

Ordering number : EN4391A

Thick Film Hybrid IC

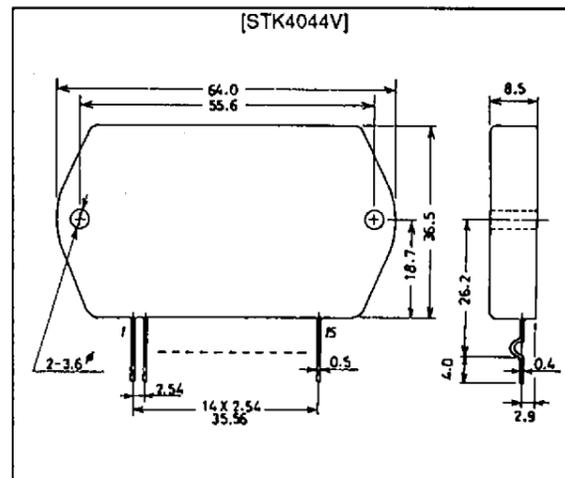
SANYO	No. 4391A	STK4044V
	AF Power Amplifier (Split Power Supply) (100 W min, THD = 0.08 %)	

Features

- Compact packaging supports slimmer set designs
- Series designed from 20 up to 100 W (200 W) and pin-compatibility (120 to 200 W have 18 pins)
- Simpler heat sink design facilitates thermal design of slim stereo sets
- Current mirror circuit application reduces distortion to 0.08%
- Supports addition of electronic circuits for thermal shutdown and load-short protection circuit as well as pop noise muting which occurs when the power supply switch is turned on and off

Package Dimensions

unit : mm
4075



Specifications

Maximum Ratings at Ta = 25°C

			Unit
Maximum supply voltage	V _{CC} max	±73	V
Thermal resistance	θ _{j-c}	1.1	°C/W
Junction temperature	T _j	150	°C
Operating substrate temperature	T _c	125	°C
Storage temperature	T _{sig}	-30 to +125	°C
Available time for load shorted	t _s *1	V _{CC} =±51V, R _L =8Ω, f=50Hz, P _O =100W	1 s

Recommended Operating Conditions at Ta = 25°C

			Unit
Recommended supply voltage	V _{CC}	±51	V
Load resistance	R _L	8	Ω

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Operating Characteristics

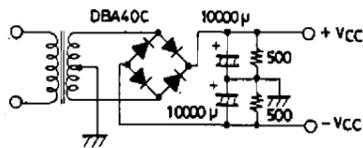
at $T_a = 25^\circ\text{C}$, $V_{CC} = \pm 51\text{V}$, $R_L = 8\Omega$, $V_G = 40\text{dB}$, $R_g = 600\Omega$, 100k LPF ON, R_L (noninductive)

			min	typ	max	Unit
Quiescent current	I_{CCO}	$V_{CC} = \pm 61\text{V}$	15		120	mA
Output power	P_O	THD = 0.08%, $f = 20\text{Hz to } 20\text{kHz}$	100			W
Total harmonic distortion	THD	$P_O = 1.0\text{W}$, $f = 1\text{kHz}$			0.08	%
Frequency response	f_L, f_H	$P_O = 1.0\text{W}$, $+0_{-3}\text{dB}$		20 to 50k		Hz
Input resistance	r_i	$P_O = 1.0\text{W}$, $f = 1\text{kHz}$		55		k Ω
Output noise voltage	V_{NO}^{*2}	$V_{CC} = \pm 61\text{V}$, $R_g = 10\text{k}\Omega$			1.2	mVrms
Neutral voltage	V_N	$V_{CC} = \pm 61\text{V}$	-70	0	+70	mV

• Use rated power supply for test unless otherwise specified.

*1 When measuring available time for load shorted and output noise voltage, use transformer power supply indicated below.

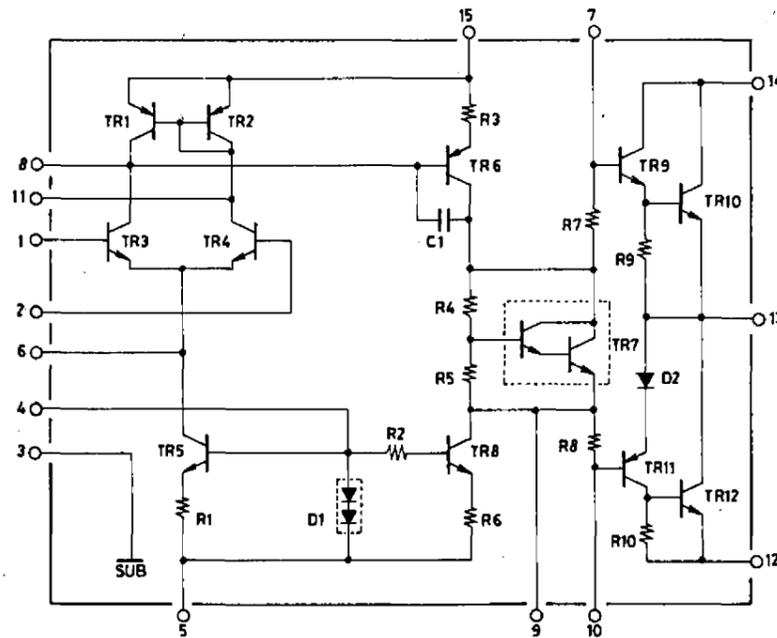
*2 Output noise voltage represents the peak value on the rms scale (VTVM). The noise voltage waveform does not include the pulse noise.



**Specified Transformer Power Supply
(MG-200 Equivalent)**

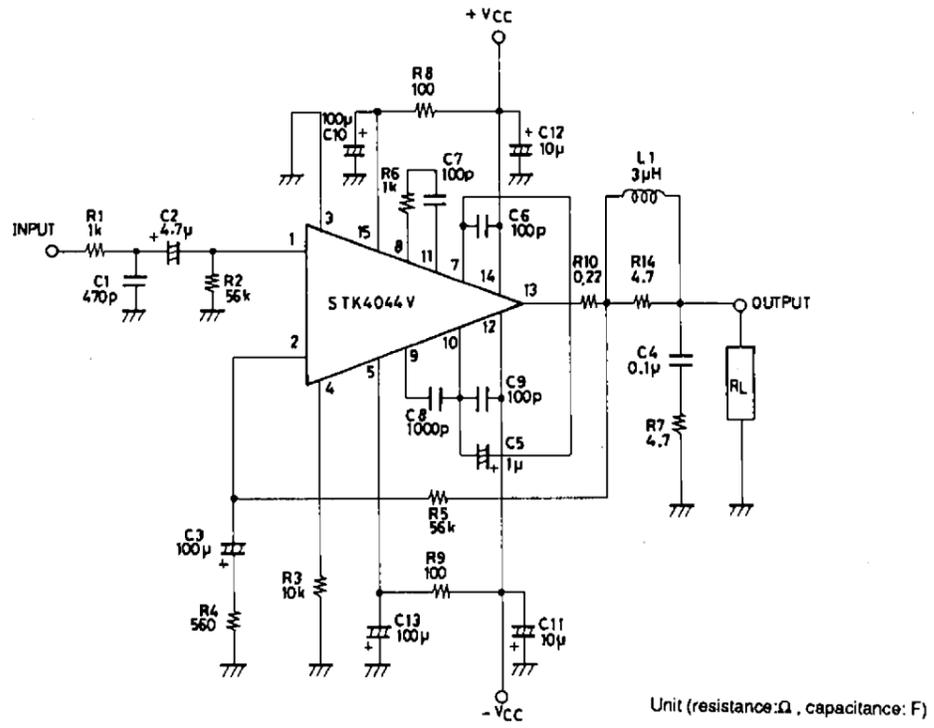
Unit (resistance: Ω , capacitance: F)

Equivalent Circuit

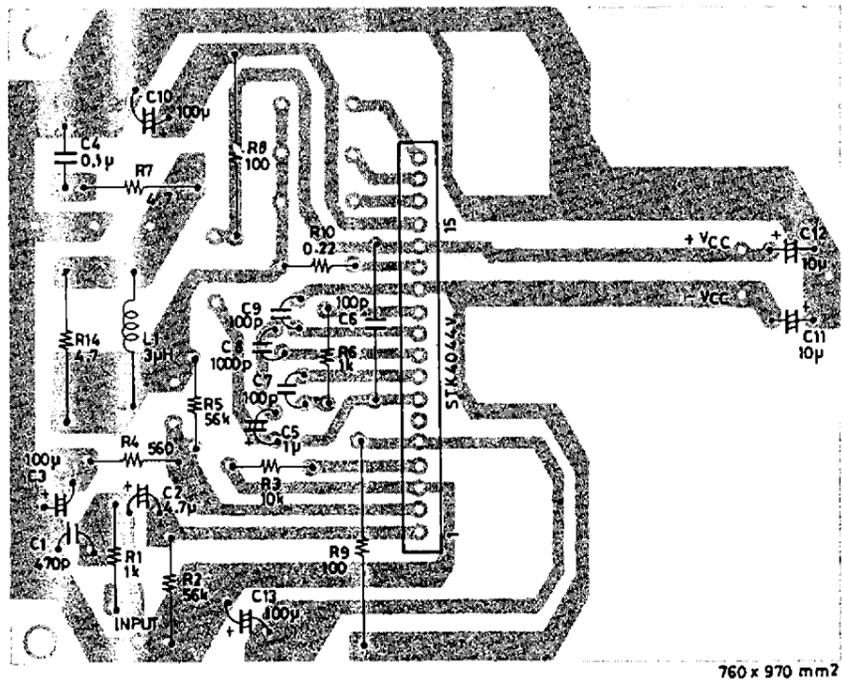


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Sample Application Circuit: 100W min Single-Channel AF Power Amplifier

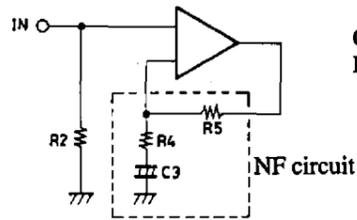


Sample Printed Circuit Pattern for Application Circuit (Copper-folled side)



Description of External Parts

- R₁, C₁ : Input filter circuit
 - Reduces high-frequency noise.
- C₂ : Input coupling capacitor
 - DC current suppression. A reduction in reactance is effective because of increases in capacitor reactance at low frequencies and 1/f noise dependence on signal source resistance which result in output noise worsening.
- R₂ : Input bias resistor
 - Biases the input pin to zero.
 - Affects V_N stability (refer to NF circuit).
 - Due to differential input, input resistance is more or less determined by this resistance value.
- R₄, R₅ : NFB circuit (AC NF circuit). Use of resistor with 1% error is suggested.
- C₃ (R₂)



- C₃ : AC NF capacitor
- R₄, R₅ : Used for VG setting.

- VG settings are obtained using R₄ and R₅ according to the following equation:

$$\log_{20} \frac{R_5}{R_4} \quad 40 \text{ dB is recommended.}$$

- Low-frequency cutoff frequency settings are obtained using R₄ and C₃ according to the following equation:

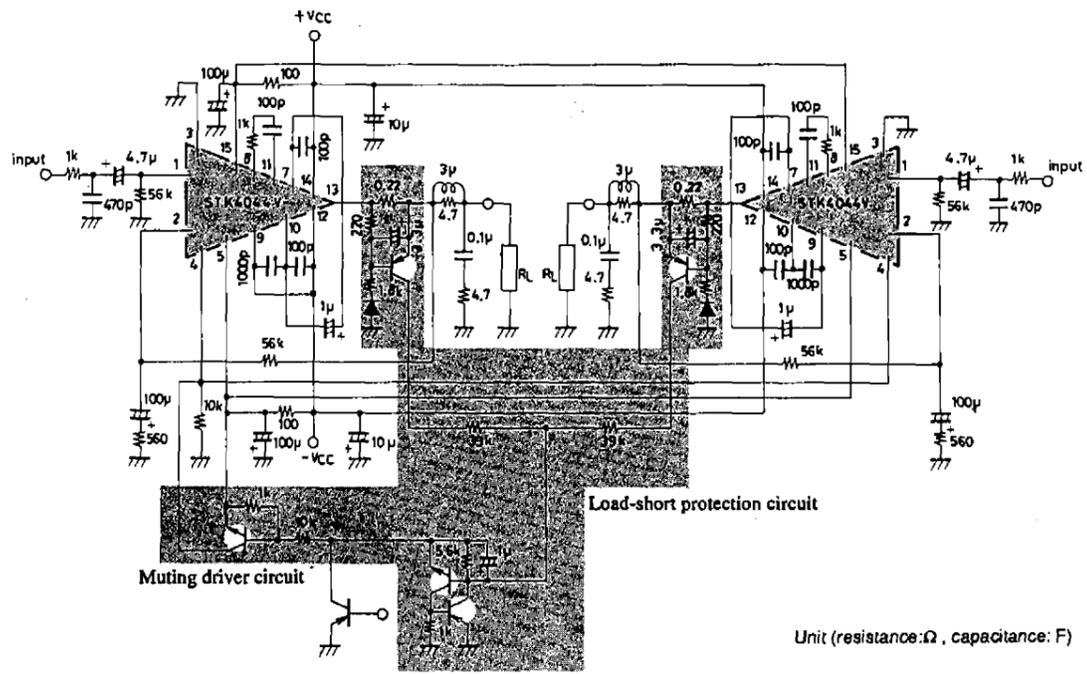
$$f_L = \frac{1}{2\pi \cdot R_4 \cdot C_3} \quad [\text{Hz}]$$

When changing the VG setting, you should change R₄ which requires a recheck of the low cutoff frequency setting. When the VG setting is changed using R₅, the setting should ensure R₂ equals R₅ so that V_N balance stability is maintained. If the resistor value is increased more than the existing value, V_N balance may be disturbed and result in deterioration of V_N temperature characteristics.

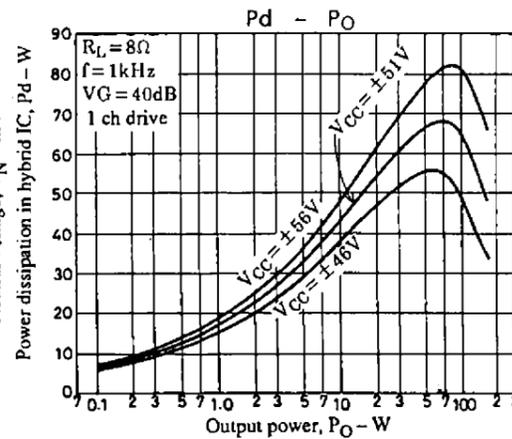
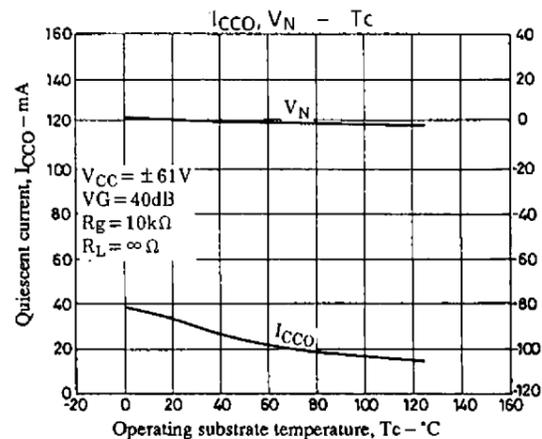
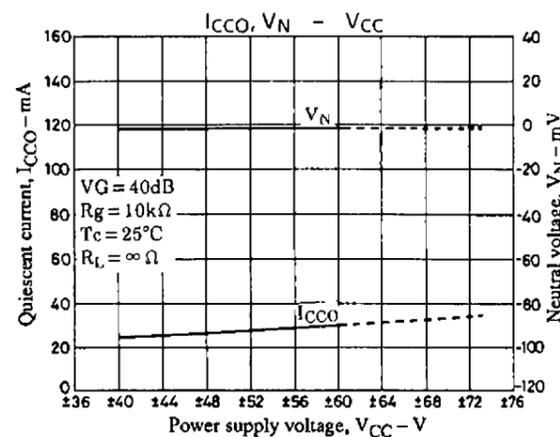
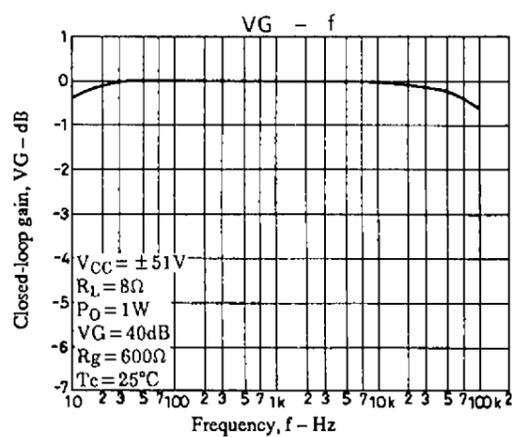
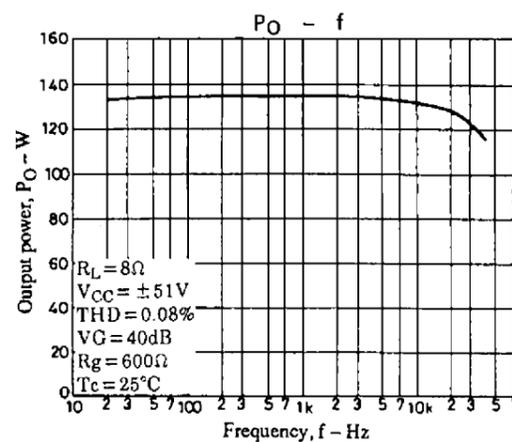
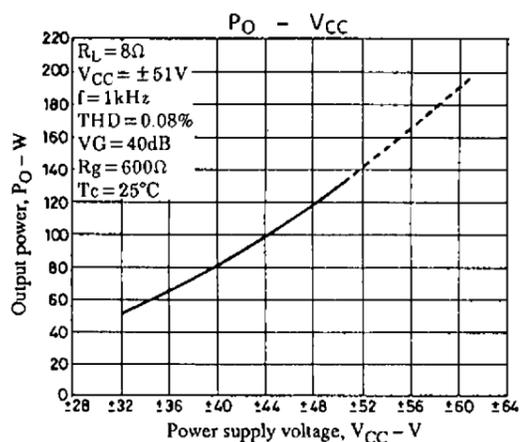
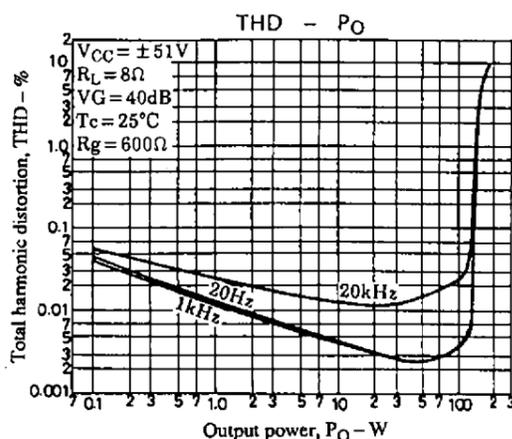
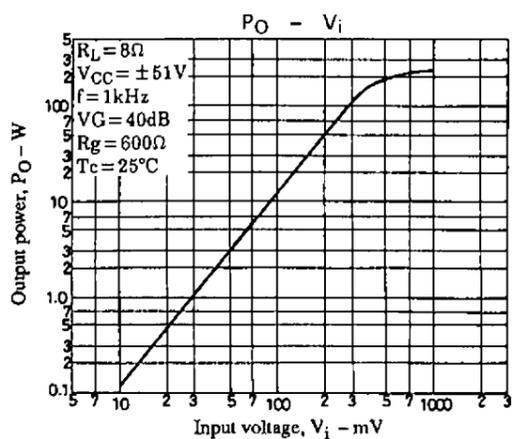
- R₃ : Differential constant-current bias resistor
- R₆, R₇ : For oscillation suppression and phase compensation applications (For use with differential stage applications)
- R₇, C₄ : For oscillation suppression and phase compensation applications (A Mylar capacitor is recommended for C₄ for use with output stage applications)
- C₆, C₉ : For oscillation suppression and phase compensation applications
Power stage (Must be connected near the pin) C₆: Positive (+) power C₉: Negative (-) power
- C₈ : For oscillation suppression and phase compensation applications (Oscillation suppression before power step clip)
- C₅ : For oscillation suppression and distortion improvement applications
- R₈, C₁₀ : Ripple filter circuit on positive (+) side.
- R₉, C₁₃ : Ripple filter circuit on negative (-) side.
- C₁₁, C₁₂ : For oscillation suppression applications
 - Used for reducing power supply impedance to stable IC operation and should be connected near the IC pin. We recommend that you use an electrolytic capacitor.
- R₁₀ : Output resistor
Increases load short handling capability during times of high output.
- R₁₄, L₁ : For oscillation suppression applications
Increases oscillation stability against capacitance loads.

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Sample Application Circuit (Protection circuit and muting circuit)



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