
ECP	YES	
SCI	3 (3-pin)	
CAN (HECC and/or SCC)	2 HECC	Two high-end CAN controllers
SPI (5-pin, 4-pin or 3-pin)	2 (5-pin)	
I2C	5	
HET with XOR Share	12 I/O	The high-resolution (HR) SHARE feature allows even-numbered HR pins to share the next higher odd-numbered HR pin structures. This HR sharing is independent of whether or not the odd pin is available externally. If an odd pin is available externally and <i>shared</i> , then the odd pin can only be used as a general-purpose I/O. For more information on HR SHARE, see the <i>TMS470R1x High-End Timer (HET) Reference Guide</i> (literature number SPNU199).
HET RAM	64-Instruction Capacity	
MibADC	10-bit, 12-channel 64-word FIFO	Both the logic and registers for a full 16-channel MibADC are present.
CORE VOLTAGE	1.8 V	
I/O VOLTAGE	3.3 V	
PINS	84	
PACKAGES	HFQ, HKP	

Functional Block Diagram



A. The enhanced RTI module is the system module with two extra bits to disable the ZPLL while in STANDBY mode.

Table 4. Terminal Functions

TERMINAL			TYPE ⁽¹⁾ (2)	CURRENT OUTPUT	INTERNAL PULLUP/ PULLDOWN ⁽³⁾	DESCRIPTION
NAME	PAD NO.	HFQ/ HKP PIN NO. ⁽⁴⁾				
HIGH-END TIMER (HET)						
HET[0]	76	42	3.3 V	2 mA -z	IPD (20 µA)	Timer input capture or output compare. The HET[8:0,18,20,22] applicable pins can be programmed as general-purpose input/output (GIO) pins. All are high-resolution pins. The high-resolution (HR) SHARE feature allows even HR pins to share the next higher odd HR pin structures. This HR sharing is independent of whether or not the odd pin is available externally. If an odd pin is available externally and shared, then the odd pin can only be used as a general-purpose I/O. For more information on HR SHARE, see the <i>TMS470R1x High-End Timer (HET) Reference Guide</i> (literature number SPNU199).
HET[1]	75	NC				
HET[2]	74	41				
HET[3]	69	38				
HET[4]	68	37				
HET[5]	66	36				
HET[6]	9	6				
HET[7]	11	NC				
HET[8]	13	7				
HET[18]	16	NC				
HET[20]	19	NC				
HET[22]	20	NC				
HIGH-END CAN CONTROLLER (HECC)						
CAN1HRX	86	49	5-V tolerant	4 mA		HECC1 receive pin or GIO pin
CAN1HTX	87	50	3.3 V	2 mA -z	IPU (20 µA)	HECC1 transmit pin or GIO pin
CAN2HRX	57	29	5-V tolerant	4 mA		HECC2 receive pin or GIO pin
CAN2HTX	58	30	3.3 V	2 mA -z	IPU (20 µA)	HECC2 transmit pin or GIO pin
GENERAL-PURPOSE I/O (GIO)						
GIOA[0]/INT[0]	144	83	5-V tolerant	4 mA		General-purpose input/output pins. GIOA[7:0]/INT[7:0] are interrupt-capable pins. GIOA[1]/INT[1]/ECLK pin is multiplexed with the external clock-out function of the external clock prescale (ECP) module.
GIOA[1]/INT[1]/ECLK	140	80				
GIOA[2]/INT[2]	138	79				
GIOA[3]/INT[3]	137	78				
GIOA[4]/INT[4]	130	73				
GIOA[5]/INT[5]	101	58				
GIOA[6]/INT[6]	81	46				
GIOA[7]/INT[7]	82	47				
GIOB[0]/EBDMAREQ0	46	NC	3.3 V	2 mA -z	IPD (20 µA)	GIOB[0], GIOC[4:0], GIOD[5:0], GIOE[7:0:], GIOF[7:0:], GIOG[7:0:], and GIOH[5:0] are multiplexed with the expansion bus module. See Table 9 .
GIOC[0]/EBOE	139	NC				
GIOC[1]/EBWR[0]	131	NC				
GIOC[2]/EBWR[1]	129	NC				
GIOC[3]/EBCS[5]	123	NC				
GIOC[4]/EBCS[6]	122	NC				

(1) PWR = power, GND = ground, REF = reference voltage, NC = no connect

(2) All I/O pins, except $\overline{RST}$, are configured as inputs while $\overline{PORRST}$ is low and immediately after $\overline{PORRST}$ goes high.

(3) IPD = internal pulldown, IPU = internal pullup (all internal pullups and pulldowns are active on input pins, independent of the $\overline{PORRST}$ state.)

(4) Any pins marked as NC are physically connected to ground internal to the package. Care must be used to keep these pins in a high impedance input state.

Table 4. Terminal Functions (continued)

TERMINAL			TYPE ⁽¹⁾ (2)	CURRENT OUTPUT	INTERNAL PULLUP/PULLDOWN ⁽³⁾	DESCRIPTION
NAME	PAD NO.	HFQ/HKP PIN NO. ⁽⁴⁾				
GIOD[0]/EBADDR[0]	45	NC	3.3 V	2 mA -z	IPD (20 µA)	GIOB[0], GIOC[4:0], GIOD[5:0], GIOE[7:0], GIOF[7:0], GIOG[7:0], and GIOH[5,0] are multiplexed with the expansion bus module. GIOF[7:0]/INT[15:8] are interrupt-capable pins. See Table 9 .
GIOD[1]/EBADDR[1]	42	NC				
GIOD[2]/EBADDR[2]	38	NC				
GIOD[3]/EBADDR[3]	33	NC				
GIOD[4]/EBADDR[4]	30	NC				
GIOD[5]/EBADDR[5]	25	NC				
GIOE[0]/EBDATA[0]	47	NC				
GIOE[1]/EBDATA[1]	50	NC				
GIOE[2]/EBDATA[2]	61	NC				
GIOE[3]/EBDATA[3]	64	NC				
GIOE[4]/EBDATA[4]	67	NC				
GIOE[5]/EBDATA[5]	70	NC				
GIOE[6]/EBDATA[6]	73	NC				
GIOE[7]/EBDATA[7]	80	NC				
GIOF[0]/INT[8]/EBADDR[6]/EBDATA[8]	83	NC				
GIOF[1]/INT[9]/EBADDR[7]/EBDATA[9]	85	NC				
GIOF[2]/INT[10]/EBADDR[8]/EBDATA[10]	92	NC				
GIOF[3]/INT[11]/EBADDR[9]/EBDATA[11]	93	NC				
GIOF[4]/INT[12]/EBADDR[10]/EBDATA[12]	96	NC				
GIOF[5]/INT[13]/EBADDR[11]/EBDATA[13]	99	NC				
GIOF[6]/INT[14]/EBADDR[12]/EBDATA[14]	102	NC				
GIOF[7]/INT[15]/EBADDR[13]/EBDATA[15]	103	NC				
GIOG[0]/EBADDR[14]/EBADDR[6]	21	NC				
GIOG[1]/EBADDR[15]/EBADDR[7]	10	NC				
GIOG[2]/EBADDR[16]/EBADDR[8]	8	NC				
GIOG[3]/EBADDR[17]/EBADDR[9]	6	NC				
GIOG[4]/EBADDR[18]/EBADDR[10]	3	NC				
GIOG[5]/EBADDR[19]/EBADDR[11]	150	NC				
GIOG[6]/EBADDR[20]/EBADDR[12]	148	NC				
GIOG[7]/EBADDR[21]/EBADDR[13]	145	NC				
GIOH[0]/EBADDR[22]/EBADDR[14]	144	NC				
GIOH[5]/EBHOLD	128	NC				

Table 4. Terminal Functions (continued)

TERMINAL			TYPE ⁽¹⁾ (2)	CURRENT OUTPUT	INTERNAL PULLUP/ PULLDOWN ⁽³⁾	DESCRIPTION
NAME	PAD NO.	HFQ/ HKP PIN NO. ⁽⁴⁾				
MULTI-BUFFERED ANALOG-TO-DIGITAL CONVERTER (MibADC)						
ADEVT	104	59	3.3 V	2 mA -z	IPD (20 μA)	MibADC event input. Can be programmed as a GIO pin.
ADIN[0]	120	NC		MibADC analog input pins		
ADIN[1]	119	NC				
ADIN[2]	118	NC				
ADIN[3]	117	NC				
ADIN[4]	116	NC				
ADIN[5]	111	NC				
ADIN[6]	110	63				
ADIN[7]	109	NC				
ADIN[8]	108	62				
ADIN[9]	107	61				
ADIN[10]	106	60				
ADIN[11]	105	NC				
AD _{REFHI}	112	64	3.3 VREF			MibADC module high-voltage reference input
AD _{REFLO}	113	65	GND REF			MibADC module low-voltage reference input
V _{CCAD}	114	66	3.3-V PWR			MibADC analog supply voltage
V _{SSAD}	115	67	GND			MibADC analog ground reference
SERIAL PERIPHERAL INTERFACE 1 (SPI1)						
SPI1CLK	4	3	5-V tolerant	4 mA		SPI1 clock. SPI1CLK can be programmed as a GIO pin.
$\overline{\text{SPI1ENA}}$	2	2				SPI1 chip enable. Can be programmed as a GIO pin.
$\overline{\text{SPI1SCS}}$	1	1				SPI1 slave chip select. Can be programmed as a GIO pin.
SPI1SIMO	5	4				SPI1 data stream. Slave in/master out. Can be programmed as a GIO pin.
SPI1SOMI	7	5				SPI1 data stream. Slave out/master in. Can be programmed as a GIO pin.
SERIAL PERIPHERAL INTERFACE 2 (SPI2)						
SPI2CLK	59	31	5-V tolerant	4 mA		SPI2 clock. Can be programmed as a GIO pin.
$\overline{\text{SPI2ENA}}$	63	34				SPI2 chip enable. Can be programmed as a GIO pin.
$\overline{\text{SPI2SCS}}$	65	35				SPI2 slave chip select. Can be programmed as a GIO pin.
SPI2SIMO	62	33				SPI2 data stream. Slave in/master out. Can be programmed as a GIO pin.
SPI2SOMI	60	32				SPI2 data stream. Slave out/master in. Can be programmed as a GIO pin.
INTER-INTEGRATED CIRCUIT 1 (I2C1)						
I2C1SDA	90	NC	5-V tolerant	4 mA		I2C1 serial data pin or GIO pin
I2C1SCL	91	NC				I2C1 serial clock pin or GIO pin

Table 4. Terminal Functions (continued)

TERMINAL			TYPE ⁽¹⁾ (2)	CURRENT OUTPUT	INTERNAL PULLUP/ PULLDOWN ⁽³⁾	DESCRIPTION
NAME	PAD NO.	HFQ/ HKP PIN NO. ⁽⁴⁾				
INTER-INTEGRATED CIRCUIT 2 (I2C2)						
I2C2SDA	97	55	5-V tolerant	4 mA		I2C2 serial data pin or GIO pin
I2C2SCL	98	56				I2C2 serial clock pin or GIO pin
INTER-INTEGRATED CIRCUIT 3 (I2C3)						
I2C3SDA	32	NC	5-V tolerant	4 mA		I2C3 serial data pin or GIO pin
I2C3SCL	31	NC				I2C3 serial clock pin or GIO pin
INTER-INTEGRATED CIRCUIT 4 (I2C4)						
I2C4SDA	44	23	5-V tolerant	4 mA		I2C4 serial data pin or GIO pin
I2C4SCL	43	22				I2C4 serial clock pin or GIO pin
INTER-INTEGRATED CIRCUIT 5 (I2C5)						
I2C5SDA	41	NC	5-V tolerant	4 mA		I2C5 serial data pin or GIO pin
I2C5SCL	40	NC				I2C5 serial clock pin or GIO pin
ZERO-PIN PHASE-LOCKED LOOP (ZPLL)						
OSCIN	36	19	1.8 V			Crystal connection pin or external clock input
OSCOUT	35	18		2 mA		External crystal connection pin
PLLDIS	100	57	3.3 V		IPD (20 µA)	Enable/disable the ZPLL. The ZPLL can be bypassed and the oscillator becomes the system clock. If not in bypass mode, TI recommends that this pin be connected to ground or pulled down to ground by an external resistor.
SERIAL COMMUNICATIONS INTERFACE 1 (SCI1)						
SCI1CLK	51	26	3.3 V	2 mA -z	IPD (20 µA)	SCI1 clock. SCI1CLK can be programmed as a GIO pin.
SCI1RX	49	25	5-V tolerant	4 mA		SCI1 data receive. SCI1RX can be programmed as a GIO pin.
SCI1TX	48	24	3.3 V	2 mA -z	IPU (20 µA)	SCI1 data transmit. SCI1TX can be programmed as a GIO pin.
SERIAL COMMUNICATIONS INTERFACE 2 (SCI2)						
SCI2CLK	54	NC	3.3 V	2 mA -z	IPD (20 µA)	SCI2 clock. SCI2CLK can be programmed as a GIO pin.
SCI2RX	53	NC	5-V tolerant	4 mA		SCI2 data receive. SCI2RX can be programmed as a GIO pin.
SCI2TX	52	NC	3.3 V	2 mA -z	IPU (20 µA)	SCI2 data transmit. SCI2TX can be programmed as a GIO pin.
SERIAL COMMUNICATIONS INTERFACE 3 (SCI3)						
SCI3CLK	27	14	3.3 V	2 mA -z	IPD (20 µA)	SCI3 clock. SCI3CLK can be programmed as a GIO pin.
SCI3RX	24	13	5-V tolerant	4 mA		SCI3 data receive. SCI3RX can be programmed as a GIO pin.
SCI3TX	23	12	3.3 V	2 mA -z	IPU (20 µA)	SCI3 data transmit. SCI3TX can be programmed as a GIO pin.
SYSTEM MODULE (SYS)						
CLKOUT	84	48	3.3 V	8 mA		Bidirectional pin. CLKOUT can be programmed as a GIO pin or the output of SYCLK, ICLK, or MCLK.
PORRST	121	68	3.3 V		IPD (20 µA)	Input master chip power-up reset. External V _{CC} monitor circuitry must assert a power-on reset.

Table 4. Terminal Functions (continued)

TERMINAL			TYPE ⁽¹⁾ (2)	CURRENT OUTPUT	INTERNAL PULLUP/ PULLDOWN ⁽³⁾	DESCRIPTION
NAME	PAD NO.	HFQ/ HKP PIN NO. ⁽⁴⁾				
$\overline{\text{RST}}$	124	69	3.3 V	4 mA	IPU (20 μA)	Bidirectional reset. The internal circuitry can assert a reset, and an external system reset can assert a device reset. On this pin, the output buffer is implemented as an open drain (drives low only). To ensure an external reset is not arbitrarily generated, TI recommends that an external pullup resistor be connected to this pin.
WATCHDOG/REAL-TIME INTERRUPT (WD/RTI)						
AWD	39	21	3.3 V	8 mA		Analog watchdog reset. The AWD pin provides a system reset if the WD KEY is not written in time by the system, providing an external RC network circuit is connected. If the user is not using AWD, TI recommends that this pin be connected to ground or pulled down to ground by an external resistor. For more details on the external RC network circuit, see the <i>TMS470R1x System Module Reference Guide</i> (literature number SPNU189).
TEST/DEBUG (T/D)						
TCK	79	45	3.3 V		IPD (20 μA)	Test clock. TCK controls the test hardware (JTAG).
TDI	77	43		8 mA	IPU (20 μA)	Test data in. TDI inputs serial data to the test instruction register, test data register, and programmable test address (JTAG).
TDO	78	44		8 mA	IPD (20 μA)	Test data out. TDO outputs serial data from the test instruction register, test data register, identification register, and programmable test address (JTAG).
TEST	127	72	3.3 V		IPD (20 μA)	Test enable. Reserved for internal use only. TI recommends that this pin be connected to ground or pulled down to ground by an external resistor.
TMS	18	11		8 mA	IPU (20 μA)	Serial input for controlling the state of the CPU test access port (TAP) controller (JTAG).
TMS2	17	10		8 mA	IPU (20 μA)	Serial input for controlling the second TAP. TI recommends that this pin be connected to V_{CCIO} or pulled up to V_{CCIO} by an external resistor.
$\overline{\text{TRST}}$	151	84			IPD (20 μA)	Test hardware reset to TAP1 and TAP2. IEEE Standard 1149-1 (JTAG) Boundary-Scan Logic. TI recommends that this pin be pulled down to ground by an external resistor.
FLASH						
FLTP2	135	77	NC	NC		Flash test pad 2. For proper operation, this pin must not be connected [no connect (NC)].
V_{CCP}	134	76	3.3-V PWR			Flash external pump voltage (3.3 V)
SUPPLY VOLTAGE CORE (1.8 V)						
V_{CC}	14	8	1.8-V PWR			Core logic supply voltage
	34	17				
	56	28				
	95	54				
	126	71				
	133	75				

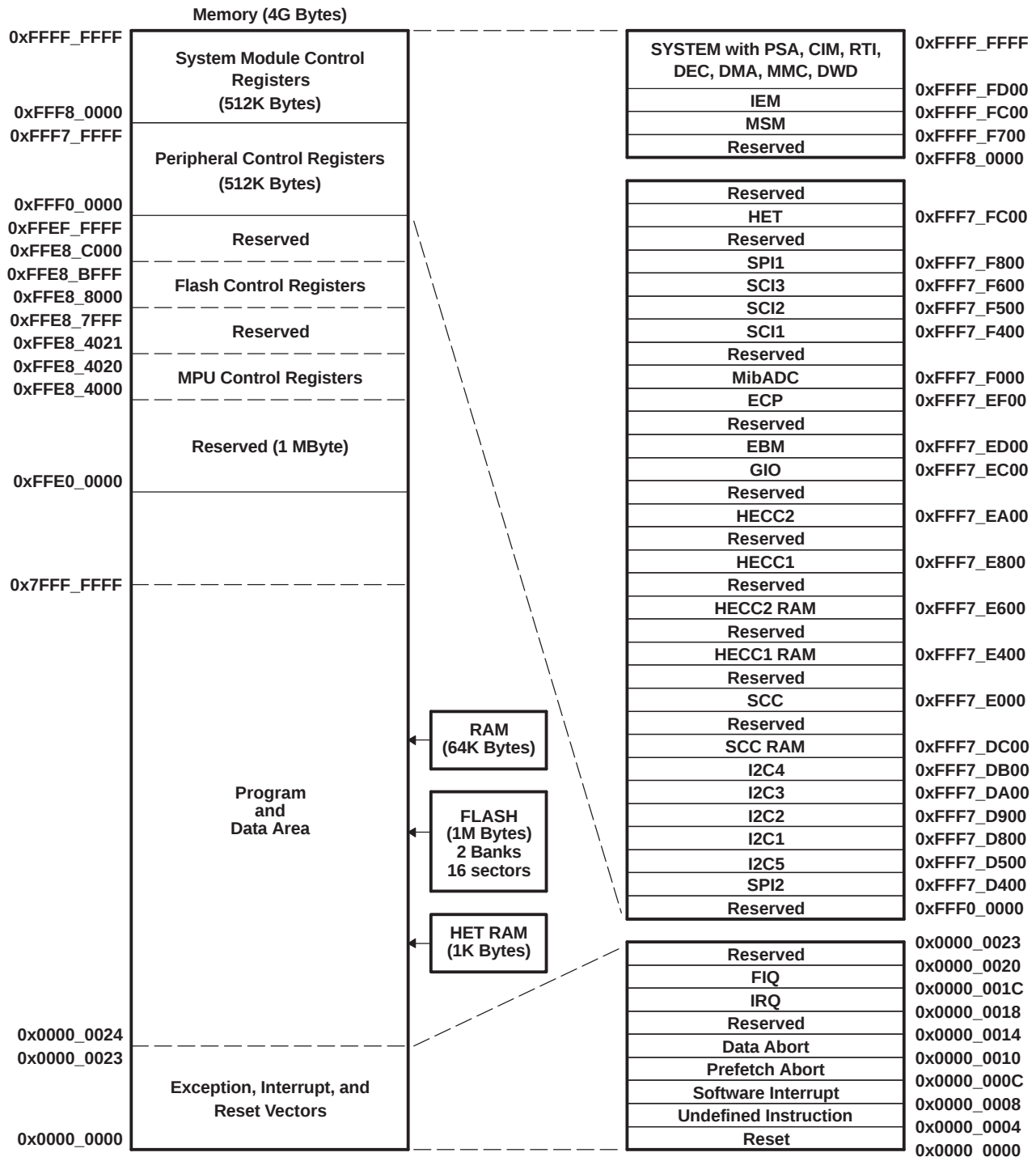
Table 4. Terminal Functions (continued)

TERMINAL			TYPE ⁽¹⁾ (2)	CURRENT OUTPUT	INTERNAL PULLUP/ PULLDOWN ⁽³⁾	DESCRIPTION
NAME	PAD NO.	HFQ/ HKP PIN NO. ⁽⁴⁾				
SUPPLY VOLTAGE DIGITAL I/O (3.3 V)						
V _{CCIO}	28	15	3.3-V PWR			Digital I/O supply voltage
	72	40				
	89	52				
	141	81				
SUPPLY GROUND CORE						
V _{SS}	15	9	GND			Core supply ground reference
	37	20				
	55	27				
	94	53				
	125	70				
	132	74				
SUPPLY GROUND DIGITAL I/O						
V _{SSIO}	29	16	GND			Digital I/O supply ground reference
	71	39				
	88	51				
	142	82				

B1M Device-Specific Information

Memory

Figure 2 shows the memory map of the B1M device.



- A. Memory addresses are configurable by the system (SYS) module within the range of 0x0000_0000 to 0xFFE0_0000.
- B. The CPU registers are not part of the memory map.

Figure 2. SM470R1B1M Memory Map

memory selects

Memory selects allow the user to address memory arrays (i.e., flash, RAM, and HET RAM) at user-defined addresses. Each memory select has its own set (low and high) of memory base address registers (MFBABRx and MFBALRx) that, together, define the array's starting (base) address, block size, and protection.

The base address of each memory select is configurable to any memory address boundary that is a multiple of the decoded block size. For more information on how to control and configure these memory select registers, see the bus structure and memory sections of the *TMS470R1x System Module Reference Guide* (literature number SPNU189).

For the memory selection assignments and the memory selected, see [Table 5](#).

Table 5. SM470R1B1M Memory Selection Assignment

MEMORY SELECT	MEMORY SELECTED (ALL INTERNAL)	MEMORY SIZE ⁽¹⁾	MPU	MSM	MEMORY BASE ADDRESS REGISTER	STATIC MEM CTL REGISTER
0 (fine)	FLASH/ROM	1 M	NO	YES	MFBABR0 and MFBALR0	
1 (fine)	FLASH/ROM		NO	YES	MFBABR1 and MFBALR1	
2 (fine)	RAM	64 K ⁽²⁾	YES	YES	MFBABR2 and MFBALR2	
3 (fine)	RAM		YES	YES	MFBABR3 and MFBALR3	
4 (fine)	HET RAM	1 K	NO	NO	MFBABR4 and MFBALR4	SMCR1
5 (coarse)	$\overline{CS}[5]/GIOC[3]$	512K x 8 (512KB) 256K x 16 (512KB)	NO	NO	MCBABR2 and MCBALR2	SMCR5
6 (coarse)	$\overline{CS}[6]/GIOC[4]$	512K x 8 (512KB) 256K x 16 (512KB)	NO	NO	MCBABR3 and MCBALR3	SMCR6

(1) x8 refers to size of memory in 8-bits; x16 refers to size of memory in 16-bits.

(2) The starting addresses for both RAM memory-select signals cannot be offset from each other by a multiple of the user-defined block size in the memory-base address register.

JTAG security module

The B1M device includes a JTAG security module to provide maximum security to the memory contents. The visible unlock code can be in the OTP sector or in the first bank of the user-programmable memory. For the B1M, the visible unlock code is in the OTP sector at address 0x0000_01F8.

memory security module

The B1M device also includes a memory security module (MSM) to provide additional security and flexibility to the memory contents' protection. The password for unlocking the MSM is located in the four words just before the flash protection keys.

RAM

The B1M device contains 64K-bytes of internal static RAM configurable by the SYS module to be addressed within the range of 0x0000_0000 to 0xFFE0_0000. This B1M RAM is implemented in one 64K-byte array selected by two memory-select signals. This B1M configuration imposes an additional constraint on the memory map for RAM; the starting addresses for both RAM memory selects cannot be offset from each other by the multiples of the size of the physical RAM (i.e., 64K bytes for the B1M device). The B1M RAM is addressed through memory selects 2 and 3.

The RAM can be protected by the memory protection unit (MPU) portion of the SYS module, allowing the user finer blocks of memory protection than is allowed by the memory selects. The MPU is ideal for protecting an operating system while allowing access to the current task. For more detailed information on the MPU portion of the SYS module and memory protection, see the memory section of the *TMS470R1x System Module Reference Guide* (literature number SPNU189).

F05 Flash

The F05 flash memory is a nonvolatile electrically erasable and programmable memory implemented with a 32-bit-wide data bus interface. The F05 flash has an external state machine for programming and erase functions. See the *Flash read* and *Flash program and erase* sections.

flash protection keys

The B1M device provides flash protection keys. These four 32-bit protection keys prevent program/erase/compaction operations from occurring until after the four protection keys have been matched by the CPU loading the correct user keys into the FMPKEY control register. The protection keys on the B1M are located in the last 4 words of the first 64K sector.

flash read

The B1M flash memory is configurable by the SYS module to be addressed within the range of 0x0000_0000 to 0xFFE0_0000. The flash is addressed through memory selects 0 and 1.

NOTE

The flash external pump voltage (V_{CCP}) is required for all operations (program, erase, and read).

flash pipeline mode

When in pipeline mode, the flash operates with a system clock frequency of up to 60 MHz (versus a system clock frequency of 30 MHz in normal mode). Flash in pipeline mode is capable of accessing 64-bit words and provides two 32-bit pipelined words to the CPU. Also, in pipeline mode the flash can be read with no wait states when memory addresses are contiguous (after the initial 1- or 2-wait-state reads).

NOTE

After a system reset, pipeline mode is disabled (ENPIPE bit [FMREGOPT.0] is a 0). In other words, the B1M device powers up and comes out of reset in non-pipeline mode. Furthermore, setting the flash configuration mode bit (GBLCTRL.4) will override pipeline mode.

flash program and erase

The B1M device flash contains two 512K-byte memory arrays (or banks), for a total of 1M-byte of flash, and consists of sixteen sectors. These sixteen sectors are sized as follows:

SECTOR NO.	SEGMENT	LOW ADDRESS	HIGH ADDRESS	MEMORY ARRAYS (OR BANKS)
OTP	2K Bytes	0x0000_0000	0x0000_007FF	BANK0 (512K Bytes)
0	64K Bytes	0x0000_0000	0x0000_FFFF	
1	64K Bytes	0x0001_0000	0x0001_FFFF	
2	64K Bytes	0x0002_0000	0x0002_FFFF	
3	64K Bytes	0x0003_0000	0x0003_FFFF	
4	64K Bytes	0x0004_0000	0x0004_FFFF	
5	64K Bytes	0x0005_0000	0x0005_FFFF	
6	64K Bytes	0x0006_0000	0x0006_FFFF	
7	64K Bytes	0x0007_0000	0x0007_FFFF	
0	64K Bytes	0x0008_0000	0x0008_FFFF	BANK1 (512K Bytes)
1	64K Bytes	0x0009_0000	0x0009_FFFF	
2	64K Bytes	0x000A_0000	0x000A_FFFF	
3	64K Bytes	0x000B_0000	0x000B_FFFF	
4	64K Bytes	0x000C_0000	0x000C_FFFF	
5	64K Bytes	0x000D_0000	0x000D_FFFF	
6	64K Bytes	0x000E_0000	0x000E_FFFF	
7	64K Bytes	0x000F_0000	0x000F_FFFF	

The minimum size for an erase operation is one sector. The maximum size for a program operation is one 16-bit word.

NOTE

The flash external pump voltage (V_{CCP}) is required for all operations (program, erase, and read).

Execution can occur from one bank while programming/erasing any or all sectors of another bank. However, execution cannot occur from any sector within a bank that is being programmed or erased.

NOTE

When the OTP sector is enabled, the rest of flash memory is disabled. The OTP memory can only be read or programmed from code executed out of RAM.

HET RAM

The B1M device contains HET RAM. The HET RAM has a 64-instruction capability. The HET RAM is configurable by the SYS module to be addressed within the range of 0x0000_0000 to 0xFFE0_0000. The HET RAM is addressed through memory select 4.

peripheral selects and base addresses

The B1M device uses 10 of the 16 peripheral selects to decode the base addresses of the peripherals. These peripheral selects are fixed and transparent to the user since they are part of the decoding scheme used by the SYS module.

Control registers for the peripherals, SYS module, and flash begin at the base addresses shown in [Table 6](#).

Table 6. B1M Peripherals, System Module, and Flash Base Addresses

CONNECTING MODULE	ADDRESS RANGE		PERIPHERAL SELECTS
	BASE ADDRESS	ENDING ADDRESS	
SYSTEM	0 x FFFF_FFCC	0 x FFFF_FFFF	N/A
RESERVED	0 x FFFF_FF70	0 x FFFF_FFCB	N/A
DWD	0xFFFF_FF60	0 x FFFF_FF6F	N/A
PSA	0 x FFFF_FF40	0 x FFFF_FF5F	N/A
CIM	0 x FFFF_FF20	0 x FFFF_FF3F	N/A
RTI	0 x FFFF_FF00	0 x FFFF_FF1F	N/A
DMA	0 x FFFF_FE80	0 x FFFF_FEFF	N/A
DEC	0 x FFFF_FE00	0 x FFFF_FE7F	N/A
RESERVED	0xFFFF_FD80	0xFFFF_FDFF	N/A
MMC	0 x FFFF_FD00	0 x FFFF_FD7F	N/A
IEM	0 x FFFF_FC00	0 x FFFF_FCFE	N/A
RESERVED	0 x FFFF_Fb00	0 x FFFF_FBFF	N/A
RESERVED	0 x FFFF_Fa00	0 x FFFF_FAFF	N/A
DMA CMD BUFFER	0 x FFFF_F800	0 x FFFF_F9FF	N/A
MSM	0xFFFF_F700	0xFFFF_F7FF	N/A
RESERVED	0xFFFF_F600	0xFFFF_F6FF	N/A
RESERVED	0 x FFF7_FD00	0xFFFF_FFFF	PS[0]
HET	0xFFFF_FC00	0xFFFF_FCFE	
RESERVED	0xFFFF_F900	0xFFFF_FBFF	PS[1]
SPI1	0xFFFF_F800	0xFFFF_F8FF	
RESERVED	0xFFFF_F700	0xFFFF_F7FF	PS[2]
SCI3	0xFFFF_F600	0xFFFF_F6FF	
SCI2	0xFFFF_F500	0xFFFF_F5FF	
SCI1	0xFFFF_F400	0xFFFF_F4FF	
RESERVED	0xFFFF_F100	0xFFFF_F3FF	PS[3]
MibADC	0xFFFF_F000	0xFFFF_F0FF	
ECP	0xFFFF_EF00	0xFFFF_EFFF	PS[4]
RESERVED	0xFFFF_EE00	0xFFFF_EEFF	
EBM	0xFFFF_ED00	0xFFFF_EDFF	
GIO	0xFFFF_EC00	0xFFFF_ECFE	
HECC2	0xFFFF_EB00	0xFFFF_EBFF	PS[5]
	0xFFFF_EA00	0xFFFF_EAFF	
HECC1	0xFFFF_E900	0xFFFF_E9FF	
	0xFFFF_E800	0xFFFF_E8FF	
HECC2 RAM	0xFFFF_E700	0xFFFF_E7FF	PS[6]
	0xFFFF_E600	0xFFFF_E6FF	
HECC1 RAM	0xFFFF_E500	0xFFFF_E5FF	
	0xFFFF_E400	0xFFFF_E4FF	
RESERVED	0xFFFF_E100	0xFFFF_E3FF	PS[7]
SCC	0xFFFF_E000	0xFFFF_E0FF	

Table 6. B1M Peripherals, System Module, and Flash Base Addresses (continued)

CONNECTING MODULE	ADDRESS RANGE		PERIPHERAL SELECTS
	BASE ADDRESS	ENDING ADDRESS	
RESERVED	0xFFFF7_DD00	0xFFFF7_DFFF	PS[8]
SCC RAM	0xFFFF7_DC00	0xFFFF7_DCFE	
I2C4	0xFFFF7_DB00	0xFFFF7_DBFF	PS[9]
I2C3	0xFFFF7_DA00	0xFFFF7_DAFF	
I2C2	0xFFFF7_D900	0xFFFF7_D9FF	
I2C1	0xFFFF7_D800	0xFFFF7_D8FF	
RESERVED	0xFFFF7_D600	0xFFFF7_D7FF	PS[10]
I2C5	0xFFFF7_D500	0xFFFF7_D5FF	
SPI2	0xFFFF7_D400	0xFFFF7_D4FF	
RESERVED	0xFFFF7_CC00	0xFFFF7_D3FF	PS[11] – PS[12]
RESERVED	0xFFFF7_C800	0xFFFF7_CBFF	PS[13]
RESERVED	0xFFFF7_C000	0xFFFF7_C7FF	PS[14] – PS[15]
RESERVED	0xFFFF0_0000	0xFFFF7_BFFF	N/A
FLASH CONTROL REGISTERS	0xFFE8_8000	0xFFE8_BFFF	N/A
RESERVED	0xFFFF8_4024	0xFFFF8_7FFF	N/A
MPU CONTROL REGISTERS	0xFFE8_4000	0xFFE8_4023	N/A
RESERVED	0xFFFF8_0000	0xFFFF8_3FFF	N/A

direct-memory access (DMA)

The direct-memory access (DMA) controller transfers data to and from any specified location in the B1M memory map (except for restricted memory locations like the system control registers area). The DMA manages up to 16 channels, and supports data transfer for both on-chip and off-chip memories and peripherals. The DMA controller is connected to both the CPU and peripheral buses, enabling these data transfers to occur in parallel with CPU activity and thus maximizing overall system performance.

Although the DMA controller has two possible configurations, for the B1M device, the DMA controller configuration is 32 control packets and 16 channels.

For the B1M DMA request hardwired configuration, see [Table 7](#).

Table 7. DMA Request Lines Connections⁽¹⁾

MODULES	DMA REQUEST INTERRUPT SOURCES		DMA CHANNEL
EBM	Expansion Bus DMA request	EBDMAREQ[0]	DMAREQ[0]
SPI1/I2C4	SPI1 end-receive/I2C4 read	SPI1DMA0/I2C4DMA0	DMAREQ[1]
SPI1/I2C4	SPI1 end-transmit/I2C4 write	SPI1DMA1/I2C4DMA1	DMAREQ[2]
MibADC/I2C1	ADC EV/I2C1 read	MibADCMA0/I2C1DMA0	DMAREQ[3]
MibADC/SCI1/I2C5	ADC G1/SCI1 end-receive/I2C5 read	MibADCMA1/SCI1DMA0/I2C5DMA0	DMAREQ[4]
MibADC/SCI1/I2C5	ADC G2/SCI1 end-transmit/I2C5 write	MibADCMA2/SCI1DMA1/I2C5DMA1	DMAREQ[5]
I2C1	I2C1 write	I2C1DMA1	DMAREQ[6]
SCI3/SPI2	SCI3 end-receive/SPI2 end-receive	SCI3DMA0/SPI2DMA0	DMAREQ[7]
SCI3/SPI2	SCI3 end-transmit/SPI2 end-transmit	SCI3DMA0/SPI2DMA1	DMAREQ[8]
I2C2	I2C2 read end-receive	I2C2DMA0	DMAREQ[9]
I2C2	I2C2 write end-transmit	I2C2DMA1	DMAREQ[10]
I2C3	I2C3 read	I2C3DMA0	DMAREQ[11]
I2C3	I2C3 write	I2C3DMA1	DMAREQ[12]
Reserved			DMAREQ[13]
SCI2	SCI2 end-receive	SCI2DMA0	DMAREQ[14]
SCI2	SCI2 end-transmit	SCI2DMA1	DMAREQ[15]

(1) For DMA channels with more than one assigned request source, *only one* of the sources listed can be the DMA request generator in a given application. The device has software control to ensure that there are no conflicts between requesting modules.

Each channel has two control packets attached to it, allowing the DMA to continuously load RAM and generate periodic interrupts so that the data can be read by the CPU. The control packets allow for the interrupt enable, and the channels determine the priority level of the interrupt.

DMA transfers occur in one of two modes:

- Non-request mode (used when transferring from memory to memory)
- Request mode (used when transferring from memory to peripheral)

For more detailed functional information on the DMA controller, see the *TMS470R1x Direct Memory Access (DMA) Controller Reference Guide* (literature number SPNU194).

interrupt priority (IEM to CIM)

Interrupt requests originating from the B1M peripheral modules (i.e., SPI1 or SPI2; SCI1 or SCI2; RTI; etc.) are assigned to channels within the 48-channel interrupt expansion module (IEM) where, via programmable register mapping, these channels are then mapped to the 32-channel central interrupt manager (CIM) portion of the SYS module.

Programming multiple interrupt sources in the IEM to the same CIM channel effectively shares the CIM channel between sources.

The CIM request channels are maskable so that individual channels can be selectively disabled. All interrupt requests can be programmed in the CIM to be of either type:

- Fast interrupt request (FIQ)
- Normal interrupt request (IRQ)

The CIM prioritizes interrupts. The precedences of request channels decrease with ascending channel order in the CIM (0 [highest] and 31 [lowest] priority). For IEM-to-CIM default mapping, channel priorities, and their associated modules, see [Table 8](#).

Table 8. Interrupt Priority (IEM and CIM)

MODULES	INTERRUPT SOURCES	DEFAULT CIM INTERRUPT LEVEL/CHANNEL	IEM CHANNEL
SPI1	SPI1 end-transfer/overflow	0	0
RTI	COMP2 interrupt	1	1
RTI	COMP1 interrupt	2	2
RTI	TAP interrupt	3	3
SPI2	SPI2 end-transfer/overflow	4	4
GIO	GIO interrupt A	5	5
Reserved		6	6
HET	HET interrupt 1	7	7
I2C1	I2C1 interrupt	8	8
SCI1/SCI2	SCI1 or SCI2 error interrupt	9	9
SCI1	SCI1 receive interrupt	10	10
Reserved		11	11
I2C2	I2C2 interrupt	12	12
HECC1	HECC1 interrupt A	13	13
SCC	SCC interrupt A	14	14
Reserved		15	15
MibADC	MibADC end event conversion	16	16
SCI2	SCI2 receive interrupt	17	17
DMA	DMA interrupt 0	18	18
I2C3	I2C3 interrupt	19	19
SCI1	SCI1 transmit interrupt	20	20
System	SW interrupt (SSI)	21	21
Reserved		22	22
HET	HET interrupt 2	23	23
HECC1	HECC1 interrupt B	24	24
SCC	SCC interrupt B	25	25
SCI2	SCI2 transmit interrupt	26	26
MibADC	MibADC end Group 1 conversion	27	27
DMA	DMA Interrupt 1	28	28
GIO	GIO interrupt B	29	29
MibADC	MibADC end Group 2 conversion	30	30
SCI3	SCI3 error interrupt	31	31

Table 8. Interrupt Priority (IEM and CIM) (continued)

MODULES	INTERRUPT SOURCES	DEFAULT CIM INTERRUPT LEVEL/CHANNEL	IEM CHANNEL
Reserved		31	32–37
HECC2	HECC2 interrupt A	31	38
HECC2	HECC2 interrupt B	31	39
SCI3	SCI3 receive interrupt	31	40
SCI3	SCI3 transmit interrupt	31	41
I2C4	I2C4 interrupt	31	42
I2C5	I2C5 interrupt	31	43
Reserved		31	44–47

For more detailed functional information on the IEM, see the *TMS470R1x Interrupt Expansion Module (IEM) Reference Guide* (literature number SPNU211). For more detailed functional information on the CIM, see the *TMS470R1x System Module Reference Guide* (literature number SPNU189).

expansion bus module (EBM)

The expansion bus module (EBM) is a standalone module used to bond out both general-purpose input/output pins and expansion bus interface pins. This module supports the multiplexing of the GIO and the expansion bus interface functions. The module also supports 8- and 16- bit expansion bus memory interface mappings as well as mapping of the following expansion bus signals:

- 27-bit address bus (EBADDR[26:0] for x8, 19-bit address bus (EBADDR[18:0] for x16)
- 8- or 16-bit data bus (EBDATA[7:0] or EBDATA[15:0])
- 2 write strobes (EBWR[1:0])
- 2 memory chip selects (EBCS[6:5])
- 1 output enable (EBOE)
- 1 external hold signal for interfacing to slow memories (EBHOLD)
- 1 DMA request line (EBDMAREQ[0])

Table 9 shows the multiplexing of I/O signals with the expansion bus interface signals. The mapping of these pins varies depending on the memory mode.

Table 9. Expansion Bus Mux Mapping⁽¹⁾

GIO	EXPANSION BUS MODULE PINS	
	x8 ⁽²⁾	x16 ⁽²⁾
GIOB[0]	EBDMAREQ[0]	EBDMAREQ[0]
GIOC[0]	$\overline{\text{EBOE}}$	$\overline{\text{EBOE}}$
GIOC[2:1]	$\overline{\text{EBWR}}[1:0]$	$\overline{\text{EBWR}}[1:0]$
GIOC[4:3]	$\overline{\text{EBCS}}[6:5]$	$\overline{\text{EBCS}}[6:5]$
GIOD[5:0]	EBADDR[5:0]	EBADDR[5:0]
GIOE[7:0]	EBDATA[7:0]	EBDATA[7:0]
GIOF[7:0]	EBADDR[13:6]	EBDATA[15:8]
GIOG[7:0]	EBADDR[21:14]	EBADDR[13:6]
GIOH[5]	$\overline{\text{EBHOLD}}$	$\overline{\text{EBHOLD}}$
I2C5SDA	EBADDR[26]	EBADDR[18]
I2C5SCL	EBADDR[25]	EBADDR[17]
I2C4SCL	EBADDR[24]	EBADDR[16]
I2C4SDA	EBADDR[23]	EBADDR[15]
GIOH[0]	EBADDR[22]	EBADDR[14]

(1) For more detailed information, see the *TMS470R1x Expansion Bus Module (EBM) Reference Guide* (literature number SPNU222) and the *TMS470R1x General Purpose Input/Output Reference Guide* (literature number SPNU192).

(2) X8 refers to size of memory in 8-bits; X16 refers to size of memory in 16-bits.

Table 10 lists the names of the expansion bus interface signals and their functions.

Table 10. Expansion Bus Pins

PIN	DESCRIPTION
EBDMAREQ	Expansion bus DMA request
$\overline{\text{EBOE}}$	Expansion bus pin enable
$\overline{\text{EBWR}}$	Expansion bus write strobe EBWR[1] controls EBDATA[15:8] and EBWR[0] controls EBDATA[7:0]
$\overline{\text{EBCS}}$	Expansion bus chip select
EBADDR	Expansion bus address pins
EBDATA	Expansion bus data pins
$\overline{\text{EBHOLD}}$	Expansion bus hold: An external device may assert this signal to add wait states to an expansion bus transaction.

MibADC

The multi-buffered analog-to-digital converter (MibADC) accepts an analog signal and converts the signal to a 10-bit digital value.

The B1M MibADC module can function in two modes: compatibility mode, where its programmer's model is compatible with the SM470R1x ADC module and its digital results are stored in digital result registers; or in buffered mode, where the digital result registers are replaced with three FIFO buffers, one for each conversion group [event, group1 (G1), and group2 (G2)]. In buffered mode, the MibADC buffers can be serviced by interrupts or by the DMA.

MibADC event trigger enhancements

The MibADC includes two major enhancements over the event-triggering capability of the SM470R1x ADC.

- Both group 1 and the event group can be configured for event-triggered operation, providing up to two event-triggered groups.
- The trigger source and polarity can be selected individually for both group1 and the event group from the options identified in [Table 11](#).

Table 11. MibADC Event Hookup Configuration

EVENT #	SOURCE SELECT BITS FOR G1 OR EVENT (G1SRC[1:0] OR EVSRC[1:0])	SIGNAL PIN NAME
EVENT1	00	ADEVT
EVENT2	01	HET18
EVENT3	10	Reserved
EVENT4	11	Reserved

For group1, these event-triggered selections are configured via the group 1 source select bits (G1SRC[1:0]) in the AD event source register (ADEVTSRC[5:4]). For the event group, these event-triggered selections are configured via the event group source select bits (EVSRC[1:0]) in the AD event source register (ADEVTSRC[1:0]).

For more detailed functional information on the MibADC, see the *TMS470R1x Multi-Buffered Analog-to-Digital Converter (MibADC) Reference Guide* (literature number SPNU206).

JTAG Interface

There are two main test access ports (TAPs) on the device:

- SM470R1x CPU TAP
- Device TAP for factory test

Some of the JTAG pins are shared among these two TAPs. The hookup is illustrated in [Figure 3](#).

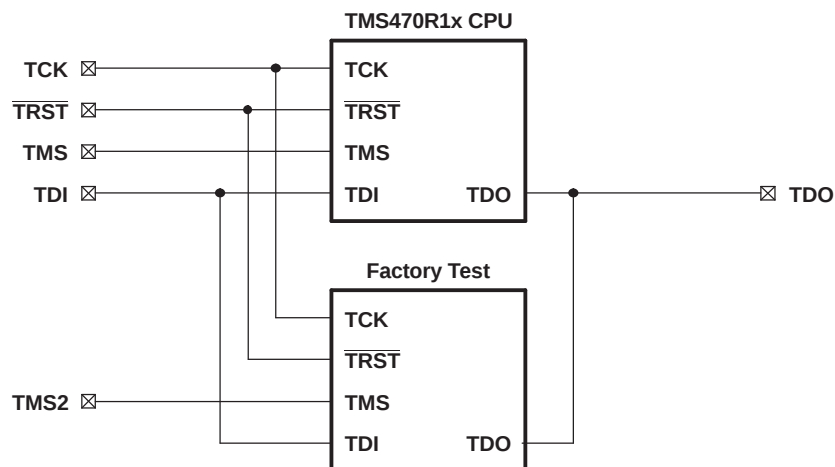


Figure 3. JTAG Interface

documentation support

Extensive documentation supports all of the SM470 microcontroller family generation of devices. The types of documentation available include data sheets with design specifications; complete user's guides for all devices and development support tools; and hardware and software applications. Useful reference documentation includes:

- Bulletin
 - *TMS470 Microcontroller Family Product Bulletin* (literature number SPNB086)
- User's Guides
 - *TMS470R1x System Module Reference Guide* (literature number SPNU189)
 - *TMS470R1x General Purpose Input/Output (GPIO) Reference Guide* (literature number SPNU192)
 - *TMS470R1x Direct Memory Access (DMA) Controller Reference Guide* (literature number SPNU194)
 - *TMS470R1x Direct Memory Access (DMA) Controller Reference Guide* (literature number SPNU194)
 - *TMS470R1x Serial Peripheral Interface (SPI) Reference Guide* (literature number SPNU195)
 - *TMS470R1x Serial Communication Interface (SCI) Reference Guide* (literature number SPNU196)
 - *TMS470R1x Controller Area Network (CAN) Reference Guide* (literature number SPNU197)
 - *TMS470R1x High End Timer (HET) Reference Guide* (literature number SPNU199)
 - *TMS470R1x External Clock Prescale (ECP) Reference Guide* (literature number SPNU202)
 - *TMS470R1x MultiBuffered Analog to Digital (MibADC) Reference Guide* (literature number SPNU206)
 - *TMS470R1x Zero Pin Phase Locked Loop (ZPLL) Clock Module Reference Guide* (literature number SPNU212)
 - *TMS470R1x Digital Watchdog Timer Reference Guide* (literature number SPNU244)
 - *TMS470R1x Interrupt Expansion Module (IEM) Reference Guide* (literature number SPNU211)
 - *TMS470R1x Class II Serial Interface B (C2SIB) Reference Guide* (literature number SPNU214)
 - *TMS470R1x Class II Serial Interface A (C2SIA) Reference Guide* (literature number SPNU218)
 - *TMS470R1x Expansion Bus Module (EBM) Reference Guide* (literature number SPNU222)
 - *TMS470R1x Inter-Integrated Circuit (I2C) Reference Guide* (literature number SPNU223)
 - *TMS470R1x JTAG Security Module (JSM) Reference Guide* (literature number SPNU245)
 - *TMS470R1x Memory Security Module (MSM) Reference Guide* (literature number SPNU246)
 - *TMS470 Peripherals Overview Reference Guide* (literature number SPNU248)
- Errata Sheet
 - *TMS470R1B1M TMS470 Microcontrollers Silicon Errata* (literature number SPNZ139)

Device and Development-Support Tool Nomenclature

To designate the stages in the product development cycle, TI assigns prefixes to the part numbers of all DSP devices and support tools. Each DSP commercial family member has one of three prefixes: TMX, TMP, or TMS (e.g., **TMS470R1B1M**). Texas Instruments recommends two of three possible prefix designators for its support tools: TMDX and TMDS. These prefixes represent evolutionary stages of product development from engineering prototypes (TMX/TMDX) through fully qualified production devices/tools (TMS/TMDS).

Device development evolutionary flow:

TMX Experimental device that is not necessarily representative of the final device's electrical specifications

TMP Final silicon die that conforms to the device's electrical specifications but has not completed quality and reliability verification

SM Fully qualified production device

Support tool development evolutionary flow:

TMDX Development-support product that has not yet completed Texas Instruments internal qualification testing.

TMDS Fully qualified development-support product

TMX and TMP devices and TMDX development-support tools are shipped against the following disclaimer:

"Developmental product is intended for internal evaluation purposes."

TMS devices and TMDS development-support tools have been characterized fully, and the quality and reliability of the device have been demonstrated fully. TI's standard warranty applies.

Predictions show that prototype devices (TMX or TMP) have a greater failure rate than the standard production devices. Texas Instruments recommends that these devices not be used in any production system because their expected end-use failure rate still is undefined. Only qualified production devices are to be used.

[Figure 4](#) illustrates the numbering and symbol nomenclature for the SM470R1x family.

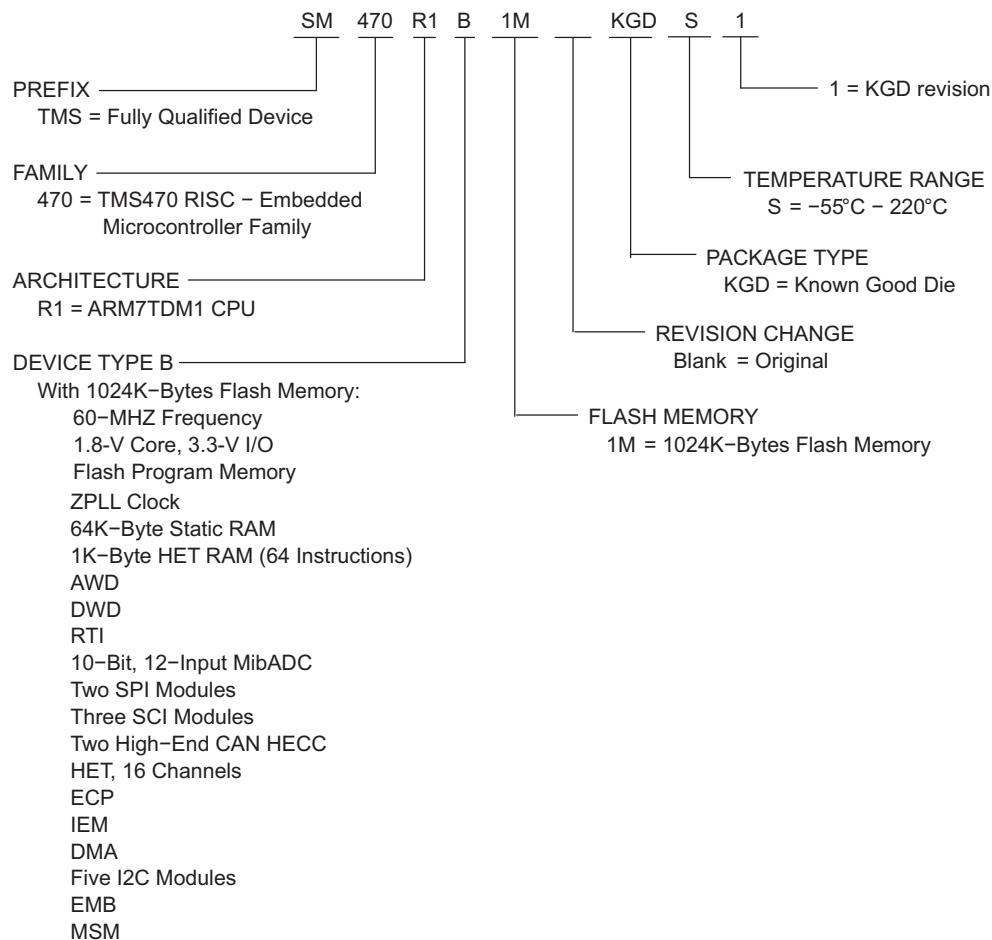


Figure 4. SM470R1x Family Nomenclature

device identification code register

The device identification code register identifies the silicon version, the technology family (TF), a ROM or flash device, and an assigned device-specific part number (see [Table 12](#)). The B1M device identification code register value is 0xnA5F.

Figure 5. SM470 Device ID Bit Allocation Register [offset = 0xFFFF_FFF0h]

31				Reserved												16		
15				12		11	10	9			3			2	1	0		
VERSION				TF	R/F	PART NUMBER									1	1	1	
R-K				R-K	R-K	R-K									R-1	R-1	R-1	

LEGEND:

For bits 3-15: R = Read only, -K = Value constant after $\overline{\text{RESET}}$.

For bits 0-2: R = Read only, -1 = Value after RESET.

Table 12. SM470 Device ID Bit Allocation Register Field Descriptions

Bit	Field	Value	Description
31-16	Reserved		Reads are undefined and writes have no effect.
15-12	VERSION		Silicon version (revision) bits These bits identify the silicon version of the device.
11	TF	0 1	Technology family bit This bit distinguishes the technology family core power supply: 3.3 V for F10/C10 devices 1.8 V for F05/C05 devices
10	R/F	0 1	ROM/flash bit This bit distinguishes between ROM and flash devices: Flash device ROM device
9-3	PART NUMBER		Device-specific part number bits These bits identify the assigned device-specific part number. The assigned device-specific part number for the B1M device is 1001011.
2-0	1		Mandatory High Bits 2, 1, and 0 are tied high by default.

DEVICE ELECTRICAL SPECIFICATIONS AND TIMING PARAMETERS

Absolute Maximum Ratings

over operating free-air temperature range, A version (unless otherwise noted)⁽¹⁾

Supply voltage range:	V_{CC} ⁽²⁾	–0.3 V to 2.5 V
Supply voltage range:	V_{CCIO} , V_{CCAD} , V_{CCP} (flash pump) ⁽²⁾	–0.3 V to 4.1 V
Input voltage range:	All 5 V tolerant input pins	–0.3 V to 6.0 V
	All other input pins	–0.3 V to 4.1 V
Input clamp current:	I_{IK} ($V_I < 0$ or $V_I > V_{CCIO}$) All pins except ADIN[0:11], $\overline{PORRST}$, $\overline{TRST}$, TEST, and TCK	±20 mA
	I_{IK} ($V_I < 0$ or $V_I > V_{CCAD}$) ADIN[0:11]	±10 mA
Operating free-air temperature range, T_A :	A version	–55°C to 220°C
Storage temperature range, T_{stg} :		–55°C to 220°C

- (1) Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
- (2) All voltage values are with respect to their associated grounds.

Device Recommended Operating Conditions⁽¹⁾

			MIN	NOM	MAX	UNIT
V_{CC}	Digital logic supply voltage (Core)	SYSCLK = 48 MHz (pipeline mode enabled)	1.71		2.05	V
		SYSCLK = 60 MHz (pipeline mode enabled)	1.81		2.05	
V_{CCIO}	Digital logic supply voltage (I/O)		3		3.6	V
V_{CCAD}	ADC supply voltage		3		3.6	V
V_{CCP}	Flash pump supply voltage		3		3.6	V
V_{SS}	Digital logic supply ground			0		V
V_{SSAD}	ADC supply ground ⁽¹⁾		–0.1		0.1	V
T_A	Operating free-air temperature		–55		220	°C

- (1) All voltages are with respect to V_{SS} , except V_{CCAD} , which is with respect to V_{SSAD} .

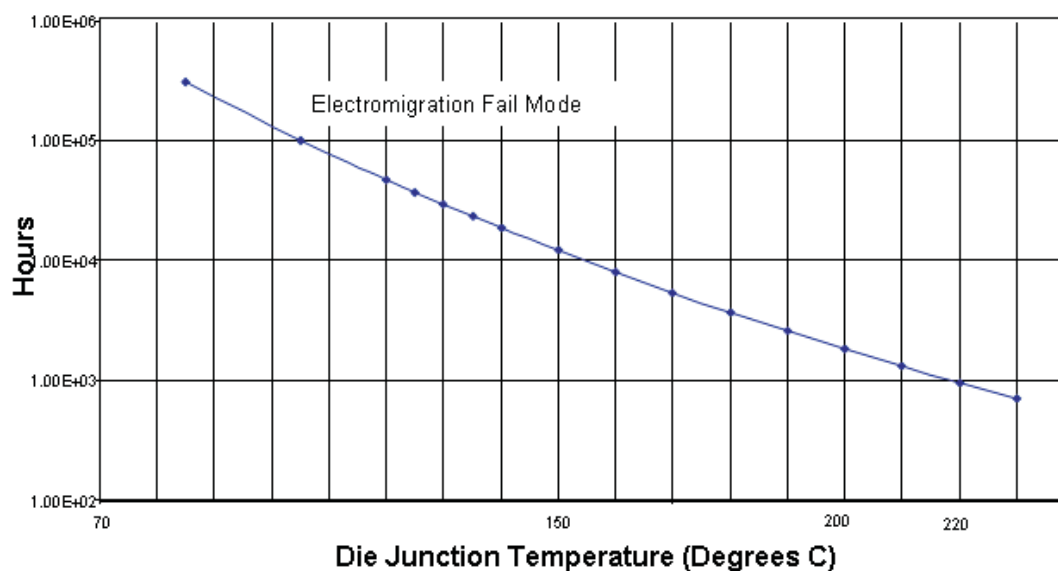


Figure 6. SM470R1B1M-HT Life Expectancy Curve

Notes:

1. See data sheet for absolute maximum and minimum recommended operating conditions.
2. Silicon operating life design goal is 10 years at 105°C junction temperature (does not include package interconnect life).

ELECTRICAL CHARACTERISTICS

Minimum and maximum parameters are characterized for operation at $T_A = 220^\circ\text{C}$ unless otherwise noted, but may not be production tested at that temperature. Production test limits with statistical guardbands are used to ensure high temperature performance. ⁽¹⁾

PARAMETER			TEST CONDITIONS	MIN	TYP ⁽²⁾	MAX	UNIT
V_{hys}	Input hysteresis			0.15			V
V_{IL}	Low-level input voltage	All inputs ⁽³⁾		-0.3		0.8	V
V_{IH}	High-level input voltage	All inputs		2		$V_{CCIO} + 0.3$	V
	Input threshold voltage	AWD only ⁽⁴⁾		1.35		1.8	
		OSCIN with digital input only		$0.7 V_{CC}$		$V_{CC} + 0.3$	
V_{OL}	Low-level output voltage ⁽⁵⁾		$I_{OL} = I_{OL\ MAX}$			$0.2 V_{CCIO}$	V
			$I_{OL} = 50\ \mu\text{A}$			0.2	
V_{OH}	High-level output voltage ⁽⁵⁾		$I_{OH} = I_{OH\ MIN}$	$0.8 V_{CCIO}$			V
			$I_{OH} = 50\ \mu\text{A}$	$V_{CCIO} - 0.2$			
I_{IC}	Input clamp current (I/O pins) ⁽⁶⁾		$V_I < V_{SSIO} - 0.3$ or $V_I > V_{CCIO} + 0.3$	-2		2	mA
I_I	Input current (3.3 V input pins)	I_{IL} Pulldown	$V_I = V_{SS}$	-1		1	μA
		I_{IH} Pulldown	$V_I = V_{CCIO}$	5		100	
		I_{IL} Pullup	$V_I = V_{SS}$	-100		-5	
		I_{IH} Pullup	$V_I = V_{CCIO}$	-1		1	
		All other pins	No pullup or pulldown	-1		1	
	Input current (5 V tolerant input pins)		$V_I = V_{SS}$	-1		1	μA
			$V_I = V_{CCIO}$	1		5	
			$V_I = 5\ \text{V}$	5		25	
			$V_I = 5.5\ \text{V}$	25		50	
I_{OL}	Low-level output current	CLKOUT, AWD, TDI, TDO, TMS, TMS2	$V_{OL} = V_{OL\ MAX}$			8	mA
		$\overline{\text{RST}}$				4	
		All other 3.3 V I/O ⁽⁷⁾				2	
		5 V tolerant				4	
I_{OH}	High-level output current	CLKOUT, TDI, TDO, TMS, TMS2	$V_{OH} = V_{OH\ MIN}$	-8			mA
		$\overline{\text{RST}}$		-4			
		All other 3.3 V I/O ⁽⁷⁾		-2			
		5 V tolerant		-4			

(1) Source currents (out of the device) are negative while sink currents (into the device) are positive.

(2) The typical values indicated in this table are the expected values during operation under normal operating conditions: nominal V_{CC} , V_{CCIO} , or V_{CCAD} , room temperature.

(3) This does not apply to the $\overline{\text{PORRST}}$ pin. For $\overline{\text{PORRST}}$ exceptions, see the $\overline{\text{RST}}$ and $\overline{\text{PORRST}}$ Timings section.

(4) These values help to determine the external RC network circuit. For more details, see the *TMS470R1x System Module Reference Guide* (literature number SPNU189).

(5) V_{OL} and V_{OH} are linear with respect to the amount of load current (I_{OL}/I_{OH}) applied.

(6) Parameter does not apply to input-only or output-only pins.

(7) Some of the 2 mA buffers on this device are zero-dominant buffers, as indicated by a -z in the Output Current column of the Terminal Functions table. If two of these buffers are shorted together and one is outputting a low level and the other is outputting a high level, the resulting value will always be low.

ELECTRICAL CHARACTERISTICS (continued)

Minimum and maximum parameters are characterized for operation at $T_A = 220^\circ\text{C}$ unless otherwise noted, but may not be production tested at that temperature. Production test limits with statistical guardbands are used to ensure high temperature performance. ⁽¹⁾

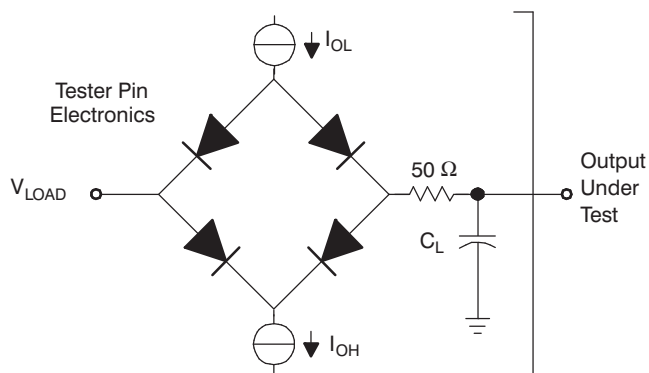
PARAMETER		TEST CONDITIONS	MIN	TYP ⁽²⁾	MAX	UNIT
I_{CC}	V_{CC} Digital supply current (operating mode)	SYSCLK = 48 MHz, ICLK = 24 MHz, $V_{CC} = 2.05\text{ V}$			110	mA
		SYSCLK = 60 MHz, ICLK = 30 MHz, $V_{CC} = 2.05\text{ V}$			125	mA
	V_{CC} Digital supply current (standby mode) ^{(8) (9)}	OSCIN = 5 MHz, $V_{CC} = 2.05\text{ V}$			1.30	mA
	V_{CC} Digital supply current (halt mode) ^{(8) (9)}	All frequencies, $V_{CC} = 2.05\text{ V}$			700	μA
I_{CCIO}	V_{CCIO} Digital supply current (operating mode)	No DC load, $V_{CCIO} = 3.6\text{ V}$ ⁽¹⁰⁾			20	mA
	V_{CCIO} Digital supply current (standby mode) ⁽⁹⁾	No DC load, $V_{CCIO} = 3.6\text{ V}$ ⁽¹⁰⁾			250	μA
	V_{CCIO} Digital supply current (halt mode) ⁽⁹⁾	No DC load, $V_{CCIO} = 3.6\text{ V}$ ⁽¹⁰⁾			225	μA
I_{CCAD}	V_{CCAD} supply current (operating mode)	All frequencies, $V_{CCAD} = 3.6\text{ V}$			15	mA
	V_{CCAD} supply current (standby mode)	All frequencies, $V_{CCAD} = 3.6\text{ V}$			10	μA
	V_{CCAD} supply current (halt mode)	All frequencies, $V_{CCAD} = 3.6\text{ V}$			10	μA
I_{CCP}	V_{CCP} pump supply current	SYSCLK = 48 MHz, $V_{CCP} = 3.6\text{ V}$ read operation			45	mA
		SYSCLK = 60 MHz, $V_{CCP} = 3.6\text{ V}$ read operation			55	mA
		$V_{CCP} = 3.6\text{ V}$ program and erase			70	mA
		$V_{CCP} = 3.6\text{ V}$ standby mode operation ⁽⁸⁾			10	μA
		$V_{CCP} = 3.6\text{ V}$ halt mode operation ⁽⁸⁾			10	μA
C_I	Input capacitance			2		pF
C_O	Output capacitance			3		pF

(8) For flash banks/pumps in sleep mode.

(9) For reduced power consumption in low power mode, CANSRX and CANSTX should be driven output LOW.

(10) I/O pins configured as inputs or outputs with no load. All pulldown inputs $\leq 0.2\text{ V}$. All pullup inputs $\geq V_{CCIO} - 0.2\text{ V}$.

Parameter Measurement Information



Where: I_{OL} = I_{OL} MAX for the respective pin ^(A)
 I_{OH} = I_{OH} MIN for the respective pin ^(A)
 V_{LOAD} = 1.5 V
 C_L = 150-pF typical load-circuit capacitance ^(B)

- A. For these values, see the "Electrical Characteristics over Recommended Operating Free-Air Temperature Range" table.
- B. All timing parameters measured using an external load capacitance of 150 pF unless otherwise noted.

Figure 7. Test Load Circuit

Timing Parameter Symbolology

Timing parameter symbols have been created in accordance with JEDEC Standard 100. To shorten the symbols, some of the pin names and other related terminology have been abbreviated as follows:

CM	Compaction, CMPCT	RD	Read
CO	CLKOUT	RST	Reset, $\overline{\text{RST}}$
ER	Erase	RX	SCInRX
ICLK	Interface clock	S	Slave mode
M	Master mode	SCC	SCInCLK
OSC, OSCI	OSCIN	SIMO	SPInSIMO
OSCO	OSCO	SOMI	SPInSOMI
P	Program, PROG	SPC	SPInCLK
R	Ready	SYS	System clock
R0	Read margin 0, RDMRGNO	TX	SCInTX
R1	Read margin 1, RDMRGN1		

Lowercase subscripts and their meanings are:

a	access time	r	rise time
c	cycle time (period)	su	setup time
d	delay time	t	transition time
f	fall time	v	valid time
h	hold time	w	pulse duration (width)

The following additional letters are used with these meanings:

H	High	X	Unknown, changing, or don't care level
L	Low	Z	High impedance
V	Valid		

External Reference Resonator/Crystal Oscillator Clock Option

The oscillator is enabled by connecting the appropriate fundamental 4–10 MHz resonator/crystal and load capacitors across the external OSCIN and OSCOUT pins as shown in Figure 8 (a). The oscillator is a single-stage inverter held in bias by an integrated bias resistor. This resistor is disabled during leakage test measurement and HALT mode. **TI strongly encourages each customer to submit samples of the device to the resonator/crystal vendors for validation.** The vendors are equipped to determine what load capacitors will best tune their resonator/crystal to the microcontroller device for optimum start-up and operation over temperature/voltage extremes. **Please note that external crystal mode is guaranteed to function only within the temperature ranges of -40°C to 150°C. Above this recommended temperature range it is strongly recommended to use an external clock signal as shown in Figure 8 (b).**

An external oscillator source can be used by connecting a 1.8-V clock signal to the OSCIN pin and leaving the OSCOUT pin unconnected (open) as shown in Figure 8b.



- A. The values of C1 and C2 should be provided by the resonator/crystal vendor.

Figure 8. Crystal/Clock Connection

ZPLL AND CLOCK SPECIFICATIONS

Timing Requirements for ZPLL Circuits Enabled or Disabled⁽¹⁾

		MIN	TYP	MAX	UNIT
$f_{(OSC)}$	Input clock frequency	4		10	MHz
$t_{c(OSC)}$	Cycle time, OSCIN	100			ns
$t_{w(OSCIL)}$	Pulse duration, OSCIN low	15			ns
$t_{w(OSCIH)}$	Pulse duration, OSCIN high	15			ns
$f_{(OSCRST)}$	OSC FAIL frequency ⁽²⁾		53		kHz

(1) Not production tested.

(2) Causes a device reset (specifically a clock reset) by setting the RST OSC FAIL (GLBCTRL.15) and the OSC FAIL flag (GLBSTAT.1) bits equal to 1. For more detailed information on these bits and device resets, see the *TMS470R1x System Module Reference Guide* (literature number SPNU189).

Switching Characteristics over Recommended Operating Conditions for Clocks^{(1)(2) (3) (4)}

PARAMETER		TEST CONDITIONS ⁽⁵⁾	MIN	MAX	UNIT
$f_{(SYS)}$	System clock frequency ⁽⁶⁾	Pipeline mode enabled		60 ⁽⁷⁾	MHz
		Pipeline mode disabled		24	MHz
$f_{(CONFIG)}$	System clock frequency - flash config mode			24	MHz
$f_{(ICLK)}$	Interface clock frequency	Pipeline mode enabled		30	MHz
		Pipeline mode disabled		24	MHz
$f_{(ECLK)}$	External clock output frequency for ECP module	Pipeline mode enabled		30	MHz
		Pipeline mode disabled		24	MHz
$t_{c(SYS)}$	Cycle time, system clock	Pipeline mode enabled	16.7		ns
		Pipeline mode disabled	41.6		ns
$t_{c(CONFIG)}$	Cycle time, system clock - flash config mode		41.6		ns
$t_{c(ICLK)}$	Cycle time, interface clock	Pipeline mode enabled	33.3		ns
		Pipeline mode disabled	41.6		ns
$t_{c(ECLK)}$	Cycle time, ECP module external clock output	Pipeline mode enabled	33.3		ns
		Pipeline mode disabled	41.6		ns

(1) Not production tested.

(2) $f_{(SYS)} = M \times f_{(OSC)}/R$, where $M = \{8\}$, $R = \{1,2,3,4,5,6,7,8\}$ when PLLDIS = 0. R is the system-clock divider determined by the CLKDIVPRE[2:0] bits in the global control register (GLBCTRL[2:0]) and M is the PLL multiplier determined by the MULT4 bit also in the GLBCTRL register (GLBCTRL.3).

$f_{(SYS)} = f_{(OSC)}/R$, where $R = \{1,2,3,4,5,6,7,8\}$ when PLLDIS = 1.

$f_{(ICLK)} = f_{(SYS)}/X$, where $X = \{1,2,3,4,5,6,7,8,9,10,11,12,13,14,15,16\}$. X is the interface clock divider ratio determined by the PCR0[4:1] bits in the SYS module.

(3) $f_{(ECLK)} = f_{(ICLK)}/N$, where $N = \{1 \text{ to } 256\}$. N is the ECP prescale value defined by the ECPCTRL[7:0] register bits in the ECP module.

(4) Only ZPLL mode is available. FM mode must not be turned on.

(5) Pipeline mode enabled or disabled is determined by the ENPIPE bit (FMREGOPT.0).

(6) Flash Vread must be set to 5 V to achieve maximum system clock frequency.

(7) Operating V_{CC} range for this system clock frequency is 1.81 to 2.05 V.

Switching Characteristics over Recommended Operating Conditions for External Clocks^{(1)(2) (3)} (4)

 (see [Figure 9](#) and [Figure 10](#))

PARAMETER		TEST CONDITIONS	MIN	MAX	UNIT
$t_{w(COL)}$	Pulse duration, CLKOUT low	SYSCLK or MCLK ⁽⁵⁾	$0.5t_{c(SYS)} - t_f$		ns
		ICLK: X is even or 1 ⁽⁶⁾	$0.5t_{c(ICLK)} - t_f$		
		ICLK: X is odd and not 1 ⁽⁶⁾	$0.5t_{c(ICLK)} + 0.5t_{c(SYS)} - t_f$		
$t_{w(COH)}$	Pulse duration, CLKOUT high	SYSCLK or MCLK ⁽⁵⁾	$0.5t_{c(SYS)} - t_f$		ns
		ICLK: X is even or 1 ⁽⁶⁾	$0.5t_{c(ICLK)} - t_f$		
		ICLK: X is odd and not 1 ⁽⁶⁾	$0.5t_{c(ICLK)} - 0.5t_{c(SYS)} - t_f$		
$t_{w(EOL)}$	Pulse duration, ECLK low	N is even and X is even or odd	$0.5t_{c(ECLK)} - t_f$		ns
		N is odd and X is even	$0.5t_{c(ECLK)} - t_f$		
		N is odd and X is odd and not 1	$0.5t_{c(ECLK)} + 0.5t_{c(SYS)} - t_f$		
$t_{w(EOH)}$	Pulse duration, ECLK high	N is even and X is even or odd	$0.5t_{c(ECLK)} - t_f$		ns
		N is odd and X is even	$0.5t_{c(ECLK)} - t_f$		
		N is odd and X is odd and not 1	$0.5t_{c(ECLK)} - 0.5t_{c(SYS)} - t_f$		

(1) Not production tested.

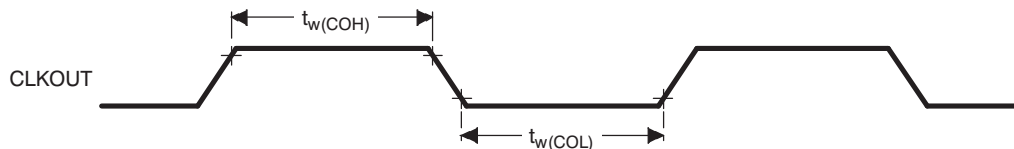
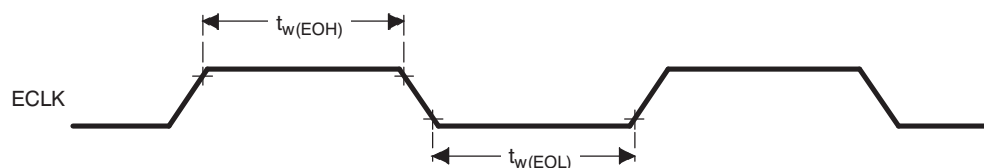
 (2) $X = \{1, 2, 3, 4, 5, 6, 7, 8, 9, 10, 11, 12, 13, 14, 15, 16\}$. X is the interface clock divider ratio determined by the PCR0[4:1] bits in the SYS module.

 (3) $N = \{1 \text{ to } 256\}$. N is the ECP prescale value defined by the ECPCTRL[7:0] register bits in the ECP module.

(4) CLKOUT/ECLK pulse durations (low/high) are a function of the OSCIN pulse durations when PLLDIS is active.

(5) Clock source bits are selected as either SYSCLK (CLKCNTL[6:5] = 11 binary) or MCLK (CLKCNTL[6:5] = 10 binary).

(6) Clock source bits are selected as ICLK (CLKCNTL[6:5] = 01 binary).


Figure 9. CLKOUT Timing Diagram

Figure 10. ECLK Timing Diagram

RST AND PORRST TIMINGS

Timing Requirements for $\overline{\text{PORRST}}^{(1)}$

(see [Figure 11](#))

		MIN	MAX	UNIT
V _{CCPORL}	V _{CC} low supply level when $\overline{\text{PORRST}}$ must be active during power up		0.6	V
V _{CCPORH}	V _{CC} high supply level when $\overline{\text{PORRST}}$ must remain active during power up and become active during power down	1.5		V
V _{CCIOPORL}	V _{CCIO} low supply level when $\overline{\text{PORRST}}$ must be active during power up		1.1	V
V _{CCIOPORH}	V _{CCIO} high supply level when $\overline{\text{PORRST}}$ must remain active during power up and become active during power down	2.75		V
V _{IL}	Low-level input voltage after V _{CCIO} > V _{CCIOPORH}		0.2 V _{CCIO}	V
V _{IL(PORRST)}	Low-level input voltage of $\overline{\text{PORRST}}$ before V _{CCIO} > V _{CCIOPORL}		0.5	V
t _{su(PORRST)r}	Setup time, $\overline{\text{PORRST}}$ active before V _{CCIO} > V _{CCIOPORL} during power up	0		ms
t _{su(VCCIO)r}	Setup time, V _{CCIO} > V _{CCIOPORL} before V _{CC} > V _{CCPORL}	0		ms
t _{h(PORRST)r}	Hold time, $\overline{\text{PORRST}}$ active after V _{CC} > V _{CCPORH}	1		ms
t _{su(PORRST)f}	Setup time, $\overline{\text{PORRST}}$ active before V _{CC} ≤ V _{CCPORH} during power down	8		μs
t _{h(PORRST)rio}	Hold time, $\overline{\text{PORRST}}$ active after V _{CC} > V _{CCIOPORH}	1		ms
t _{h(PORRST)d}	Hold time, $\overline{\text{PORRST}}$ active after V _{CC} < V _{CCPORL}	0		ms
t _{su(PORRST)fio}	Setup time, $\overline{\text{PORRST}}$ active before V _{CC} ≤ V _{CCIOPORH} during power down	0		ns
t _{su(VCCIO)f}	Setup time, V _{CC} < V _{CCPORL} before V _{CCIO} < V _{CCIOPORL}	0		ns

(1) Not production tested.

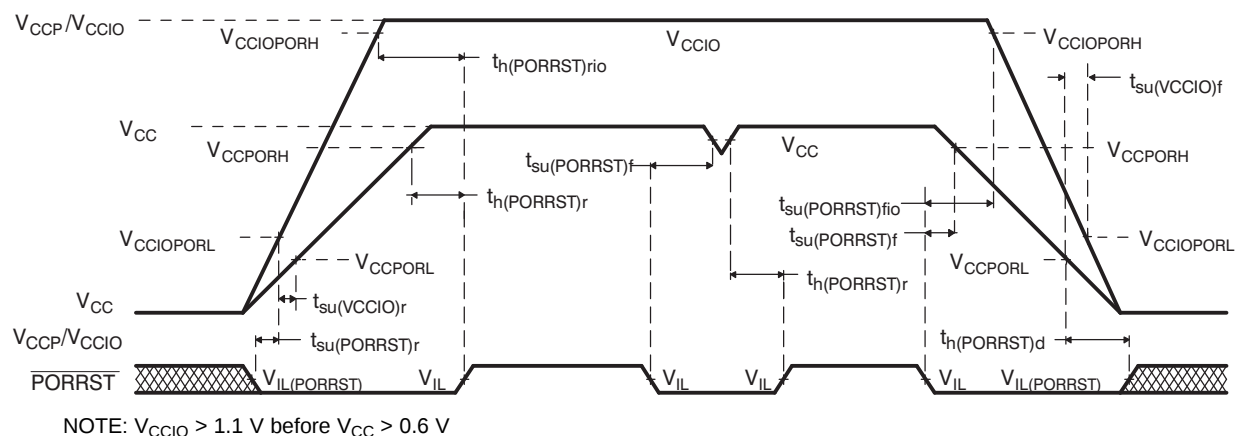


Figure 11. $\overline{\text{PORRST}}$ Timing Diagram

Switching Characteristics over Recommended Operating Conditions for $\overline{\text{RST}}^{(1) (2)}$

PARAMETER		MIN	MAX	UNIT
t _{V(RST)}	Valid time, $\overline{\text{RST}}$ active after $\overline{\text{PORRST}}$ inactive	4112	t _{c(OSC)}	ns
	Valid time, $\overline{\text{RST}}$ active (all others)	8	t _{c(SYS)}	
t _{FSU}	Flash start up time, from $\overline{\text{RST}}$ inactive to fetch of first instruction from flash (flash pump stabilization time)	836	t _{c(OSC)}	ns

(1) Not production tested.

(2) Specified values do NOT include rise/fall times. For rise and fall timings, see the "switching characteristics for output timings versus load capacitance" table.

JTAG SCAN INTERFACE TIMING (JTAG CLOCK SPECIFICATION 10-MHz AND 50-pF LOAD ON TDO OUTPUT)⁽¹⁾

		MIN	MAX	UNIT
$t_{c(JTAG)}$	Cycle time, JTAG low and high period	50		ns
$t_{su(TDI/TMS - TCKr)}$	Setup time, TDI, TMS before TCK rise (TCKr)	15		ns
$t_h(TCKr - TDI/TMS)$	Hold time, TDI, TMS after TCKr	15		ns
$t_h(TCKf - TDO)$	Hold time, TDO after TCKf	10		ns
$t_d(TCKf - TDO)$	Delay time, TDO valid after TCK fall (TCKf)		45	ns

(1) Not production tested.

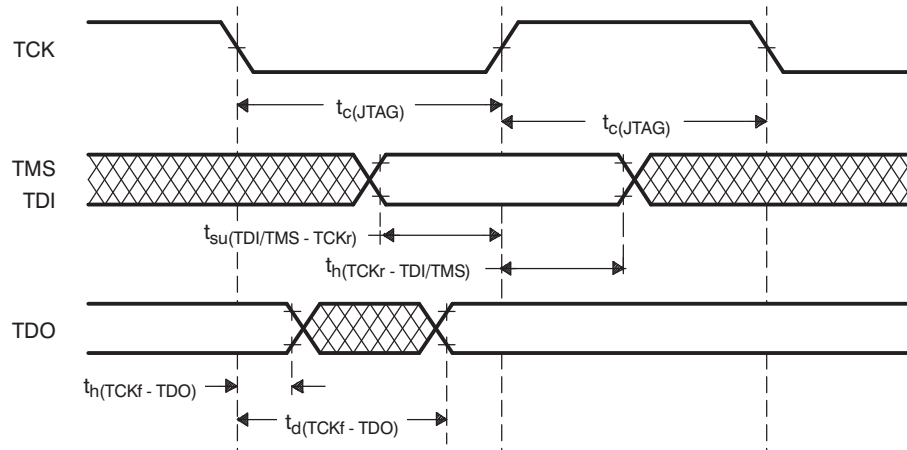


Figure 12. JTAG Scan Timings

OUTPUT TIMINGS

Switching Characteristics for Output Timings versus Load Capacitance C_L ⁽¹⁾

(see Figure 13)

PARAMETER			MIN	MAX	UNIT
t_r	Rise time, AWD, CLKOUT, $\overline{\text{TDI}}$, TDO, TMS, TMS2	$C_L = 15 \text{ pF}$	0.5	2.5	ns
		$C_L = 50 \text{ pF}$	1.5	5.0	
		$C_L = 100 \text{ pF}$	3.0	9.0	
		$C_L = 150 \text{ pF}$	4.5	12.5	
t_f	Fall time, AWD, CLKOUT, $\overline{\text{TDI}}$, TDO, TMS, TMS2	$C_L = 15 \text{ pF}$	0.5	2.5	ns
		$C_L = 50 \text{ pF}$	1.5	5.0	
		$C_L = 100 \text{ pF}$	3.0	9.0	
		$C_L = 150 \text{ pF}$	4.5	12.5	
t_r	Rise time, $\overline{\text{RST}}$	$C_L = 15 \text{ pF}$	2.5	8	ns
		$C_L = 50 \text{ pF}$	5	14	
		$C_L = 100 \text{ pF}$	9	23	
		$C_L = 150 \text{ pF}$	13	32	
t_r	Rise time, 4mA, 5 V tolerant pins	$C_L = 15 \text{ pF}$	3	10	ns
		$C_L = 50 \text{ pF}$	3.5	12	
		$C_L = 100 \text{ pF}$	7	21	
		$C_L = 150 \text{ pF}$	9	28	
		$C_L = 400 \text{ pF}$	18	40	
t_f	Fall time, 4mA, 5 V tolerant pins	$C_L = 15 \text{ pF}$	2	8	ns
		$C_L = 50 \text{ pF}$	2.5	9	
		$C_L = 100 \text{ pF}$	8	25	
		$C_L = 150 \text{ pF}$	11	35	
		$C_L = 400 \text{ pF}$	20	45	
t_r	Rise time, all other output pins	$C_L = 15 \text{ pF}$	2.5	10	ns
		$C_L = 50 \text{ pF}$	6.0	25	
		$C_L = 100 \text{ pF}$	12	45	
		$C_L = 150 \text{ pF}$	18	65	
t_f	Fall time, all other output pins	$C_L = 15 \text{ pF}$	3	10	ns
		$C_L = 50 \text{ pF}$	8.5	25	
		$C_L = 100 \text{ pF}$	16	45	
		$C_L = 150 \text{ pF}$	23	65	

(1) Not production tested.

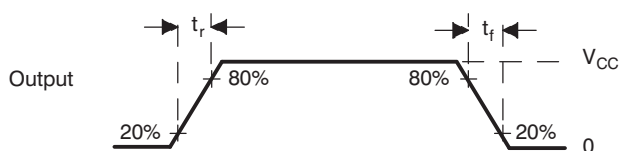


Figure 13. CMOS-Level Outputs

INPUT TIMINGS

Timing Requirements for Input Timings⁽¹⁾⁽²⁾

(see [Figure 14](#))

		MIN	MAX	UNIT
t_{pw}	Input minimum pulse width	$t_{c(ICK)} + 10$		ns

(1) Not production tested.

(2) $t_{c(ICK)}$ = interface clock cycle time = $1/f_{(ICK)}$

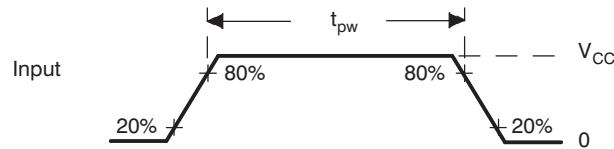


Figure 14. CMOS-Level Inputs

FLASH TIMINGS

Timing Requirements for Program Flash⁽¹⁾⁽²⁾

		MIN	TYP	MAX	UNIT
$t_{prog(16-bit)}$	Half word (16-bit) programming time	4	16	200	μ s
$t_{prog(Total)}$	1M-byte programming time ⁽³⁾		8	32	s
$t_{erase(secter)}$	Sector erase time, $T_A = -40^{\circ}\text{C}$ to 150°C		1.7		s
t_{wec}	Write/erase cycles at $T_A = -40^{\circ}\text{C}$ to 85°C	50000			cycles
$t_{fp(RST)}$	Flash pump settling time from $\overline{RST}$ to SLEEP		$167t_{c(SYS)}$		ns
$t_{fp(SLEEP)}$	Initial flash pump settling time from SLEEP to STANDBY		$167t_{c(SYS)}$		ns
$t_{fp(STANDBY)}$	Initial flash pump settling time from STANDBY to ACTIVE		$84t_{c(SYS)}$		ns

(1) Not production tested.

(2) For more detailed information on the flash core sectors, see the *flash program and erase* section of this data sheet.

(3) The 1M-byte programming time includes overhead of state machine.

SPI_{IN} MASTER MODE TIMING PARAMETERS

SPI_{IN} Master Mode External Timing Parameters

(CLOCK PHASE = 0, SPI_{IN}CLK = output, SPI_{IN}SIMO = output, and SPI_{IN}SOMI = input)^{(1)(2) (3) (4)} (see Figure 15)

NO.			MIN	MAX	UNIT
1	$t_{c(SPC)M}$	Cycle time, SPI _{IN} CLK ⁽⁵⁾	100	$256t_{c(I)CLK}$	ns
2 ⁽⁶⁾	$t_{w(SPCH)M}$	Pulse duration, SPI _{IN} CLK high (clock polarity = 0)	$0.5t_{c(SPC)M} - t_r$	$0.5t_{c(SPC)M} + 5$	
	$t_{w(SPCL)M}$	Pulse duration, SPI _{IN} CLK low (clock polarity = 1)	$0.5t_{c(SPC)M} - t_f$	$0.5t_{c(SPC)M} + 5$	
3 ⁽⁶⁾	$t_{w(SPCL)M}$	Pulse duration, SPI _{IN} CLK low (clock polarity = 0)	$0.5t_{c(SPC)M} - t_f$	$0.5t_{c(SPC)M} + 5$	
	$t_{w(SPCH)M}$	Pulse duration, SPI _{IN} CLK high (clock polarity = 1)	$0.5t_{c(SPC)M} - t_r$	$0.5t_{c(SPC)M} + 5$	
4 ⁽⁶⁾	$t_{d(SPCH-SIMO)M}$	Delay time, SPI _{IN} CLK high to SPI _{IN} SIMO valid (clock polarity = 0)		10	
	$t_{d(SPCL-SIMO)M}$	Delay time, SPI _{IN} CLK low to SPI _{IN} SIMO valid (clock polarity = 1)		10	
5 ⁽⁶⁾	$t_{v(SPCL-SIMO)M}$	Valid time, SPI _{IN} SIMO data valid after SPI _{IN} CLK low (clock polarity = 0)	$t_{c(SPC)M} - 5 - t_f$		
	$t_{v(SPCH-SIMO)M}$	Valid time, SPI _{IN} SIMO data valid after SPI _{IN} CLK high (clock polarity = 1)	$t_{c(SPC)M} - 5 - t_r$		
6 ⁽⁶⁾	$t_{su(SOMI-SPCL)M}$	Setup time, SPI _{IN} SOMI before SPI _{IN} CLK low (clock polarity = 0)	6		
	$t_{su(SOMI-SPCH)M}$	Setup time, SPI _{IN} SOMI before SPI _{IN} CLK high (clock polarity = 1)	6		
7 ⁽⁶⁾	$t_{v(SPCL-SOMI)M}$	Valid time, SPI _{IN} SOMI data valid after SPI _{IN} CLK low (clock polarity = 0)	4		
	$t_{v(SPCH-SOMI)M}$	Valid time, SPI _{IN} SOMI data valid after SPI _{IN} CLK high (clock polarity = 1)	4		

- (1) Not production tested.
- (2) The MASTER bit (SPI_{IN}CTRL2.3) is set and the CLOCK PHASE bit (SPI_{IN}CTRL2.0) is cleared.
- (3) $t_{c(I)CLK}$ = interface clock cycle time = $1/f_{(I)CLK}$
- (4) For rise and fall timings, see the "Switching Characteristics for Output Timings versus Load Capacitance" table.
- (5) When the SPI is in master mode, the following must be true:
For PS values from 1 to 255: $t_{c(SPC)M} \geq (PS + 1)t_{c(I)CLK} \geq 100$ ns, where PS is the prescale value set in the SPI_{IN}CTL1[12:5] register bits.
For PS values of 0: $t_{c(SPC)M} = 2t_{c(I)CLK} \geq 100$ ns.
- (6) The active edge of the SPI_{IN}CLK signal referenced is controlled by the CLOCK POLARITY bit (SPI_{IN}CTRL2.1).

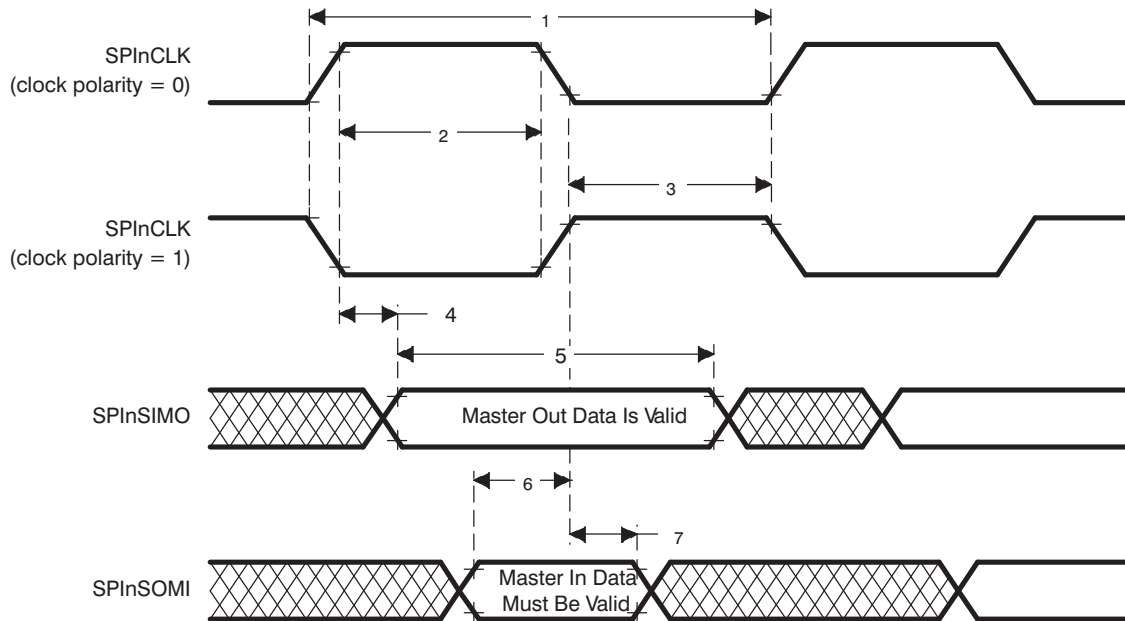


Figure 15. SPI_{IN} Master Mode External Timing (CLOCK PHASE = 0)

SPIn Master Mode External Timing Parameters⁽¹⁾

(CLOCK PHASE = 1, SPInCLK = output, SPInSIMO = output, and SPInSOMI = input)^{(2) (3) (4)} (see Figure 16)

NO.			MIN	MAX	UNIT
1	$t_{c(SPC)M}$	Cycle time, SPInCLK ⁽⁵⁾	100	$256t_{c(ICK)M}$	
2 ⁽⁶⁾	$t_{w(SPCH)M}$	Pulse duration, SPInCLK high (clock polarity = 0)	$0.5t_{c(SPC)M} - t_r$	$0.5t_{c(SPC)M} + 5$	ns
	$t_{w(SPCL)M}$	Pulse duration, SPInCLK low (clock polarity = 1)	$0.5t_{c(SPC)M} - t_f$	$0.5t_{c(SPC)M} + 5$	
3 ⁽⁶⁾	$t_{w(SPCL)M}$	Pulse duration, SPInCLK low (clock polarity = 0)	$0.5t_{c(SPC)M} - t_f$	$0.5t_{c(SPC)M} + 5$	
	$t_{w(SPCH)M}$	Pulse duration, SPInCLK high (clock polarity = 1)	$0.5t_{c(SPC)M} - t_r$	$0.5t_{c(SPC)M} + 5$	
4 ⁽⁶⁾	$t_{v(SIMO-SPCH)M}$	Valid time, SPInCLK high after SPInSIMO data valid (clock polarity = 0)	$0.5t_{c(SPC)M} - 10$		
	$t_{v(SIMO-SPCL)M}$	Valid time, SPInCLK low after SPInSIMO data valid (clock polarity = 1)	$0.5t_{c(SPC)M} - 10$		
5 ⁽⁶⁾	$t_{v(SPCH-SIMO)M}$	Valid time, SPInSIMO data valid after SPInCLK high (clock polarity = 0)	$0.5t_{c(SPC)M} - 5 - t_r$		
	$t_{v(SPCL-SIMO)M}$	Valid time, SPInSIMO data valid after SPInCLK low (clock polarity = 1)	$0.5t_{c(SPC)M} - 5 - t_f$		
6 ⁽⁶⁾	$t_{su(SOMI-SPCH)M}$	Setup time, SPInSOMI before SPInCLK high (clock polarity = 0)	6		
	$t_{su(SOMI-SPCL)M}$	Setup time, SPInSOMI before SPInCLK low (clock polarity = 1)	6		
7 ⁽⁶⁾	$t_{v(SPCH-SOMI)M}$	Valid time, SPInSOMI data valid after SPInCLK high (clock polarity = 0)	4		
	$t_{v(SPCL-SOMI)M}$	Valid time, SPInSOMI data valid after SPInCLK low (clock polarity = 1)	4		

(1) Not production tested.

(2) The MASTER bit (SPInCTRL2.3) is set and the CLOCK PHASE bit (SPInCTRL2.0) is set.

(3) $t_{c(ICK)} = \text{interface clock cycle time} = 1/f_{(ICK)}$

(4) For rise and fall timings, see the "Switching Characteristics for Output Timings versus Load Capacitance" table.

(5) When the SPI is in master mode, the following must be true:

For PS values from 1 to 255: $t_{c(SPC)M} \geq (PS + 1)t_{c(ICK)} \geq 100 \text{ ns}$, where PS is the prescale value set in the SPInCTL1[12:5] register bits.

For PS values of 0: $t_{c(SPC)M} = 2t_{c(ICK)} \geq 100 \text{ ns}$.

(6) The active edge of the SPInCLK signal referenced is controlled by the CLOCK POLARITY bit (SPInCTRL2.1).

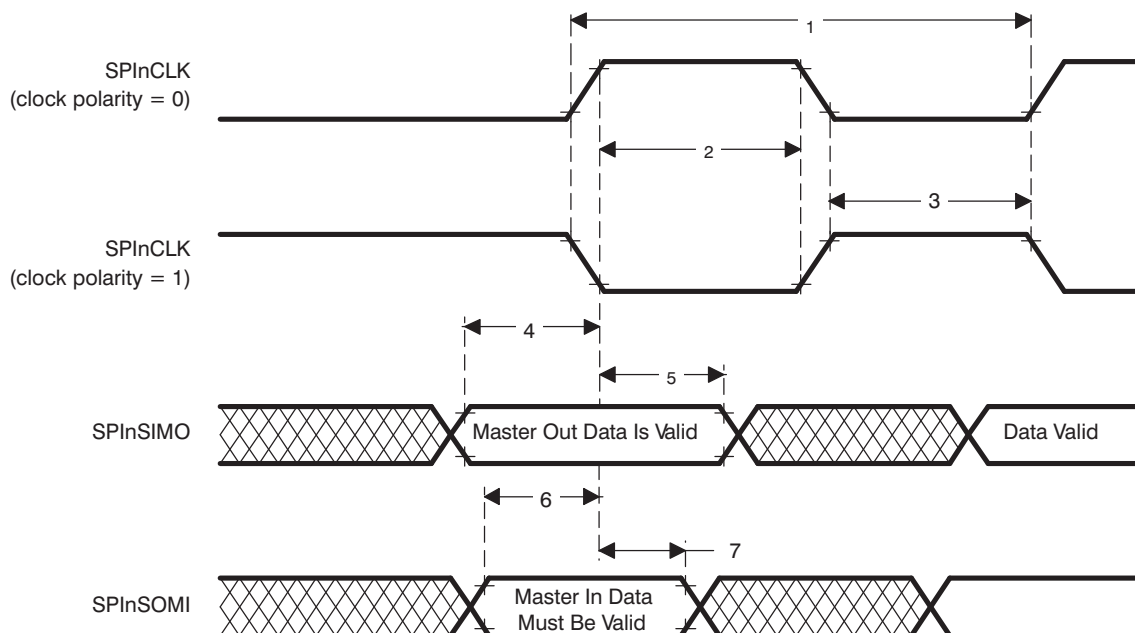


Figure 16. SPIn Master Mode External Timing (CLOCK PHASE = 1)

SPIn SLAVE MODE TIMING PARAMETERS

SPIn Slave Mode External Timing Parameters⁽¹⁾

(CLOCK PHASE = 0, SPInCLK = input, SPInSIMO = input, and SPInSOMI = output)^{(2) (3) (4) (5)} (see Figure 17)

NO.			MIN	MAX	UNIT
1	$t_{c(SPC)S}$	Cycle time, SPInCLK ⁽⁶⁾	100	$256t_{c(ICLK)}$	ns
2 ⁽⁷⁾	$t_{w(SPCH)S}$	Pulse duration, SPInCLK high (clock polarity = 0)	$0.5t_{c(SPC)S} - 0.25t_{c(ICLK)}$	$0.5t_{c(SPC)S} + 0.25t_{c(ICLK)}$	
	$t_{w(SPCL)S}$	Pulse duration, SPInCLK low (clock polarity = 1)	$0.5t_{c(SPC)S} - 0.25t_{c(ICLK)}$	$0.5t_{c(SPC)S} + 0.25t_{c(ICLK)}$	
3 ⁽⁷⁾	$t_{w(SPCL)S}$	Pulse duration, SPInCLK low (clock polarity = 0)	$0.5t_{c(SPC)S} - 0.25t_{c(ICLK)}$	$0.5t_{c(SPC)S} + 0.25t_{c(ICLK)}$	
	$t_{w(SPCH)S}$	Pulse duration, SPInCLK high (clock polarity = 1)	$0.5t_{c(SPC)S} - 0.25t_{c(ICLK)}$	$0.5t_{c(SPC)S} + 0.25t_{c(ICLK)}$	
4 ⁽⁷⁾	$t_{d(SPCH-SOMI)S}$	Delay time, SPInCLK high to SPInSOMI valid (clock polarity = 0)		$6 + t_r$	
	$t_{d(SPCL-SOMI)S}$	Delay time, SPInCLK low to SPInSOMI valid (clock polarity = 1)		$6 + t_r$	
5 ⁽⁷⁾	$t_{v(SPCH-SOMI)S}$	Valid time, SPInSOMI data valid after SPInCLK high (clock polarity = 0)	$t_{c(SPC)S} - 6 - t_r$		
	$t_{v(SPCL-SOMI)S}$	Valid time, SPInSOMI data valid after SPInCLK low (clock polarity = 1)	$t_{c(SPC)S} - 6 - t_r$		
6 ⁽⁷⁾	$t_{su(SIMO-SPCL)S}$	Setup time, SPInSIMO before SPInCLK low (clock polarity = 0)	6		
	$t_{su(SIMO-SPCH)S}$	Setup time, SPInSIMO before SPInCLK high (clock polarity = 1)	6		
7 ⁽⁷⁾	$t_{v(SPCL-SIMO)S}$	Valid time, SPInSIMO data valid after SPInCLK low (clock polarity = 0)	6		
	$t_{v(SPCH-SIMO)S}$	Valid time, SPInSIMO data valid after SPInCLK high (clock polarity = 1)	6		

(1) Not production tested.

(2) The MASTER bit (SPInCTRL2.3) is cleared and the CLOCK PHASE bit (SPInCTRL2.0) is cleared.

(3) If the SPI is in slave mode, the following must be true: $t_{c(SPC)S} \geq (PS + 1)t_{c(ICLK)}$, where PS = prescale value set in SPInCTL1[12:5].

(4) For rise and fall timings, see the "Switching Characteristics for Output Timings versus Load Capacitance" table.

(5) $t_{c(ICLK)}$ = interface clock cycle time = $1/f_{(ICLK)}$

(6) When the SPIn is in slave mode, the following must be true:

For PS values from 1 to 255: $t_{c(SPC)S} \geq (PS + 1)t_{c(ICLK)} \geq 100$ ns, where PS is the prescale value set in the SPInCTL1[12:5] register bits.

For PS values of 0: $t_{c(SPC)S} = 2t_{c(ICLK)} \geq 100$ ns.

(7) The active edge of the SPInCLK signal referenced is controlled by the CLOCK POLARITY bit (SPInCTRL2.1).

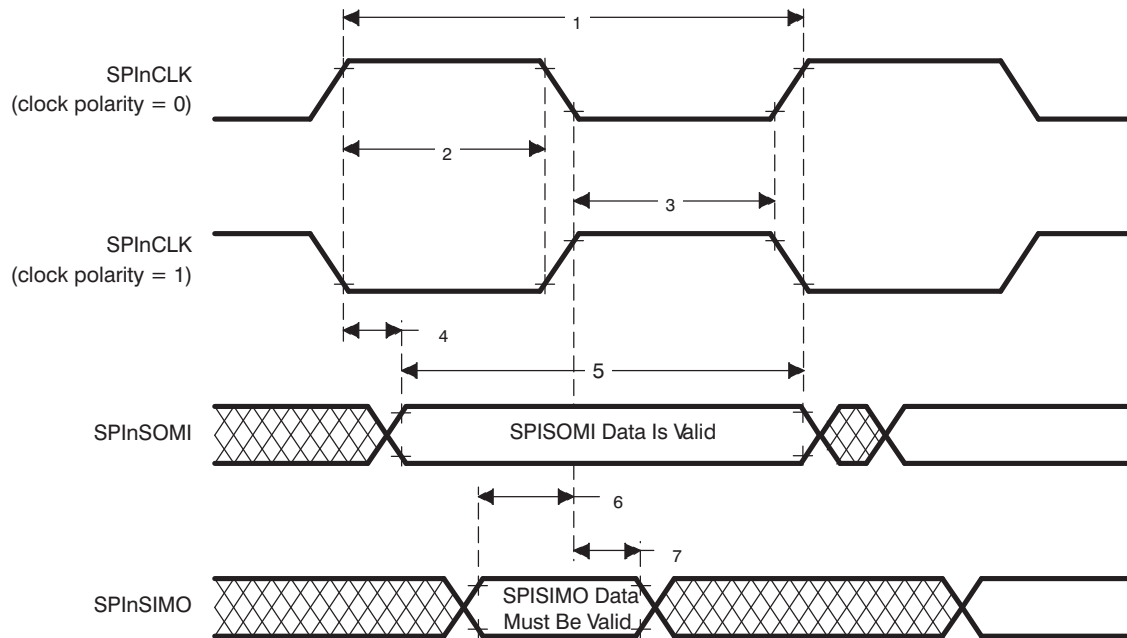


Figure 17. SPIn Slave Mode External Timing (CLOCK PHASE = 0)

SPIn Slave Mode External Timing Parameters⁽¹⁾

(CLOCK PHASE = 1, SPInCLK = input, SPInSIMO = input, and SPInSOMI = output)^{(2) (3) (4) (5)} (see Figure 18)

NO.			MIN	MAX	UNIT
1	$t_{c(SPC)S}$	Cycle time, SPInCLK ⁽⁶⁾	100	$256t_{c(ICLK)}$	
2 ⁽⁷⁾	$t_{w(SPCH)S}$	Pulse duration, SPInCLK high (clock polarity = 0)	$0.5t_{c(SPC)S} - 0.25t_{c(ICLK)}$	$0.5t_{c(SPC)S} + 0.25t_{c(ICLK)}$	
	$t_{w(SPCL)S}$	Pulse duration, SPInCLK low (clock polarity = 1)	$0.5t_{c(SPC)S} - 0.25t_{c(ICLK)}$	$0.5t_{c(SPC)S} + 0.25t_{c(ICLK)}$	
3 ⁽⁷⁾	$t_{w(SPCL)S}$	Pulse duration, SPInCLK low (clock polarity = 0)	$0.5t_{c(SPC)S} - 0.25t_{c(ICLK)}$	$0.5t_{c(SPC)S} + 0.25t_{c(ICLK)}$	
	$t_{w(SPCH)S}$	Pulse duration, SPInCLK high (clock polarity = 1)	$0.5t_{c(SPC)S} - 0.25t_{c(ICLK)}$	$0.5t_{c(SPC)S} + 0.25t_{c(ICLK)}$	
4 ⁽⁷⁾	$t_{v(SOMI-SPCH)S}$	Valid time, SPInCLK high after SPInSOMI data valid (clock polarity = 0)	$0.5t_{c(SPC)S} - 6 - t_r$		ns
	$t_{v(SOMI-SPCL)S}$	Valid time, SPInCLK low after SPInSOMI data valid (clock polarity = 1)	$0.5t_{c(SPC)S} - 6 - t_f$		
5 ⁽⁷⁾	$t_{v(SPCH-SOMI)S}$	Valid time, SPInSOMI data valid after SPInCLK high (clock polarity = 0)	$0.5t_{c(SPC)S} - 6 - t_r$		
	$t_{v(SPCL-SOMI)S}$	Valid time, SPInSOMI data valid after SPInCLK low (clock polarity = 1)	$0.5t_{c(SPC)S} - 6 - t_f$		
6 ⁽⁷⁾	$t_{su(SIMO-SPCH)S}$	Setup time, SPInSIMO before SPInCLK high (clock polarity = 0)	6		
	$t_{su(SIMO-SPCL)S}$	Setup time, SPInSIMO before SPInCLK low (clock polarity = 1)	6		
7 ⁽⁷⁾	$t_{v(SPCH-SIMO)S}$	Valid time, SPInSIMO data valid after SPInCLK high (clock polarity = 0)	6		
	$t_{v(SPCL-SIMO)S}$	Valid time, SPInSIMO data valid after SPInCLK low (clock polarity = 1)	6		

(1) Not production tested.

(2) The MASTER bit (SPInCTRL2.3) is cleared and the CLOCK PHASE bit (SPInCTRL2.0) is set.

(3) If the SPI is in slave mode, the following must be true: $t_{c(SPC)} \geq (PS + 1)t_{c(ICLK)}$, where PS = prescale value set in SPInCTL1[12:5].

(4) For rise and fall timings, see the "Switching Characteristics for Output Timings versus Load Capacitance" table.

(5) $t_{c(ICLK)}$ = interface clock cycle time = $1/f_{ICLK}$

(6) When the SPIn is in slave mode, the following must be true:

For PS values from 1 to 255: $t_{c(SPC)S} \geq (PS + 1)t_{c(ICLK)} \geq 100$ ns, where PS is the prescale value set in the SPInCTL1[12:5] register bits.

For PS values of 0: $t_{c(SPC)S} \geq 2t_{c(ICLK)} \geq 100$ ns.

(7) The active edge of the SPInCLK signal referenced is controlled by the CLOCK POLARITY bit (SPInCTRL2.1).

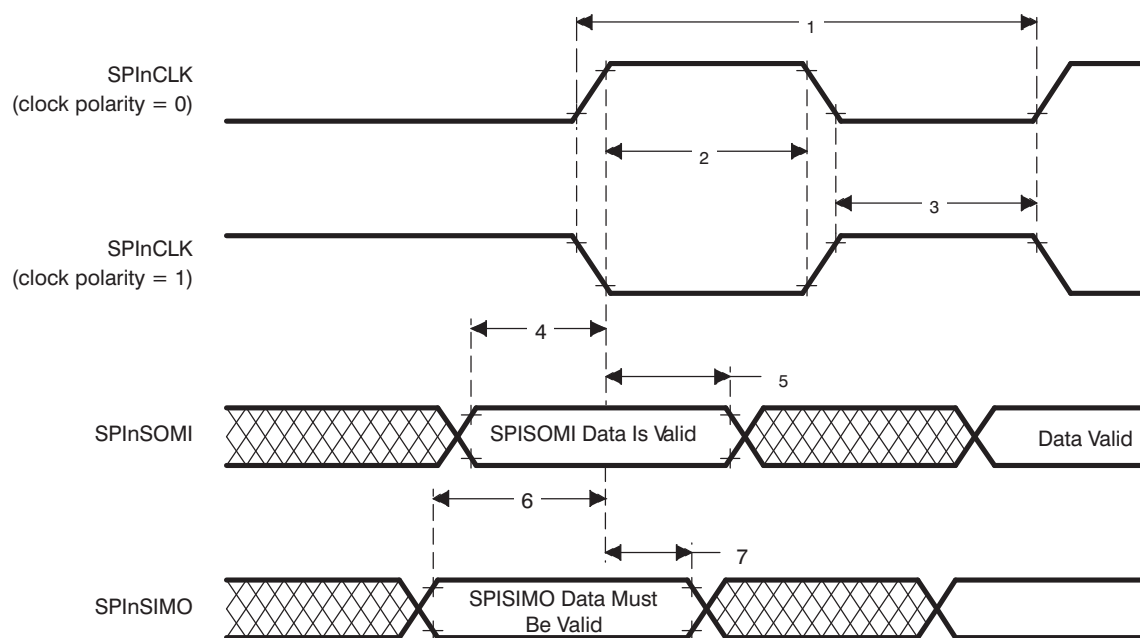


Figure 18. SPIn Slave Mode External Timing (CLOCK PHASE = 1)

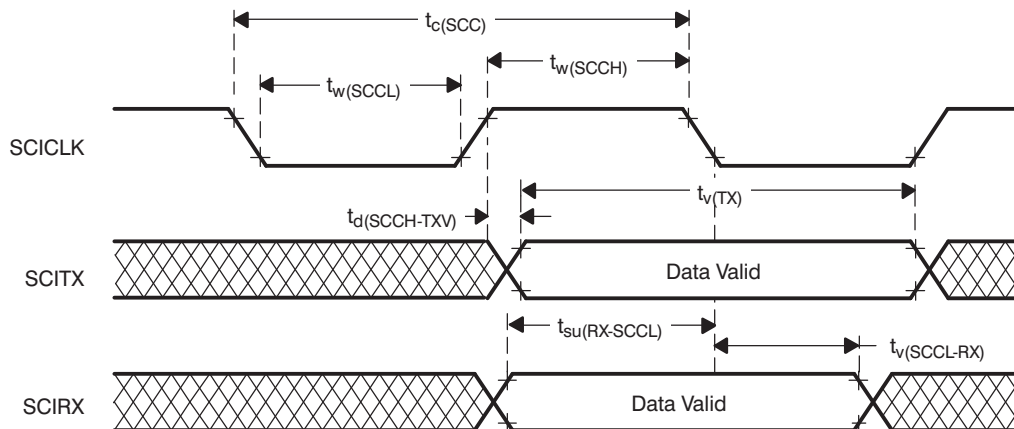
SCIn ISOSYNCHRONOUS MODE TIMINGS - INTERNAL CLOCK

Timing Requirements for Internal Clock SCIn Isosynchronous Mode^{(1)(2) (3) (4)}

(see Figure 19)

		(BAUD + 1) IS EVEN OR BAUD = 0		(BAUD + 1) IS ODD AND BAUD ≠ 0		UNIT
		MIN	MAX	MIN	MAX	
$t_{c(SCC)}$	Cycle time, SCInCLK	$2t_{c(ICKL)}$	$2^{24} t_{c(ICKL)}$	$3t_{c(ICKL)}$	$(2^{24} - 1) t_{c(ICKL)}$	ns
$t_{w(SCCL)}$	Pulse duration, SCInCLK low	$0.5t_{c(SCC)} - t_f$	$0.5t_{c(SCC)} + 5$	$0.5t_{c(SCC)} + 0.5t_{c(ICKL)} - t_f$	$0.5t_{c(SCC)} + 0.5t_{c(ICKL)}$	ns
$t_{w(SCCH)}$	Pulse duration, SCInCLK high	$0.5t_{c(SCC)} - t_r$	$0.5t_{c(SCC)} + 5$	$0.5t_{c(SCC)} - 0.5t_{c(ICKL)} - t_r$	$0.5t_{c(SCC)} - 0.5t_{c(ICKL)}$	ns
$t_{d(SCCH-TXV)}$	Delay time, SCInCLK high to SCInTX valid		10		10	ns
$t_{v(TX)}$	Valid time, SCInTX data after SCInCLK low	$t_{c(SCC)} - 10$		$t_{c(SCC)} - 10$		ns
$t_{su(RX-SCCL)}$	Setup time, SCInRX before SCInCLK low	$t_{c(ICKL)} + t_f + 20$		$t_{c(ICKL)} + t_f + 20$		ns
$t_{v(SCCL-RX)}$	Valid time, SCInRX data after SCInCLK low	$-t_{c(ICKL)} + t_f + 20$		$-t_{c(ICKL)} + t_f + 20$		ns

- (1) Not production tested.
(2) BAUD = 24-bit concatenated value formed by the SCI[H,M,L]BAUD registers.
(3) $t_{c(ICKL)}$ = interface clock cycle time = $1/f_{(ICKL)}$
(4) For rise and fall timings, see the "switching characteristics for output timings versus load capacitance" table.



- B. Data transmission/reception characteristics for isosynchronous mode with internal clocking are similar to the asynchronous mode. Data transmission occurs on the SCICLK rising edge, and data reception occurs on the SCICLK falling edge.

Figure 19. SCIn Isosynchronous Mode Timing Diagram for Internal Clock

SCIn ISOSYNCHRONOUS MODE TIMINGS - EXTERNAL CLOCK

Timing Requirements for External Clock SCIn Isosynchronous Mode^{(1)(2) (3)}

(see [Figure 20](#))

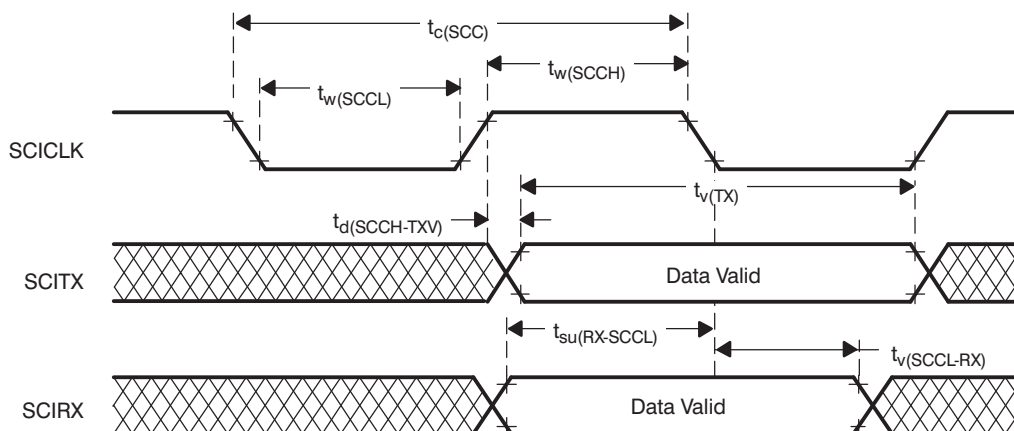
		MIN	MAX	UNIT
$t_{c(SCC)}$	Cycle time, SCInCLK ⁽⁴⁾	$8t_{c(ICK)}$		ns
$t_{w(SCCH)}$	Pulse duration, SCInCLK high	$0.5t_{c(SCC)} - 0.25t_{c(ICK)}$	$0.5t_{c(SCC)} + 0.25t_{c(ICK)}$	ns
$t_{w(SCCL)}$	Pulse duration, SCInCLK low	$0.5t_{c(SCC)} - 0.25t_{c(ICK)}$	$0.5t_{c(SCC)} + 0.25t_{c(ICK)}$	ns
$t_{d(SCCH-TXV)}$	Delay time, SCInCLK high to SCInTX valid		$2t_{c(ICK)} + 12 + t_r$	ns
$t_{v(TX)}$	Valid time, SCInTX data after SCInCLK low	$2t_{c(SCC)} - 10$		ns
$t_{su(RX-SCCL)}$	Setup time, SCInRX before SCInCLK low	0		ns
$t_{v(SCCL-RX)}$	Valid time, SCInRX data after SCInCLK low	$2t_{c(ICK)} + 10$		ns

(1) Not production tested.

(2) $t_{c(ICK)} = \text{interface clock cycle time} = 1/f_{(ICK)}$

(3) For rise and fall timings, see the "switching characteristics for output timings versus load capacitance" table.

(4) When driving an external SCInCLK, the following must be true: $t_{c(SCC)} \geq 8t_{c(ICK)}$.



- C. Data transmission / reception characteristics for isosynchronous mode with external clocking are similar to the asynchronous mode. Data transmission occurs on the SCInCLK rising edge, and data reception occurs on the SCInCLK falling edge.

Figure 20. SCIn Isosynchronous Mode Timing Diagram for External Clock

I2C TIMING

I2C Signals (SDA and SCL) Switching Characteristics⁽⁵⁾⁽⁶⁾ Assumes testing over recommended operating conditions.

I2C Signals (SDA and SCL) Switching Characteristics⁽¹⁾⁽²⁾

PARAMETER		STANDARD MODE		FAST MODE		UNIT
		MIN	MAX	MIN	MAX	
$t_{c(I2CCLK)}$	Cycle time, I2C module clock	75	150	75	150	ns
$t_{c(SCL)}$	Cycle time, SCL	10		2.5		μs
$t_{su(SCLH-SDAL)}$	Setup time, SCL high before SDA low (for a repeated START condition)	4.7		0.6		μs
$t_{h(SCLL-SDAL)}$	Hold time, SCL low after SDA low (for a repeated START condition)	4		0.6		μs
$t_{w(SCLL)}$	Pulse duration, SCL low	4.7		1.3		μs
$t_{w(SCLH)}$	Pulse duration, SCL high	4		0.6		μs
$t_{su(SDA-SCLH)}$	Setup time, SDA valid before SCL high	250		100		ns
$t_{h(SDA-SCLL)}$	Hold time, SDA valid after SCL low	0	3.45 ⁽³⁾	0	0.9	μs
$t_{w(SDAH)}$	Pulse duration, SDA high between STOP and START conditions	4.7		1.3		μs
$t_r(SCL)$	Rise time, SCL		1000	20+0.1C _b ⁽⁴⁾	300	ns
$t_r(SDA)$	Rise time, SDA		1000	20+0.1C _b ⁽⁴⁾	300	ns
$t_f(SCL)$	Fall time, SCL		300	20+0.1C _b ⁽⁴⁾	300	ns
$t_f(SDA)$	Fall time, SDA		300	20+0.1C _b ⁽⁴⁾	300	ns
$t_{su(SCLH-SDAH)}$	Setup time, SCL high before SDA high (for STOP condition)	4.0		0.6		μs
$t_{w(SP)}$	Pulse duration, spike (must be suppressed)			0	50	ns
C _b ⁽⁴⁾	Capacitive load for each bus line		400		400	pF

(5) Not production tested.

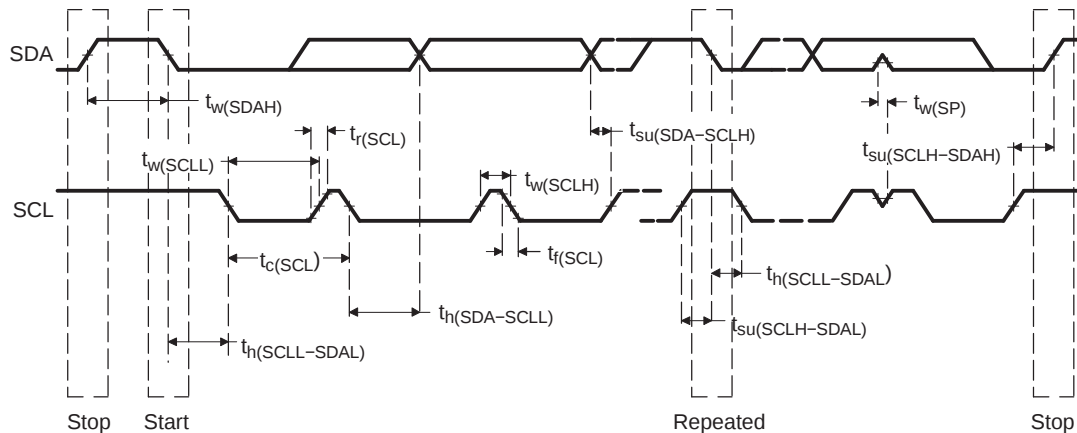
(6) The I2C pins SDA and SCL do not feature fail-safe I/O buffers. These pins could potentially draw current when the device is powered down.

(1) Not production tested.

(2) The I2C pins SDA and SCL do not feature fail-safe I/O buffers. These pins could potentially draw current when the device is powered down.

(3) The maximum $t_{h(SDA-SCLL)}$ for I2C bus devices needs to be met only if the device does not stretch the low period ($t_{w(SCLL)}$) of the SCL signal.

(4) C_b = The total capacitance of one bus line in pF. If mixed with HS-mode devices, faster fall-times are allowed.



- D. A device must internally provide a hold time of at least 300 ns for the SDA signal (referred to the V_{IHmin} of the SCL signal) to bridge the undefined region of the falling edge of SCL.
- E. The maximum $t_h(SDA-SCLL)$ needs only be met if the device does not stretch the LOW period ($t_w(SCLL)$) of the SCL signal.
- F. A fast-mode I2C-bus device can be used in a standard-mode I2C-bus system, but the requirement $t_{su}(SDA-SCLH) \geq 250$ ns must then be met. This will automatically be the case if the device does not stretch the LOW period of the SCL signal. If such a device does stretch the LOW period of the SCL signal, it must output the next data bit to the SDA line $t_r \max + t_{su}(SDA-SCLH)$.
- G. C_b = total capacitance of one bus line in pF. If mixed with HS-mode devices, faster fall-times are allowed.

Figure 21. I2C Timings

STANDARD CAN CONTROLLER (SCC) MODE TIMINGS

Dynamic Characteristics for the CANSTX and CANSRX Pins⁽¹⁾

PARAMETER		MIN	MAX	UNIT
$t_d(\text{CANSTX})$	Delay time, transmit shift register to CANSTX pin ⁽²⁾		15	ns
$t_d(\text{CANSRX})$	Delay time, CANSRX pin to receive shift register		5	ns

(1) Not production tested.

(2) These values do not include the rise/fall times of the output buffer.

EXPANSION BUS MODULE TIMING

Expansion Bus Timing Parameters⁽¹⁾

–55°C ≤ T_A ≤ 220°C, 3.0 V ≤ V_{CC} ≤ 3.6 V (see [Figure 22](#) and [Figure 23](#))

		MIN	MAX	UNIT
t _c (CO)	Cycle time, CLKOUT	20.8		ns
t _d (COH-EBADV)	Delay time, CLKOUT high to EBADDR valid		21.4	ns
t _h (COH-EBADIV)	Hold time, EBADDR invalid after CLKOUT high		12.4	ns
t _d (COH-EBOE)	Delay time, CLKOUT high to $\overline{\text{EBOE}}$ fall		11.4	ns
t _h (COH-EBOEH)	Hold time, $\overline{\text{EBOE}}$ rise after CLKOUT high		11.4	ns
t _d (COL-EBWR)	Delay time, CLKOUT low to write strobe ($\overline{\text{EBWR}}$) low		11.3	ns
t _h (COL-EBWRH)	Hold time, $\overline{\text{EBWR}}$ high after CLKOUT low		11.6	ns
t _{su} (EBRDATV-COH)	Setup time, EBDATA valid before CLKOUT high (READ) ⁽²⁾	15.2		ns
t _h (COH-EBRDATIV)	Hold time, EBDATA invalid after CLKOUT high (READ)		(–14.7)	ns
t _d (COL-EBWDATV)	Delay time, CLKOUT low to EBDATA valid (WRITE) ⁽³⁾		16.1	ns
t _h (COL-EBWDATIV)	Hold time, EBDATA invalid after CLKOUT low (WRITE)		14.7	ns
SECONDARY TIMES				
t _d (COH-EBCS0)	Delay, CLKOUT high to $\overline{\text{EBCS0}}$ fall		13.6	ns
t _h (COH-EBCS0H)	Hold, $\overline{\text{EBCS0}}$ rise after CLKOUT high		13.2	ns
t _{su} (COH-EBHOLDL)	Setup time, $\overline{\text{EBHOLD}}$ low to CLKOUT high ⁽²⁾	10.9		ns
t _{su} (COH-EBHOLDH)	Setup time, $\overline{\text{EBHOLD}}$ high to CLKOUT high ⁽²⁾	10.5		ns

(1) Not production tested.

(2) Setup time is the minimum time under worst case conditions. Data with less setup time will not work.

(3) Valid after CLKOUT goes low for write cycles.

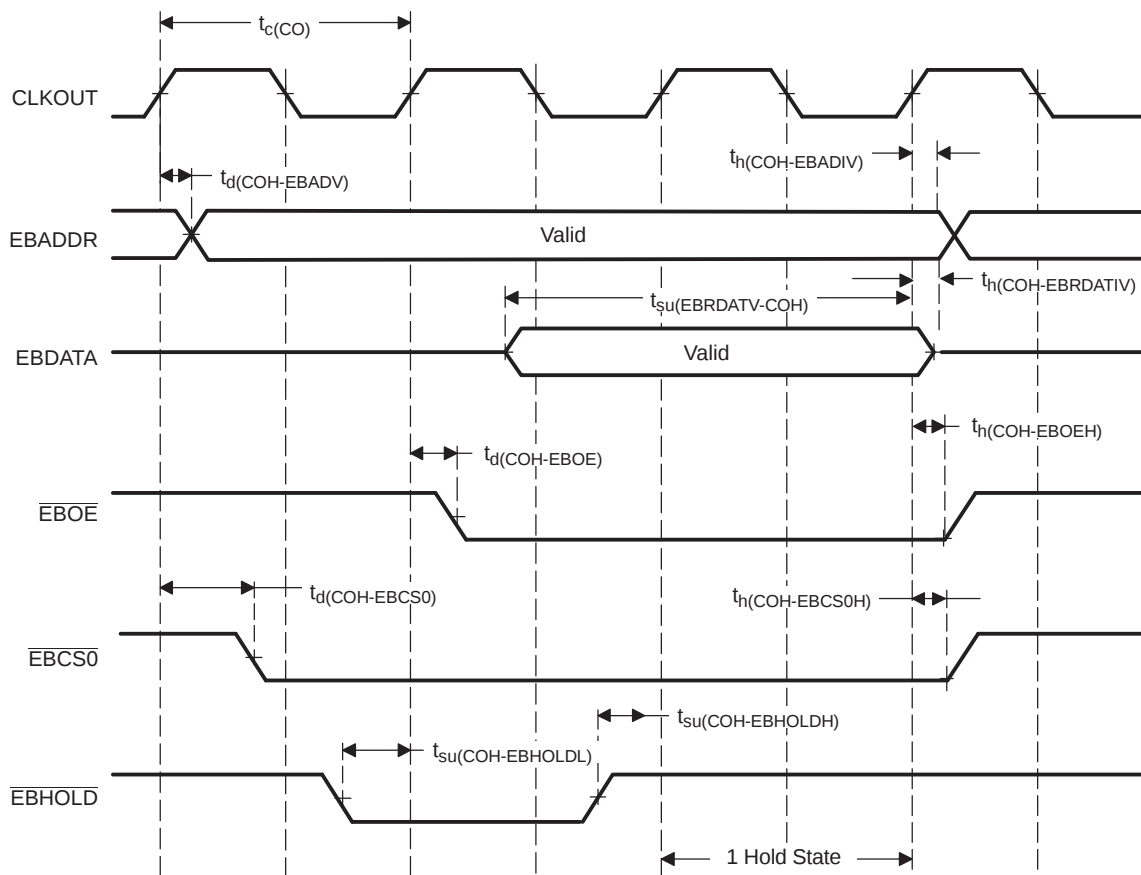


Figure 22. Expansion Memory Signal Timing - Reads

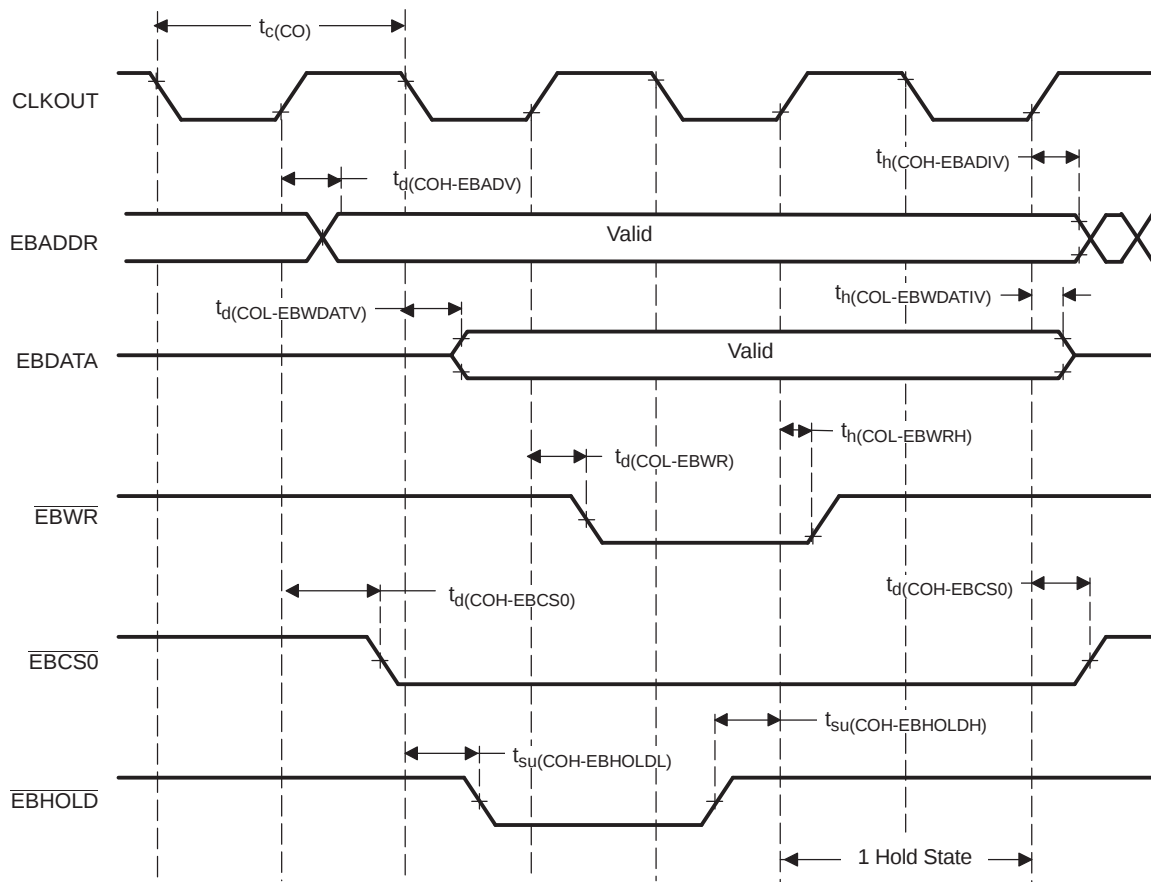


Figure 23. Expansion Memory Signal Timing - Writes

HIGH-END TIMER (HET) TIMINGS

Minimum PWM Output Pulse Width:

This is equal to one high resolution clock period (HRP). The HRP is defined by the 6-bit high resolution prescale factor (hr), which is user defined, giving prescale factors of 1 to 64, with a linear increment of codes.

Therefore, the minimum PWM output pulse width = $\text{HRP}(\text{min}) = \text{hr}(\text{min})/\text{SYSCLK} = 1/\text{SYSCLK}$

For example, for a SYSCLK of 30 MHz, the minimum PWM output pulse width = $1/30 = 33.33\text{ns}$

Minimum Input Pulses that Can Be Captured:

The input pulse width must be greater or equal to the low resolution clock period (LRP), i.e., the HET loop (the HET program must fit within the LRP). The LRP is defined by the 3-bit loop-resolution prescale factor (lr), which is user defined, with a power of 2 increment of codes. That is, the value of lr can be 1, 2, 4, 8, 16, or 32.

Therefore, the minimum input pulse width = $\text{LRP}(\text{min}) = \text{hr}(\text{min}) * \text{lr}(\text{min})/\text{SYSCLK} = 1 * 1/\text{SYSCLK}$

For example, with a SYSCLK of 30 MHz, the minimum input pulse width = $1 * 1/30 = 33.33 \text{ ns}$

NOTE

Once the input pulse width is greater than LRP, the resolution of the measurement is still HRP. (That is, the captured value gives the number of HRP clocks inside the pulse.)

Abbreviations:

hr = HET high resolution divide rate = 1, 2, 3,...63, 64

lr = HET low resolution divide rate = 1, 2, 4, 8, 16, 32

High resolution clock period = $\text{HRP} = \text{hr}/\text{SYSCLK}$

Loop resolution clock period = $\text{LRP} = \text{hr} * \text{lr}/\text{SYSCLK}$

MULTI-BUFFERED A-TO-D CONVERTER (MibADC)

The multi-buffered A-to-D converter (MibADC) has a separate power bus for its analog circuitry that enhances the A-to-D performance by preventing digital switching noise on the logic circuitry, which could be present on V_{SS} and V_{CC}, from coupling into the A-to-D analog stage. All A-to-D specifications are given with respect to AD_{REFLO} unless otherwise noted.

Resolution	10 bits (1024 values)
Monotonic	Assured
Output conversion code	00h to 3FFh [00 for V _{AI} ≤ AD _{REFLO} ; 3FF for V _{AI} ≥ AD _{REFHI}]

Table 13. MibADC Recommended Operating Conditions⁽¹⁾

		MIN	MAX	UNIT
AD _{REFHI}	A-to-D high-voltage reference source	V _{SSAD}	V _{CCAD}	V
AD _{REFLO}	A-to-D low-voltage reference source	V _{SSAD}	V _{CCAD}	V
V _{AI}	Analog input voltage	V _{SSAD} – 0.3	V _{CCAD} + 0.3	V
I _{AI}	Analog input clamp current ⁽²⁾ (V _{AI} < V _{SSAD} – 0.3 or V _{AI} > V _{CCAD} + 0.3)	–2	2	mA

(1) For V_{CCAD} and V_{SSAD} recommended operating conditions, see the "Device Recommended Operating Conditions" table.

(2) Input currents into any ADC input channel outside the specified limits could affect conversion results of other channels.

Table 14. Operating Characteristics over Full Ranges of Recommended Operating Conditions^{(1)(2)(3) (4)}

PARAMETER		DESCRIPTION/CONDITIONS		MIN	TYP	MAX	UNIT
R _I	Analog input resistance	See Figure 24.		250		500	Ω
C _I	Analog input capacitance	See Figure 24.	Conversion			10	pF
			Sampling			30	pF
I _{AIL}	Analog input leakage current	See Figure 24.		–1		1	μA
I _{ADREFHI}	AD _{REFHI} input current	AD _{REFHI} = 3.6 V, AD _{REFLO} = V _{SSAD}				5	mA
CR	Conversion range over which specified accuracy is maintained	AD _{REFHI} - AD _{REFLO}		3		3.6	V
E _{DNL}	Differential nonlinearity error	Difference between the actual step width and the ideal value. See Figure 25.				±1.5	LSB
E _{INL}	Integral nonlinearity error	Maximum deviation from the best straight line through the MibADC. MibADC transfer characteristics, excluding the quantization error. See Figure 26.				±2	LSB
E _{TOT}	Total error/Absolute accuracy	Maximum value of the difference between an analog value and the ideal midstep value. See Figure 27.				±2.5	LSB

(1) Not production tested.

(2) INL and DNL values are valid for a max ADCCLK frequency of 15 MHz. For frequencies greater than 15 MHz missing codes are expected at higher temperature.

(3) V_{CCAD} = AD_{REFHI}

(4) 1 LSB = (AD_{REFHI} – AD_{REFLO})/2¹⁰ for the MibADC

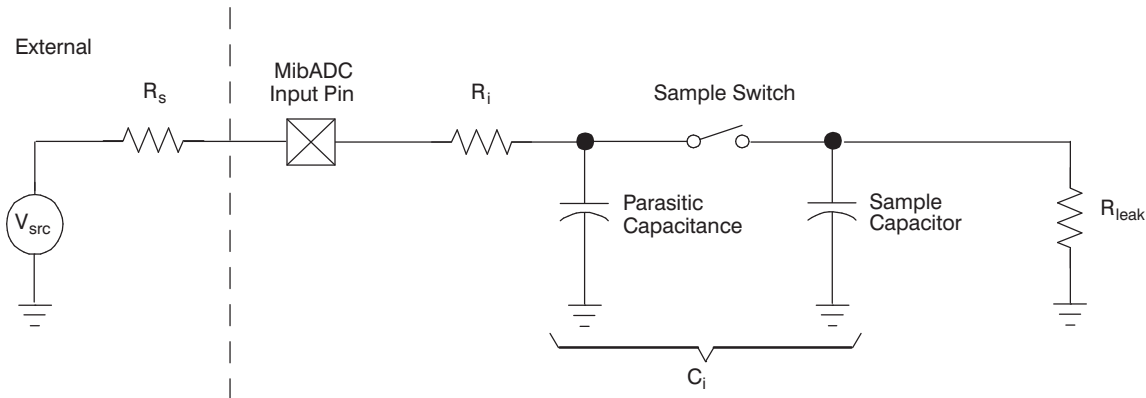


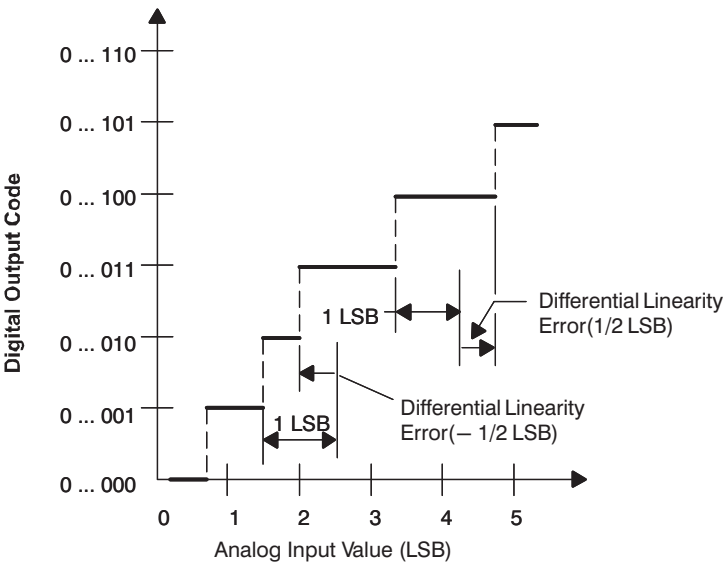
Figure 24. MibADC Input Equivalent Circuit

Table 15. Multi-Buffer ADC Timing Requirements⁽¹⁾

		MIN	NOM	MAX	UNIT
$t_{c(ADCLK)}$	Cycle time, MibADC clock	0.067			μ s
$t_{d(SH)}$	Delay time, sample and hold time	1			μ s
$t_{d@}$	Delay time, conversion time	0.55			μ s
$t_{d(SHC)}^{(2)}$	Delay time, total sample/hold and conversion time	1.55			μ s

- (1) Not production tested.
(2) This is the minimum sample/hold and conversion time that can be achieved. These parameters are dependent on many factors; for more details, see the *TMS470R1x Multi-Buffered Analog-to-Digital Converter (MibADC) Reference Guide* (literature number SPNU206).

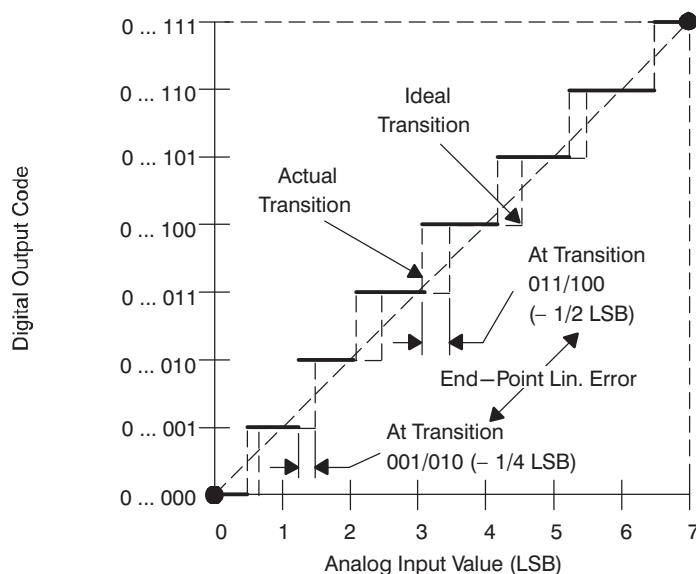
The differential nonlinearity error shown in Figure 25 (sometimes referred to as differential linearity) is the difference between an actual step width and the ideal value of 1 LSB.



A. $1 \text{ LSB} = (AD_{REFHI} - AD_{REFLO})/2^{10}$

Figure 25. Differential Nonlinearity (DNL)

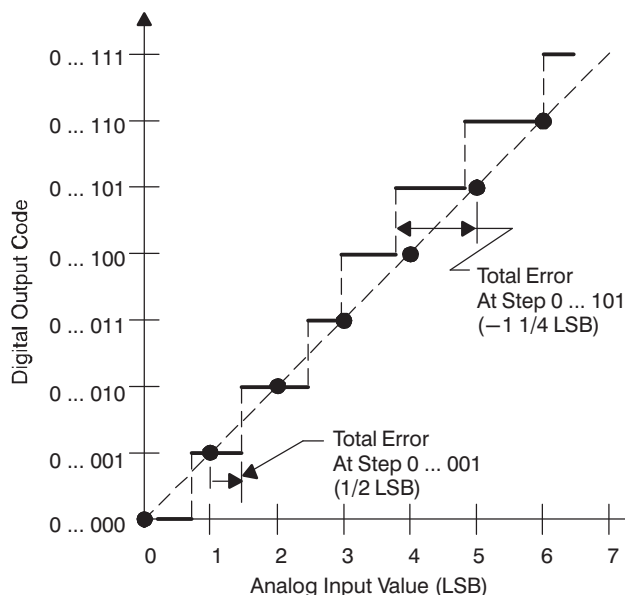
The integral nonlinearity error shown in Figure 26 (sometimes referred to as linearity error) is the deviation of the values on the actual transfer function from a straight line.



A. $1 \text{ LSB} = (AD_{REFHI} - AD_{REFLO})/2^{10}$

Figure 26. Integral Nonlinearity (INL) Error

The absolute accuracy or total error of an MibADC as shown in Figure 27 is the maximum value of the difference between an analog value and the ideal midstep value.



A. $1 \text{ LSB} = (AD_{REFHI} - AD_{REFLO})/2^{10}$

Figure 27. Absolute Accuracy (Total) Error

REVISION HISTORY

Changes from Revision E (July 2012) to Revision F	Page
• Changed HFQ package ordering information	3
• Added HKP package ordering information and corresponding data throughout data sheet	3

PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead/Ball Finish	MSL Peak Temp (3)	Op Temp (°C)	Top-Side Markings (4)	Samples
SM470R1B1MHFQS	ACTIVE	CFP	HFQ	84	1	TBD	AU	N / A for Pkg Type	-55 to 220	SM470R1B1 MHFQS	Samples
SM470R1B1MHKPS	ACTIVE	CFP	HKP	84	1	TBD	Call TI	N / A for Pkg Type	-55 to 220	SM470R1B1 MHKPS	Samples
SM470R1B1MKGDS1	ACTIVE	XCEPT	KGD	0	36	TBD	Call TI	N / A for Pkg Type	-55 to 220		Samples

⁽¹⁾ The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

⁽²⁾ Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check <http://www.ti.com/productcontent> for the latest availability information and additional product content details.

TBD: The Pb-Free/Green conversion plan has not been defined.

Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

Pb-Free (RoHS Exempt): This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

⁽³⁾ MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

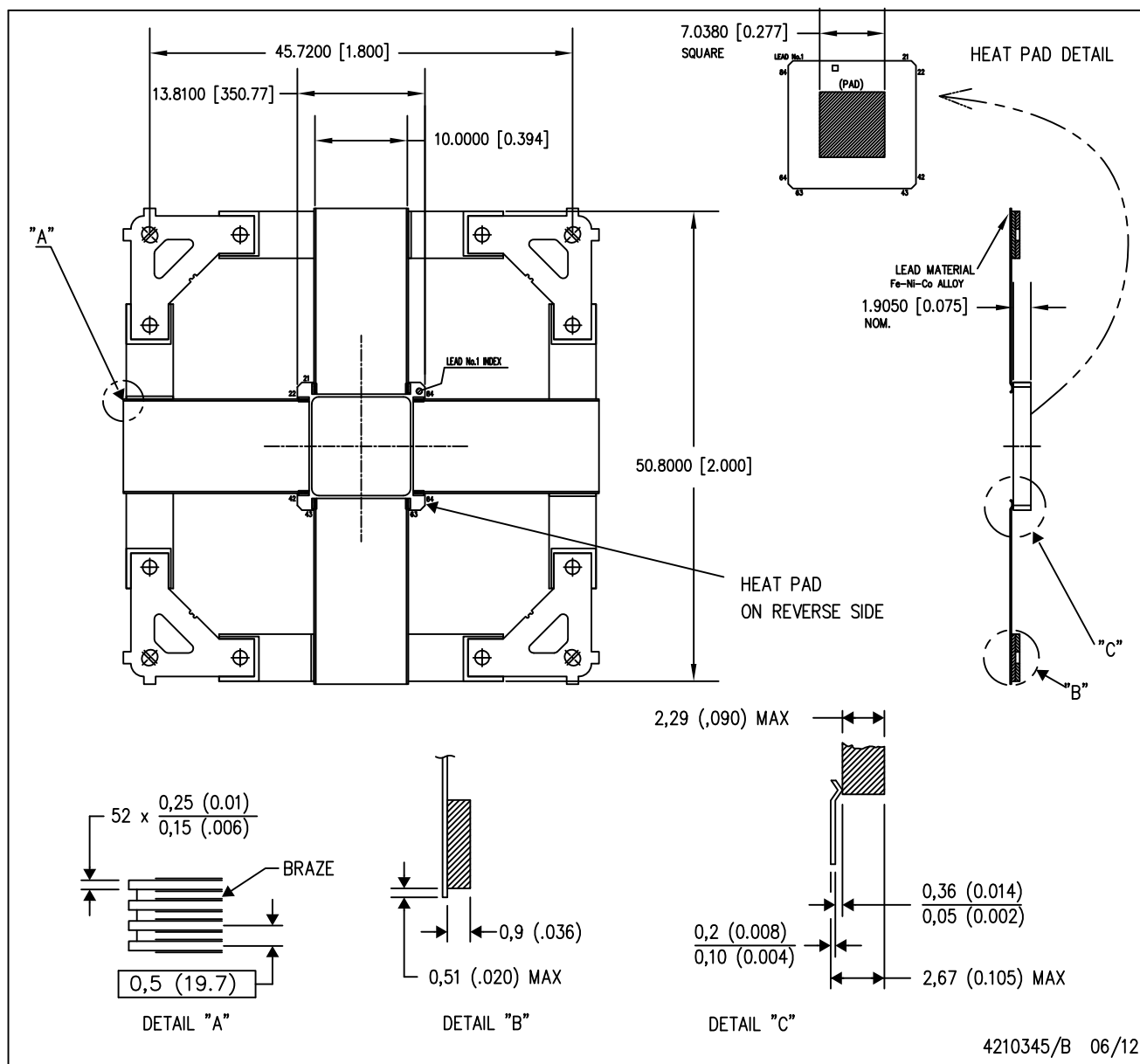
⁽⁴⁾ Only one of markings shown within the brackets will appear on the physical device.

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HFQ (S-CQFP-F84)

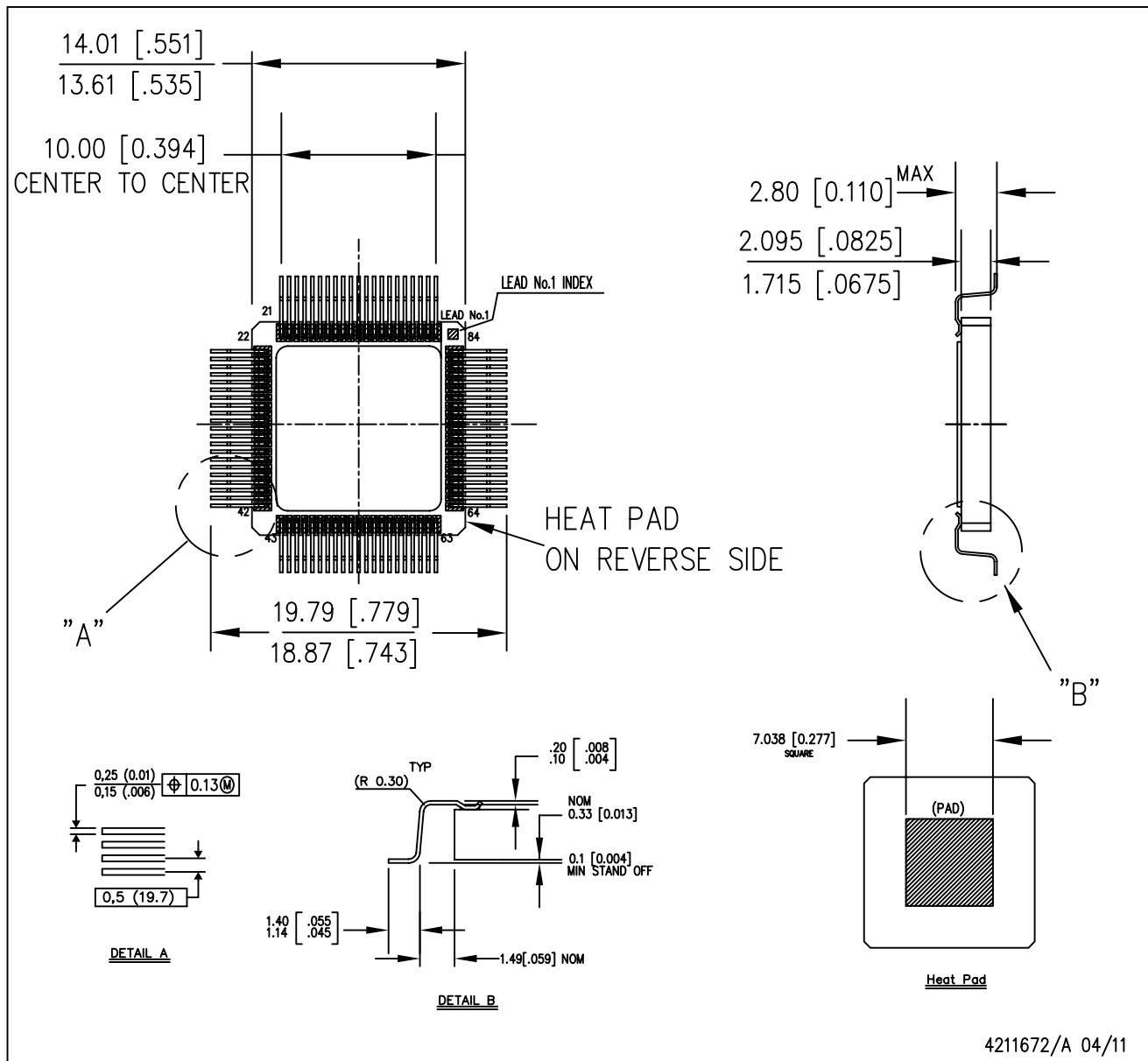
CERAMIC QUAD FLATPACK WITH NCTB



- NOTES:
- All linear dimensions are in millimeters (inches).
 - This drawing is subject to change without notice.
 - Ceramic quad flatpack with flat leads brazed to non-conductive tie bar carrier.
 - This package is hermetically sealed with a metal lid.
 - The leads are gold plated and can be solderdipped.
 - Leads not shown for clarity purposes.
 - Lid and heat sink are connected to GND leads.

HKP (S-CQFP-G84)

CERAMIC QUAD FLATPACK – GULL WING



- NOTES:
- A. All linear dimensions are in millimeters (inches).
 - B. This drawing is subject to change without notice.
 - C. Ceramic quad flatpack with formed leads.
 - D. This package is hermetically sealed with a metal lid.
 - E. The leads are gold plated and can be solderdipped.
 - F. Lid and heat pad are electrically connected.

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